

Hybrid Drive Systems for Vehicles

- L5
 - Alternative drive train Components





Drawback with conventional drivetrains

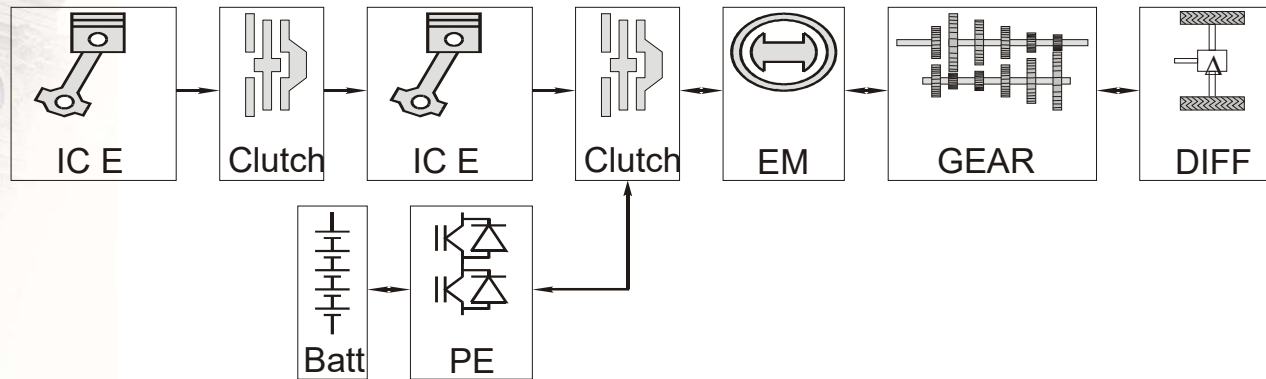
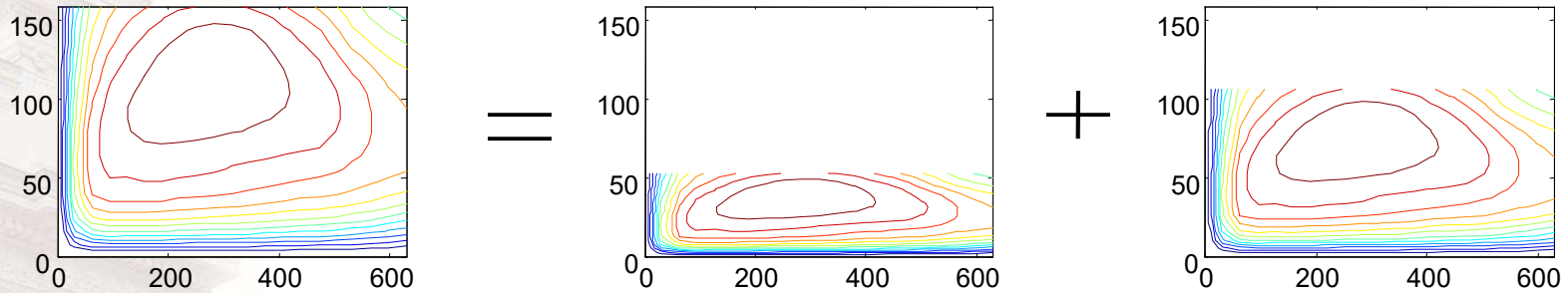
- Limited ability to optimize operating point
- No ability to regenerate braking power



Solutions

- Smaller Engine
 - + Reach higher operating points
 - Still cannot regenerate
- Store energy in the vehicle mass
 - Speed variations
 - Still cannot regenerate
- Secondary energy storage
 - + Selectable operating point ($P_{\text{storage}} = P_{\text{ice}} - P_{\text{road}}$)
 - + Can regenerate
 - Expensive

