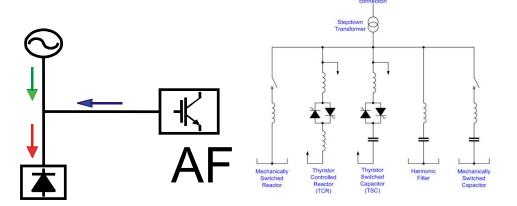
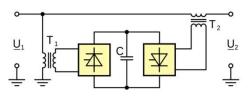
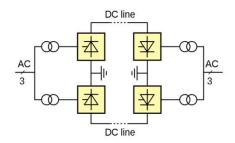
Grid Connected Power Electronics

Acronyms

- APF Active Power Filter
- UPFC Unified Power Flow Controller
- SVC Static Var Converter
- HVDC High Voltage Direct Current
- ...







All made to improve "Power Quality"

L11 - Static VAr compensation

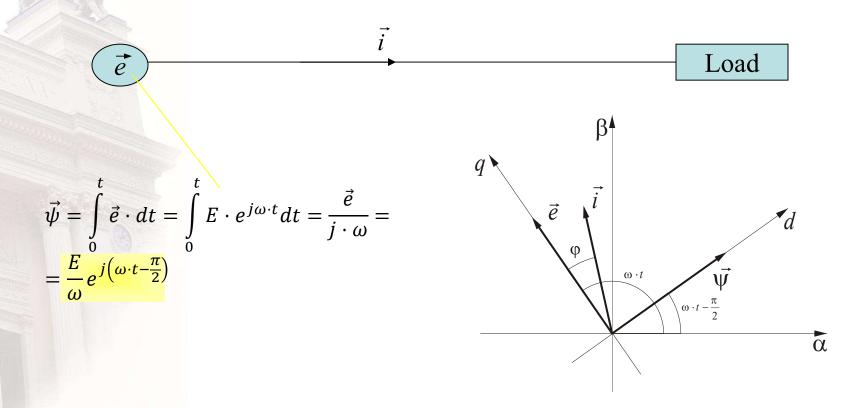
Non ideal loads

- are loads that:
 - are non-resistive -> consume reactive power
 - vary with time or phase -> consume harmonic current components.
 - are different in different phases
 consume negative sequence
 currents

Ways to improve the loads

- Self improvement
 - Solutions that draw as "ideal" current as possible from the grid
- Compensation
 - A parallel unit is used to counteract the non ideal currents drawn by the main load

Load model



L11 - Static VAr compensation

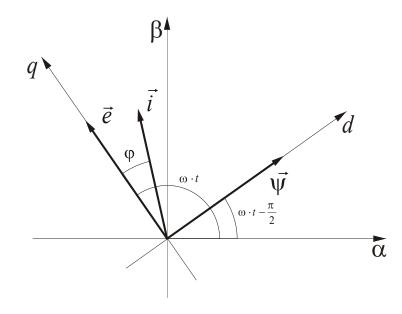
Reactive power

A phase lag between the voltage and the current:

$$\sqrt{\frac{3}{2}} \cdot \hat{\imath} \cdot e^{j(\omega t - \phi)}$$

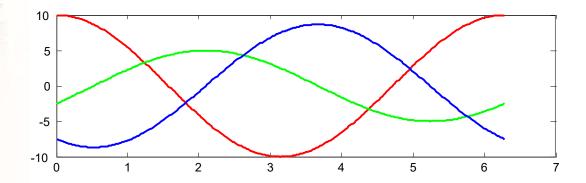
In flux coordinates:

$$\sqrt{\frac{3}{2}} \cdot \hat{\imath} \cdot e^{j\left(\frac{\pi}{2} - \phi\right)}$$



Assymetric load

 The phase currents are not equal in amplitude or phase lag ...