Smaller Engine – cylinder deactivation

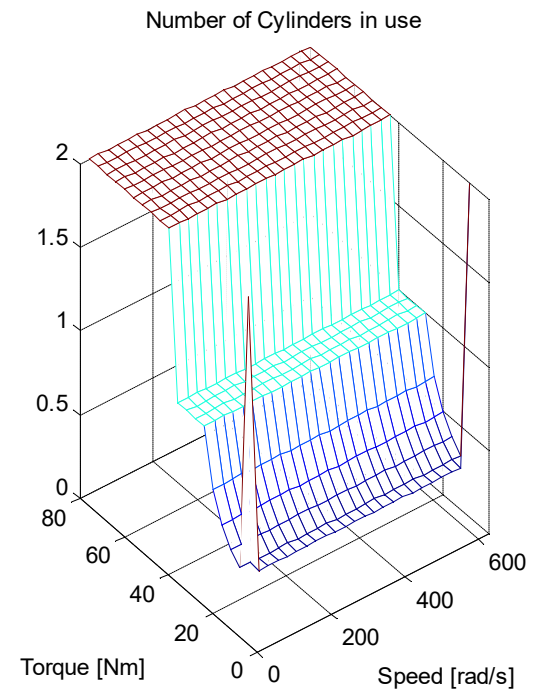
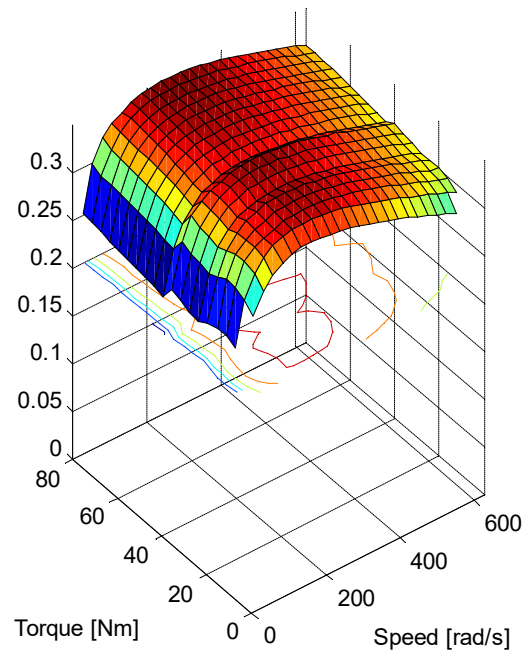
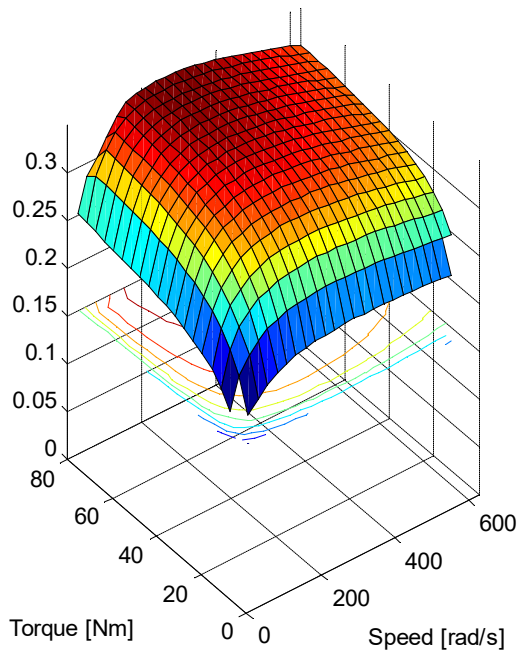


Combustion On Demand

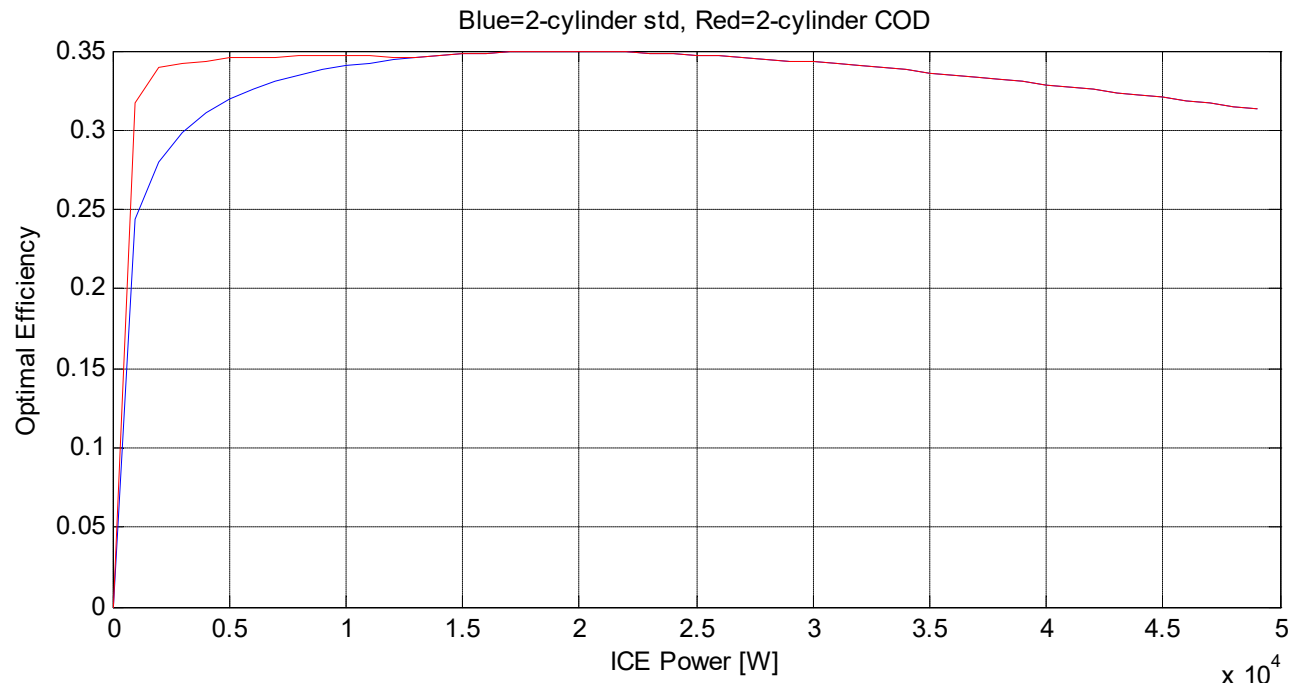
- Cylinder deactivation via free valve control
- 4-stroke, 6-stroke, 8-stroke ...



Optimized efficiency



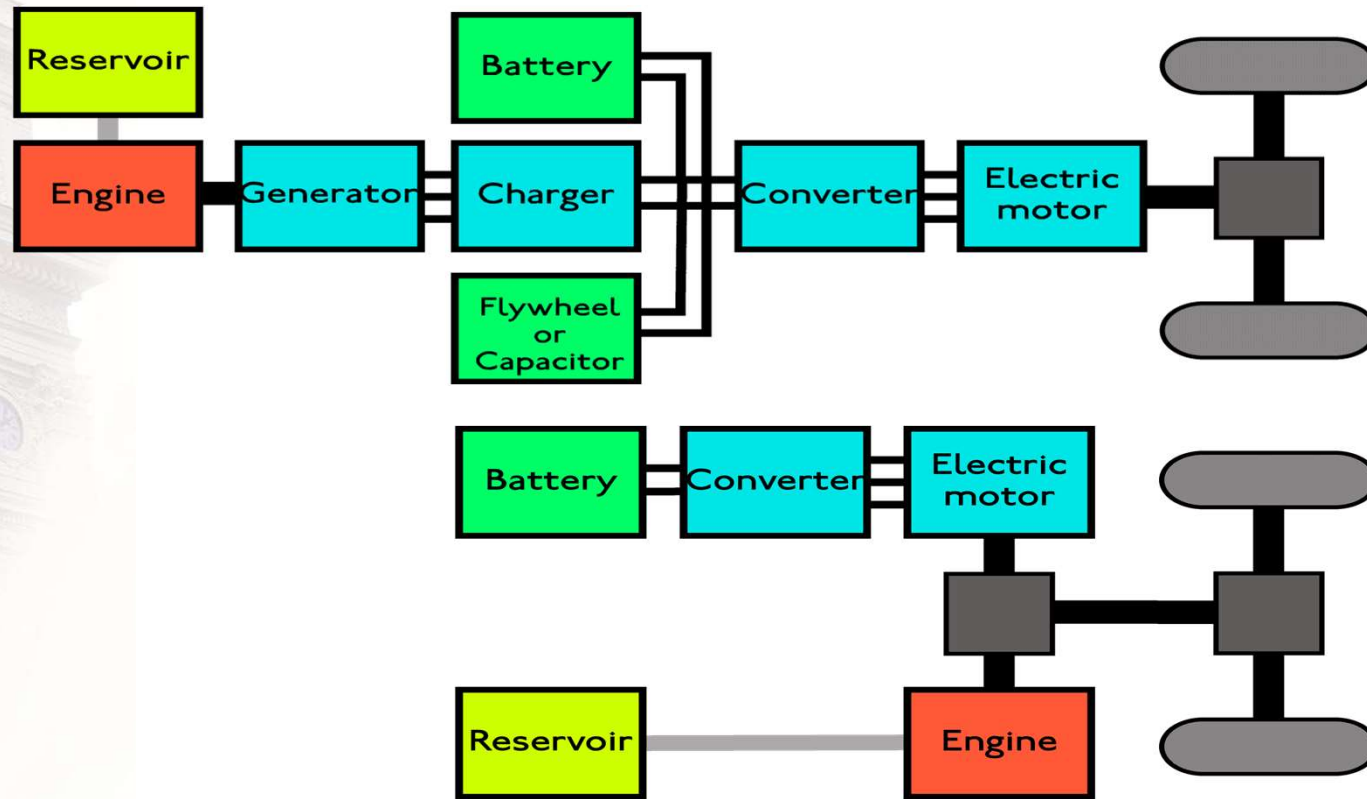
Max efficiency



Secondary Energy Storage selection

- Why electric?
 - Efficient secondary energy converters (Electrical machines)
 - Safe, quiet, flexible installation
 - Increasing need for Auxilliary electric power
 - High torque density of Electrical Mchines
 - *Up to 30 Nm/kg*
 - *ICE: <2 Nm/kg*