L11 - Static VAr compensation

An assymetric load current vector in the (α,β) -frame

$$\vec{l} = \sqrt{\frac{2}{3}} \cdot \left(\hat{l}_{a} \cdot \frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{a})}}{2} \cdot 1 + \hat{l}_{b} \cdot \frac{e^{j(\omega t - \phi_{b} - \frac{2\pi}{3})} + e^{-j(\omega t - \phi_{b} - \frac{2\pi}{3})}}{2} \cdot e^{j\frac{2\pi}{3}} + \hat{l}_{c} \cdot \frac{e^{j(\omega t - \phi_{c} - \frac{4\pi}{3})} + e^{-j(\omega t - \phi_{c} - \frac{4\pi}{3})}}{2} \cdot e^{j\frac{4\pi}{3}} \right) = \sqrt{\frac{2}{3}} \cdot \left(\hat{l}_{a} \cdot \frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{a})}}{2} + \hat{l}_{b} \cdot \frac{e^{j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{b} - \frac{4\pi}{3})}}{2} + \hat{l}_{c} \cdot \frac{e^{j(\omega t - \phi_{c})} + e^{-j(\omega t - \phi_{c} - \frac{2\pi}{3})}}{2} \right) = \sqrt{\frac{2}{3}} \cdot \left(\hat{l}_{a} \cdot \frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{a})}}{2} + \hat{l}_{b} \cdot \frac{e^{j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{b} - \frac{4\pi}{3})}}{2} \right) = \sqrt{\frac{2}{3}} \cdot \left(\hat{l}_{a} \cdot \frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{a})}}{2} + \hat{l}_{b} \cdot \frac{e^{j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{b} - \frac{4\pi}{3})}}{2} \right) = \sqrt{\frac{2}{3}} \cdot \left(\hat{l}_{a} \cdot \frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{a})}}{2} + \hat{l}_{b} \cdot \frac{e^{j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{b} - \frac{4\pi}{3})}}{2} \right) = \sqrt{\frac{2}{3}} \cdot \left(\hat{l}_{a} \cdot \frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{a})}}{2} + \hat{l}_{b} \cdot \frac{e^{j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{b})}}{2} + e^{-j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{b})} \right) = \sqrt{\frac{2}{3}} \cdot \left(\frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{a})}}{2} + e^{-j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{b})} \right) = \sqrt{\frac{2}{3}} \cdot \left(\frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{a})}}{2} + e^{-j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{b})} + e^{-j(\omega t - \phi_{b})} \right) + e^{-j(\omega t - \phi_{b})} \cdot \left(\frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{b})}}{2} \right) = \sqrt{\frac{2}{3}} \cdot \left(\frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{a})}}{2} + e^{-j(\omega t - \phi_{b})} \right) + e^{-j(\omega t - \phi_{b})} \cdot \left(\frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{b})}}}{2} \right) = \sqrt{\frac{2}{3}} \cdot \left(\frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{b})}}{2} \right) + e^{-j(\omega t - \phi_{b})} \cdot \left(\frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{b})}}{2} \right) + e^{-j(\omega t - \phi_{b})} \cdot \left(\frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{b})}}{2} \right) + e^{-j(\omega t - \phi_{b})} \cdot \left(\frac{e^{j(\omega t - \phi_{a})} + e^{-j(\omega t - \phi_{b})}}{2} \right) + e^{-j(\omega t - \phi_{b})} \cdot \left(\frac$$

$$= \begin{cases} \hat{\imath}_{x} = \hat{\imath}, \phi_{x} = \phi \to \sqrt{\frac{2}{3}} \cdot \frac{3 \cdot \hat{\imath}}{2} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{2}{3}} \cdot \frac{3 \cdot \hat{\imath}}{2} \left(e^{-j(\omega t - \phi)} + e^{-j\left(\omega t - \phi - \frac{4\pi}{3}\right)} + e^{-j\left(\omega t - \phi - \frac{2\pi}{3}\right)} \right) = \sqrt{\frac{3}{2}} \cdot \hat{\imath} \cdot e^{j(\omega t - \phi)} \\ \hat{\imath}_{x} = \hat{\imath}, \hat{\imath}_{c} \neq \hat{\imath}, \phi_{x} = \phi \to \sqrt{\frac{3}{2}} \cdot \hat{\imath} \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{2}{3}} \cdot (\hat{\imath}_{c} - \hat{\imath}) \cdot \frac{e^{j(\omega t - \phi_{c})} + e^{-j\left(\omega t - \phi_{c} - \frac{2\pi}{3}\right)}}{2} \end{cases}$$