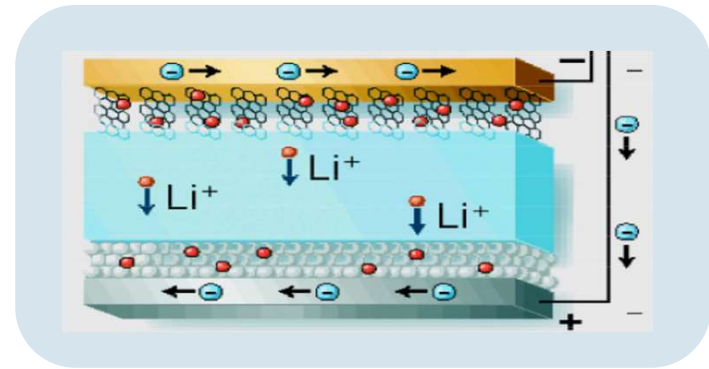
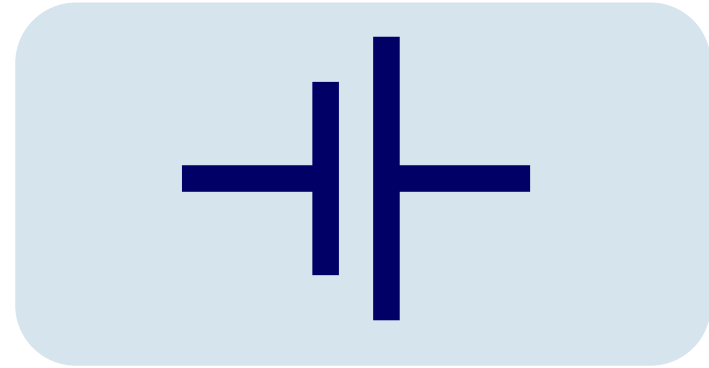
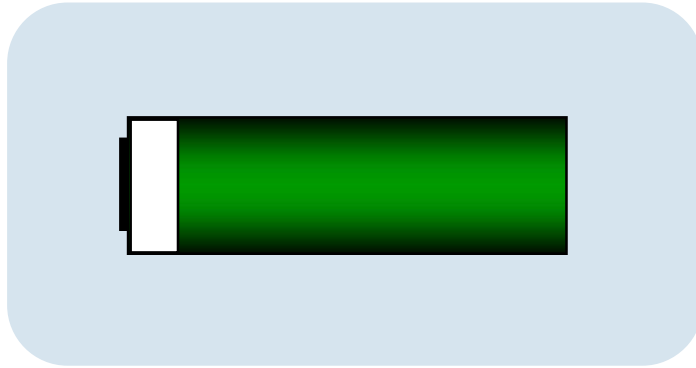
Secondary energy storage – Hybridisation



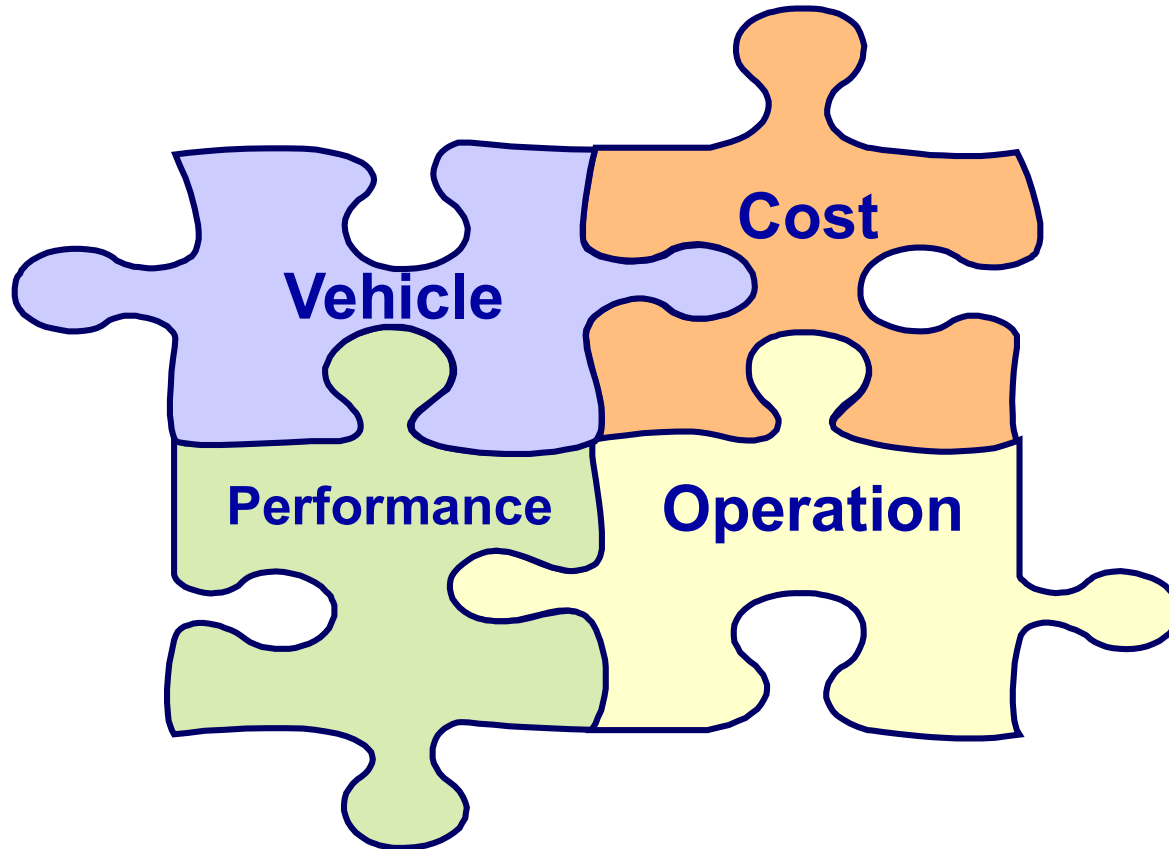


Energy Storage Systems

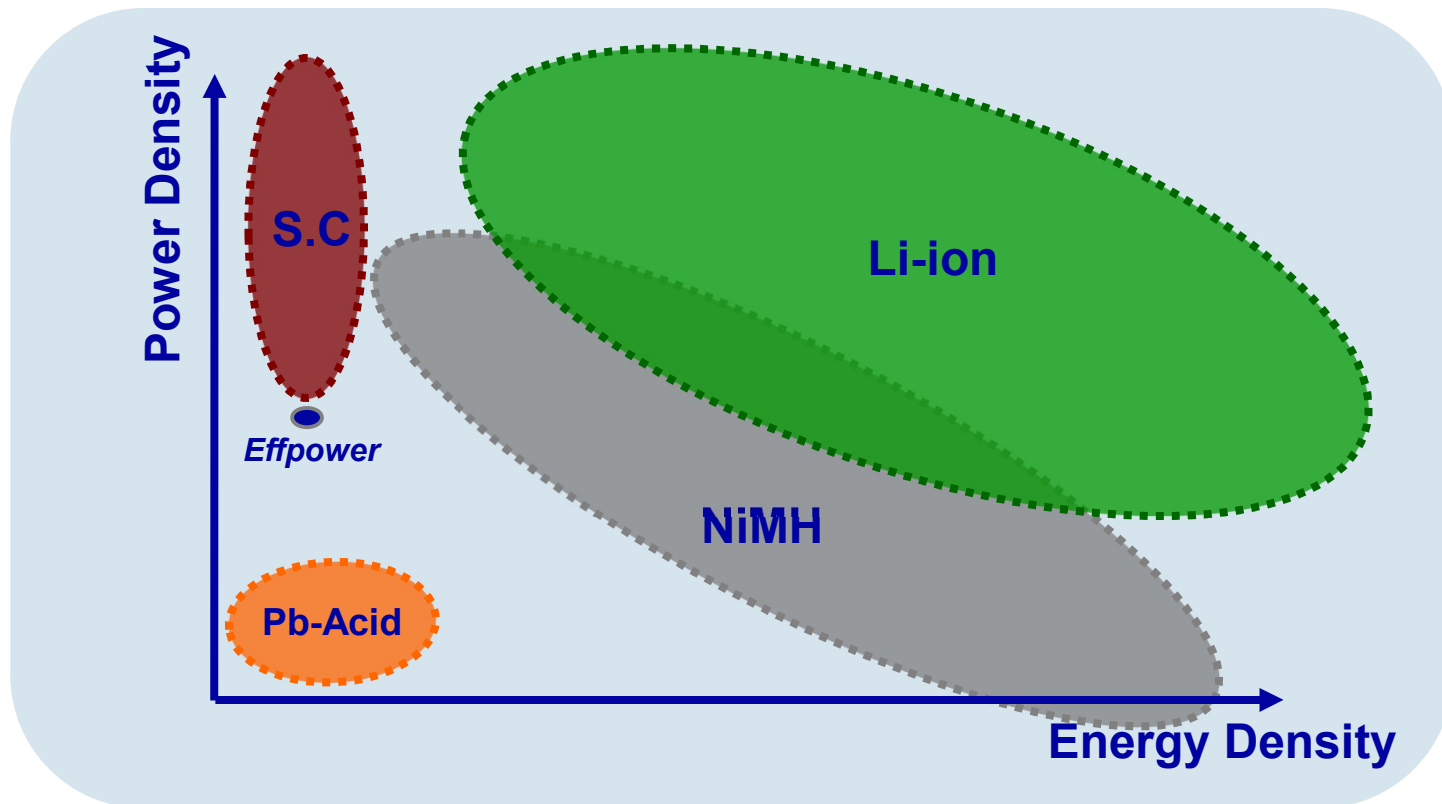
How does a battery look?