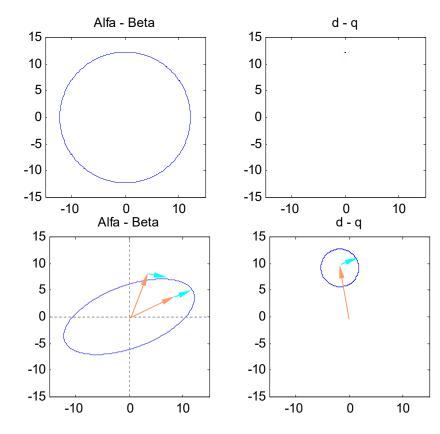
An assymetric load current vector in the (d,q)-frame

$$\vec{\iota}^{dq} = \vec{\iota}^{\alpha\beta} \cdot e^{-j\left(\omega t \frac{\pi}{2}\right)} = \begin{cases} = \\ \\ \hat{\iota}_{x} = \hat{\iota}, \phi_{x} = \phi \rightarrow \sqrt{\frac{3}{2}} \cdot \hat{\iota} \cdot e^{j(\omega t - \phi)} \cdot e^{-j\left(\omega t - \frac{\pi}{2}\right)} \\ \\ \hat{\iota}_{x} = \hat{\iota}, \hat{\iota}_{c} \neq \hat{\iota}, \phi_{x} = \phi \rightarrow \sqrt{\frac{3}{2}} \cdot \hat{\iota} \cdot e^{j(\omega t - \phi)} \cdot e^{-j\left(\omega t - \frac{\pi}{2}\right)} + \sqrt{\frac{2}{3}} \cdot (\hat{\iota}_{c} - \hat{\iota}) \cdot \frac{e^{j(\omega t - \phi_{c})} + e^{-j\left(\omega t - \phi_{c} - \frac{2\pi}{3}\right)}}{2} \cdot e^{-j\left(\omega t \frac{\pi}{2}\right)} \\ = \begin{cases} \frac{3}{2} \cdot \hat{\iota} \cdot e^{j\left(\frac{\pi}{2} - \phi\right)} \\ \sqrt{\frac{3}{2}} \cdot \hat{\iota} \cdot e^{j\left(\frac{\pi}{2} - \phi\right)} \\ \sqrt{\frac{3}{2}} \cdot \hat{\iota} \cdot e^{j\left(\frac{\pi}{2} - \phi\right)} \end{cases} + \underbrace{e^{-j\left(2\omega t - \phi_{c} - \frac{2\pi}{3} - \frac{\pi}{2}\right)}}_{Rotating} \\ \frac{Rotating}{backwards} \\ \frac{Botaking}{with 2\omega!} \end{cases}$$