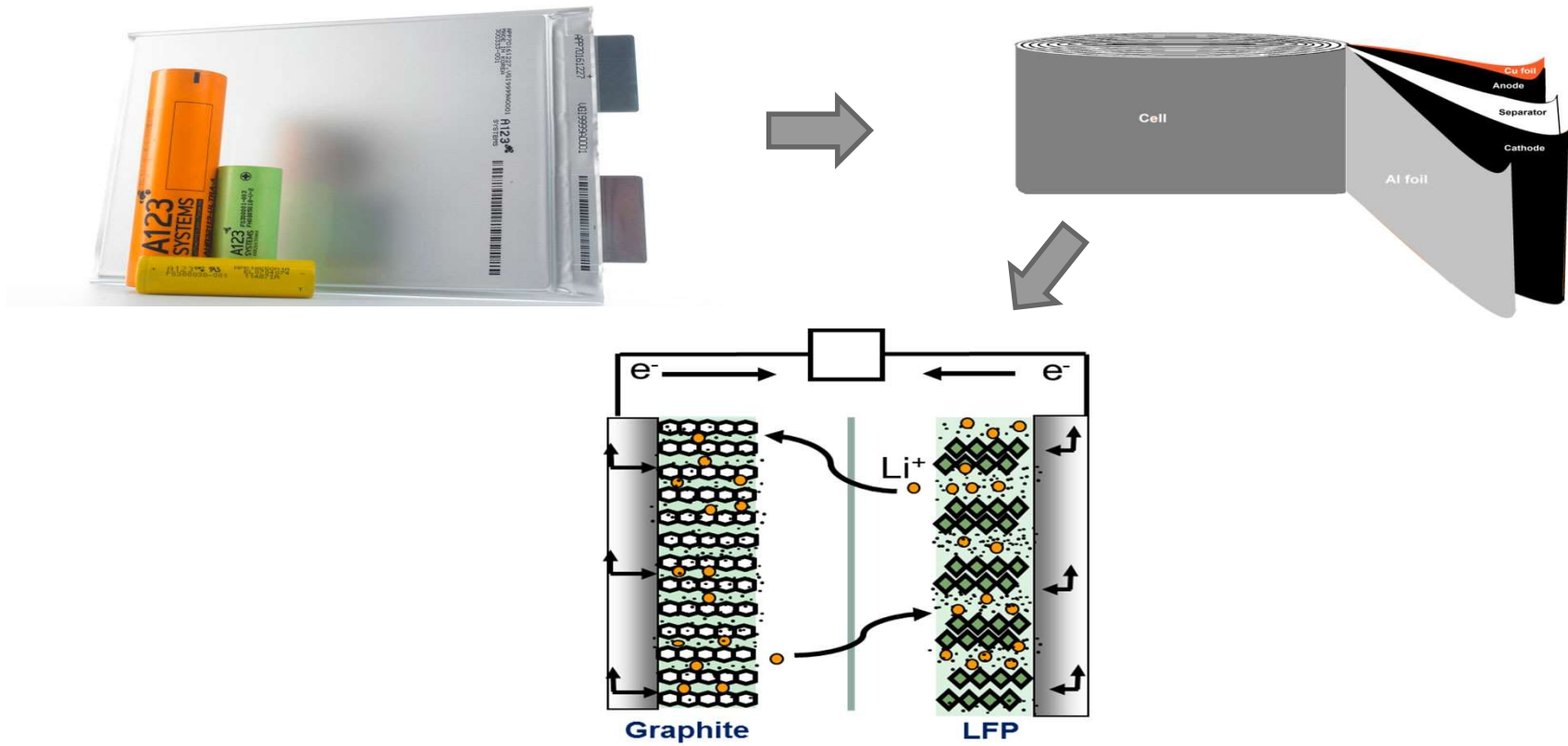
Design criteria



Performance



What is a battery cell?