M-file "Load current vectors" - demo

Symmetric load

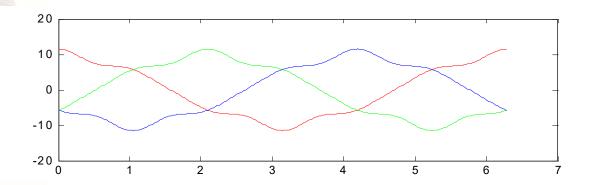
Assymetric



L11 - Static VAr compensation

Harmonics

Non linear load impedance

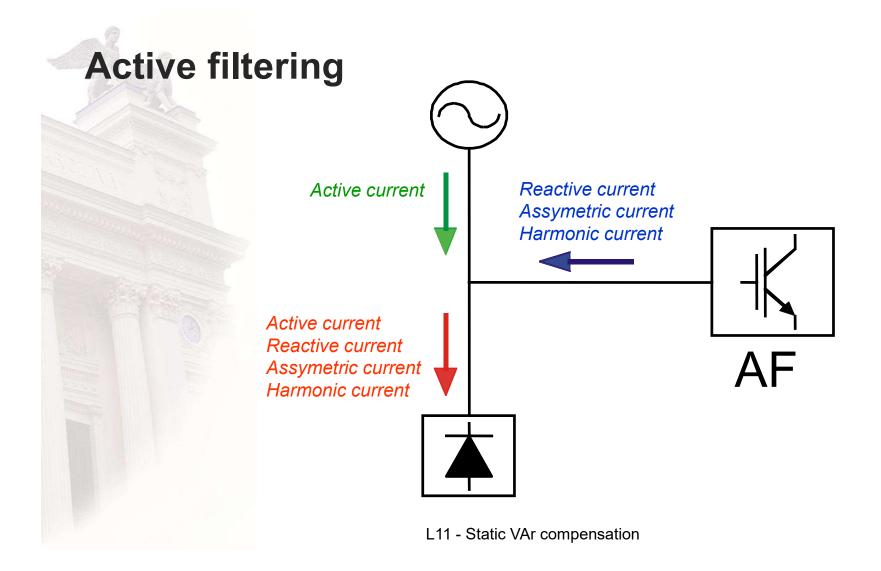


L11 - Static VAr compensation

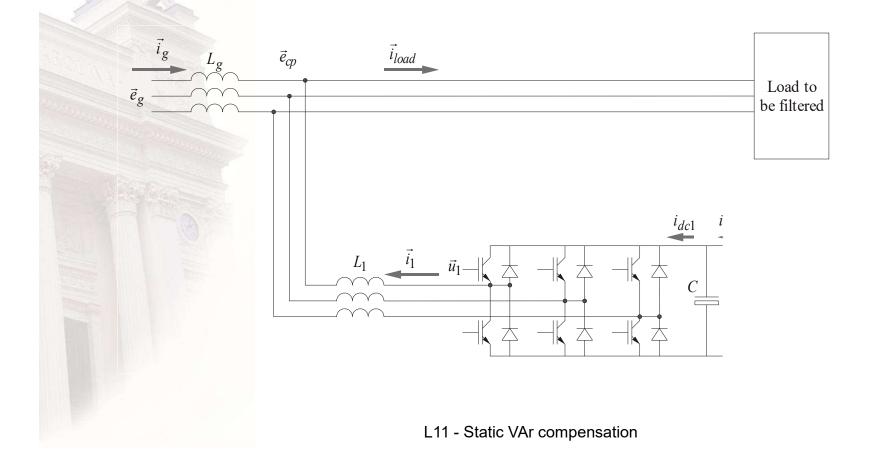
5'th and 7'th harmonic example with Simulink

$$\vec{\imath}^{\alpha\beta} = \sqrt{\frac{3}{2}} \cdot \hat{\imath}_1 \cdot e^{j(\omega t - \phi)} + \sqrt{\frac{3}{2}} \cdot \hat{\imath}_5 \cdot e^{j(-5\omega t - \cdot)} + \sqrt{\frac{3}{2}} \cdot \hat{\imath}_7 \cdot e^{j(7\omega t - \phi)}$$

$$\vec{\iota}^{dq} = \sqrt{\frac{3}{2}} \cdot \hat{\iota}_1 \cdot e^{j\left(\frac{\pi}{2} - \phi\right)} + \sqrt{\frac{3}{2}} \cdot \hat{\iota}_5 \cdot e^{j(-6\omega t - \cdot)} + \sqrt{\frac{3}{2}} \cdot \hat{\iota}_7 \cdot e^{j(6\omega t - \phi)}$$



Shunt Active Filter

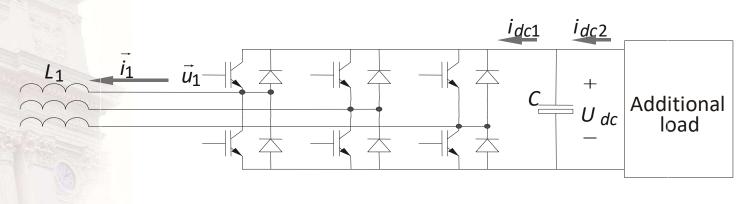


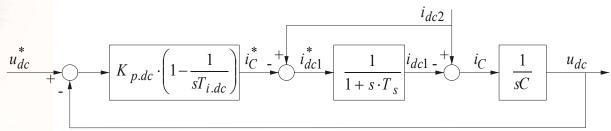
AC side Current Control

 Vector Control with Field Orientation

$$\vec{u}_{1}^{*}(k) = \left(\frac{L_{1}}{T_{s}} + \frac{R_{1}}{2}\right) \cdot \left(\left(\vec{i}_{1}^{*}(k) - \hat{\vec{i}}_{1}(k)\right) + \frac{T_{s}}{\left(\frac{L}{R} + \frac{T_{s}}{2}\right)} \cdot \sum_{n=0}^{n=k-1} \left(\vec{i}_{1}^{*}(n) - \hat{\vec{i}}_{1}(n)\right)\right) + \hat{\vec{e}}_{cp}(k)$$

DC link Voltage Control System





L11 - Static VAr compensation

Controller Parameters ...

Use Symmetric Optimum

$$\zeta = \frac{a-1}{2}$$

$$T_{i.dc} = a^2 \cdot T_s$$
, where $a > 1$

$$K_{p.dc} = \frac{a \cdot C}{T_{i.dc}}$$

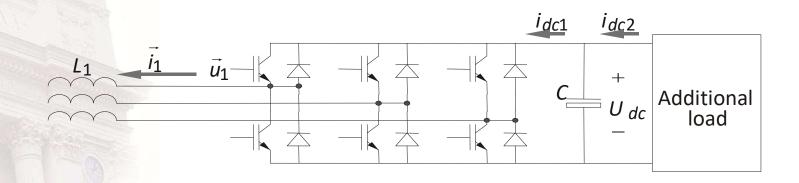
L11 - Static VAr compensation

Convert DC to AC current references

$$p(t) = Ri_{1d}^{2} + Ri_{1q}^{2} + L\frac{di_{1d}}{dt}i_{1d} + L\frac{di_{1q}}{dt}i_{1q} + e_{cp,q}i_{1q} = u_{dc} \cdot i_{dc1} \approx e_{cp,q}i_{1q}$$

$$\downarrow i_{dc1} = \frac{e_{cp,q}}{u_{dc}} \cdot i_{1q}$$

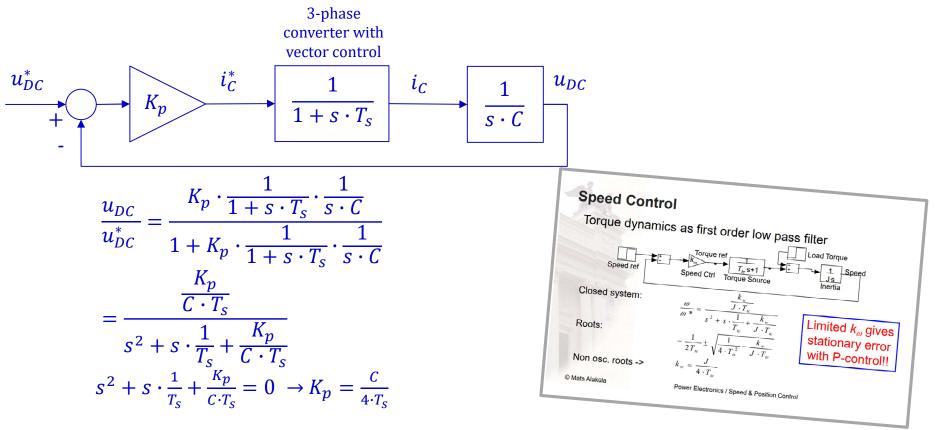
DC link voltage controller



$$i_{1q}^* = \frac{u_{dc}}{e_{cp}} \left(i_{dc2} - K_{p.dc} \cdot \left(1 - \frac{1}{sT_{i.dc}} \right) \cdot (u_{dc}^* - u_{dc}) \right)$$