Different cell formats



Prismatic

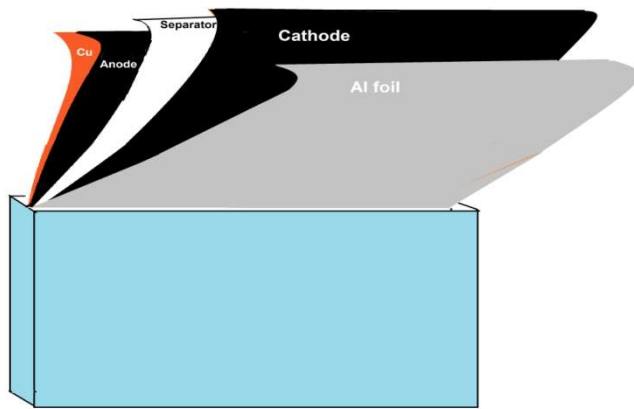


Cylindrical



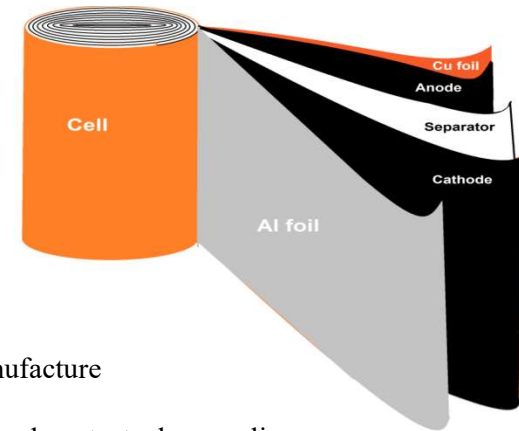
Pouch

Same principle inside



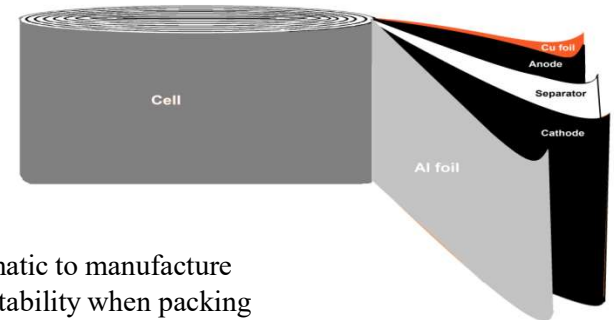
Prismatic

Simpler building block for module
More mechanically stable than pouch
(still needs mechanical support in the module)



Cylindrical

Easy and cheap to manufacture
More difficult to pack
More difficult to get good contact when cooling