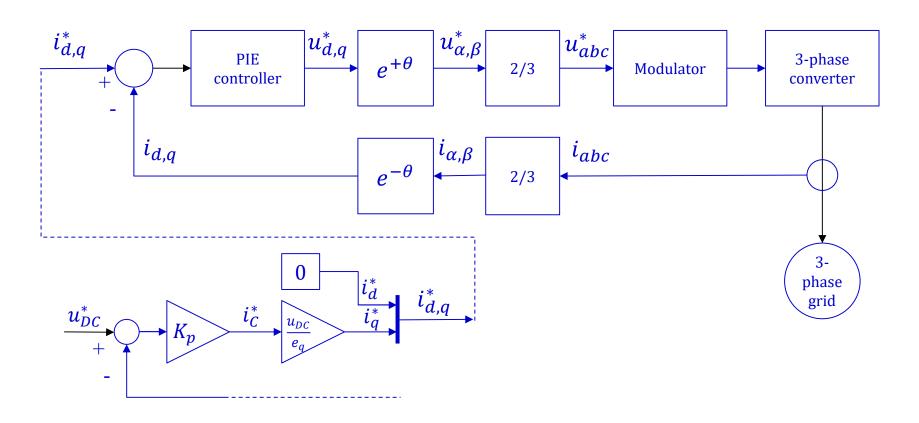
L11 - Static VAr compensation

DC Linc Voltage Control vs Speed Control



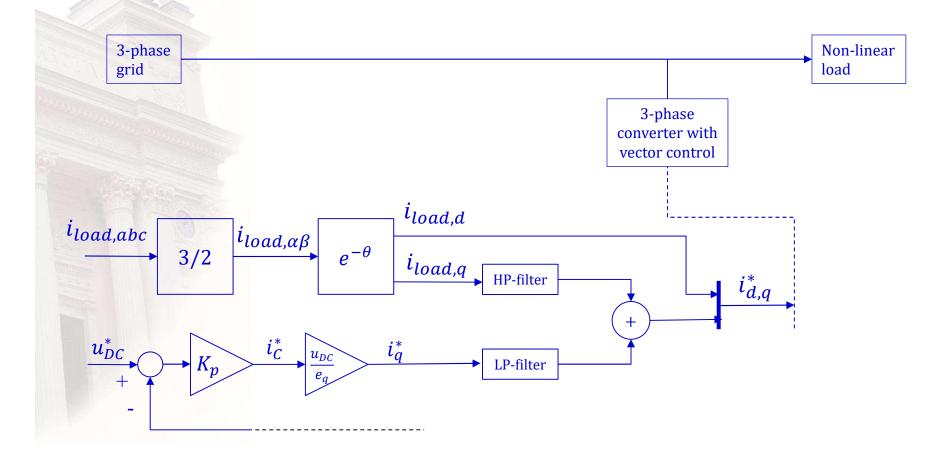
L11 - Static VAr compensation

The 3phase Grid connected converter is the DC link current source

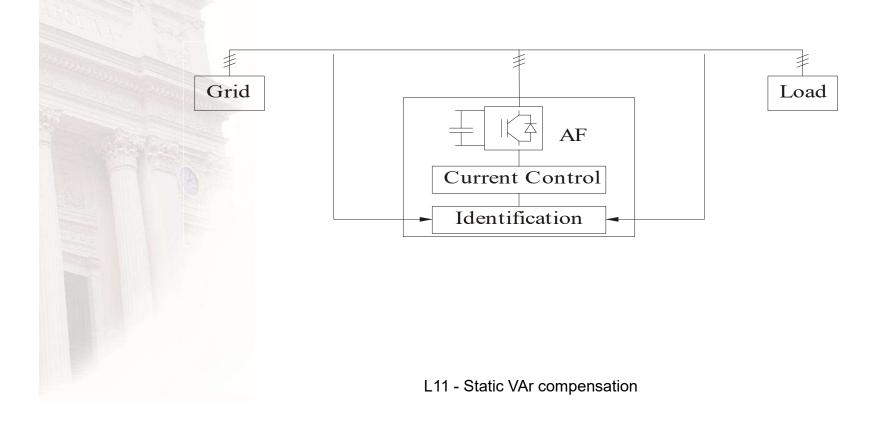


L11 - Static VAr compensation

The full control system



Active filter control



Example of DC voltage control

```
>> L=0.01;

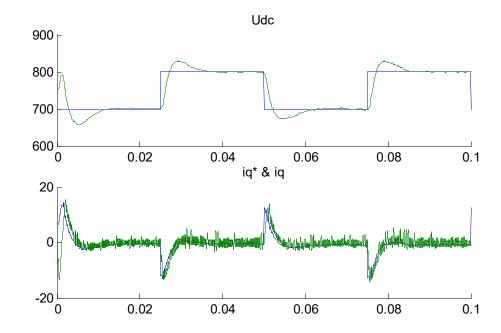
>> R=1;

>> Ts=0.0005;

>> Tidc=9*Ts;

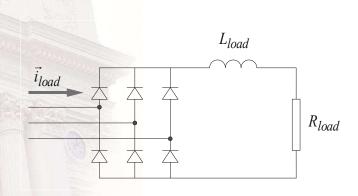
>> Kpdc=3*Cdc/Tidc;

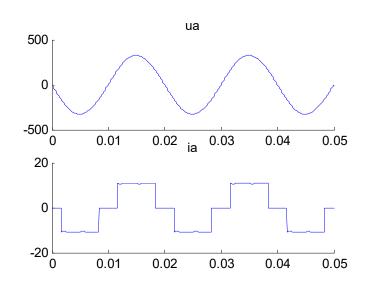
>> Cdc=1e-4;
```



L11 - Static VAr compensation

Example with active filtering





Filter Current References

$$i_{d,ActiveFilter}^* = i_{d,load}$$

$$i_{q,ActiveFilter}^* = i_{q,load} \cdot \frac{s \cdot T_f}{1 + s \cdot T_f} + \frac{u_{dc}}{e_{cp}} \left(i_{dc2} - K_{p.dc} \cdot \left(1 - \frac{1}{sT_{i.dc}} \right) \cdot \left(u_{dc}^* - u_{dc} \right) \right) \cdot \frac{1}{1 + s \cdot T_f}$$

Filter Currents

```
>> L=0.01;

>> R=1;

>> Ts=0.00005;

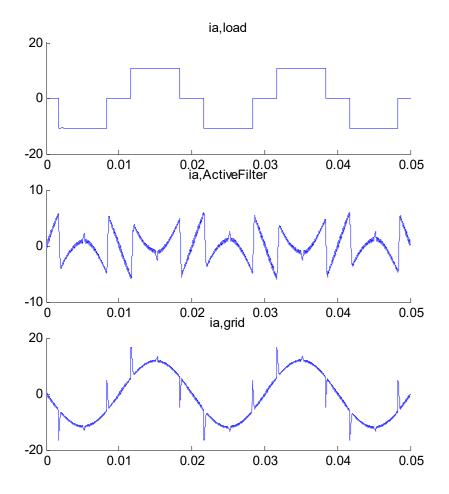
>> Rload=50;

>> Lload=0.1;

>> Tf=10e-3;

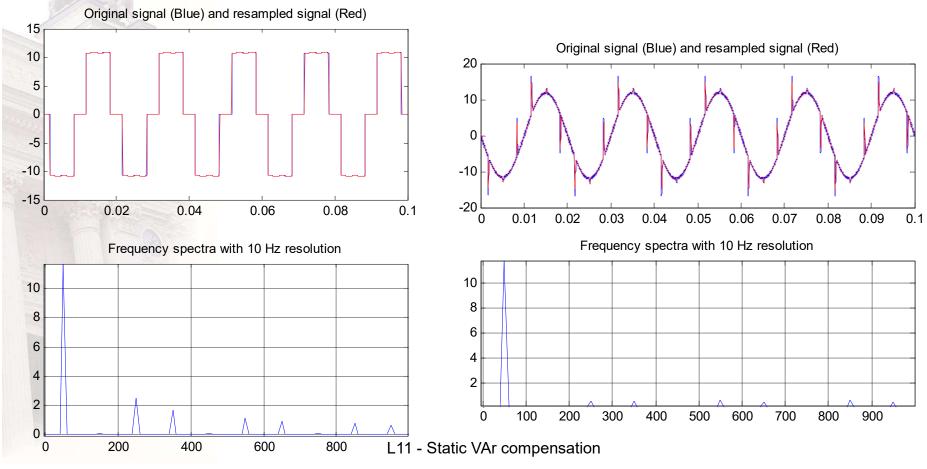
>> Tidc=9*Tf;

>> Kpdc=3*Cdc/Tidc;
```



L11 - Static VAr compensation

Spectra



Exercises

- 3-phase vectors
 - 1.11, 1.12, 1.13, 1,14
- 3-phase current control
 - 2.4