Pouch

Cheaper than prismatic to manufacture
Need mechanical stability when packing

Short introduction to Battery Characteristics

The Battery Cell – a Chemical Reactor

Closed

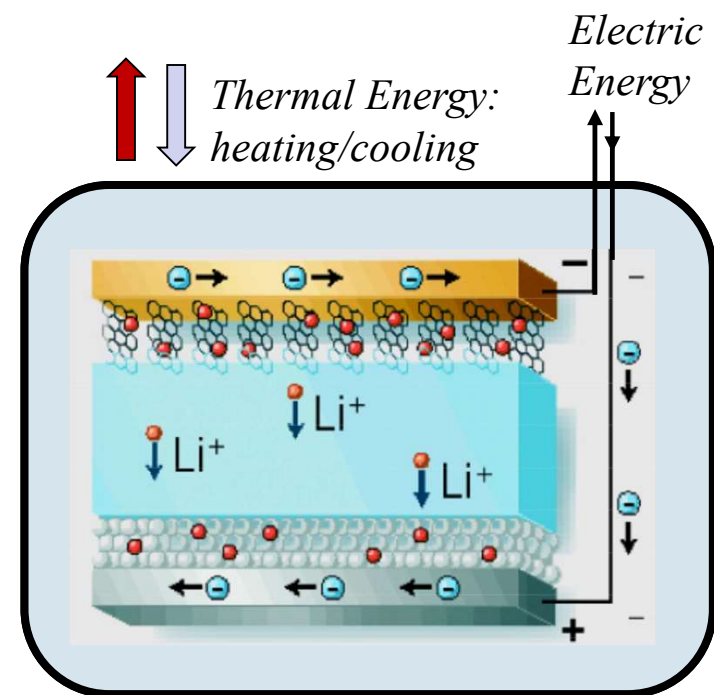
no material transfer

Controlled

no spontaneous reaction
different energy levels

Reversible

rechargeable
high efficiency



Short introduction to Battery Characteristics

SOC – state-of-charge

- Charge level related to reference capacity¹
- 0-1 or 0-100%

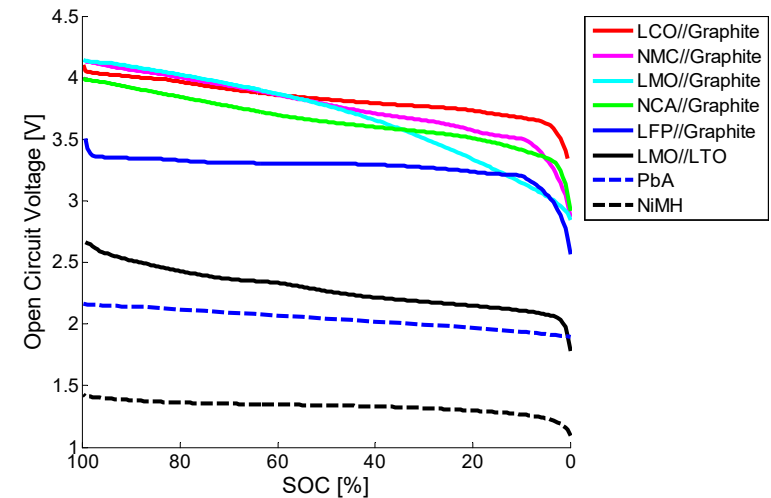
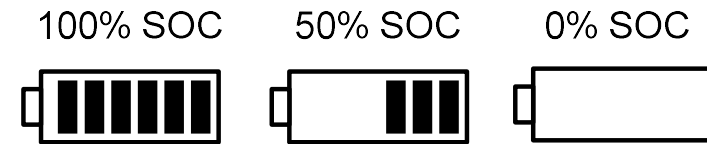
OCV – open circuit voltage

- Battery voltage at rest^o
- Measured after 15-60min rest period²
- Affected by temperature, age & hysteresis²

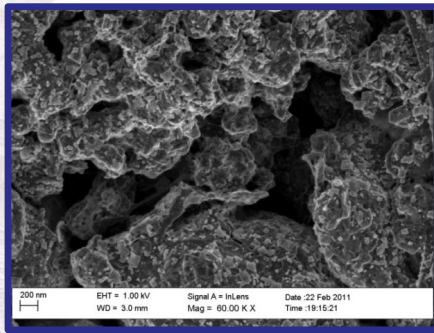
Nominal Voltage

- Electrochemical voltage \approx average OCV

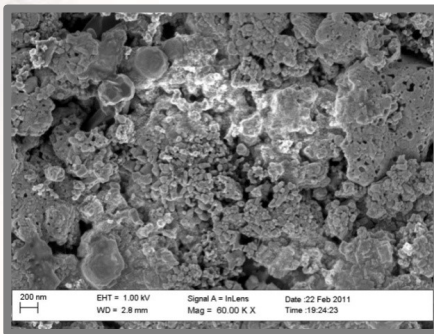
1. Often 1C-rate discharge at +23°C
2. Applicable for NiMH & LFP. See separate slide.



Physical Background & Model Design

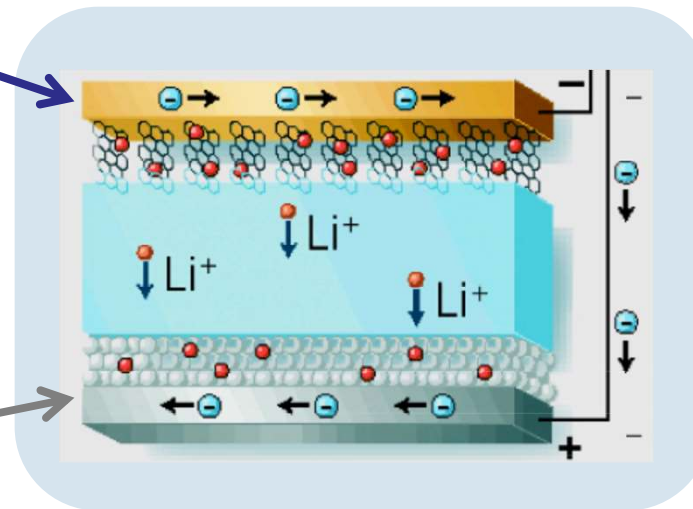


LiC_6 : SEM 60k

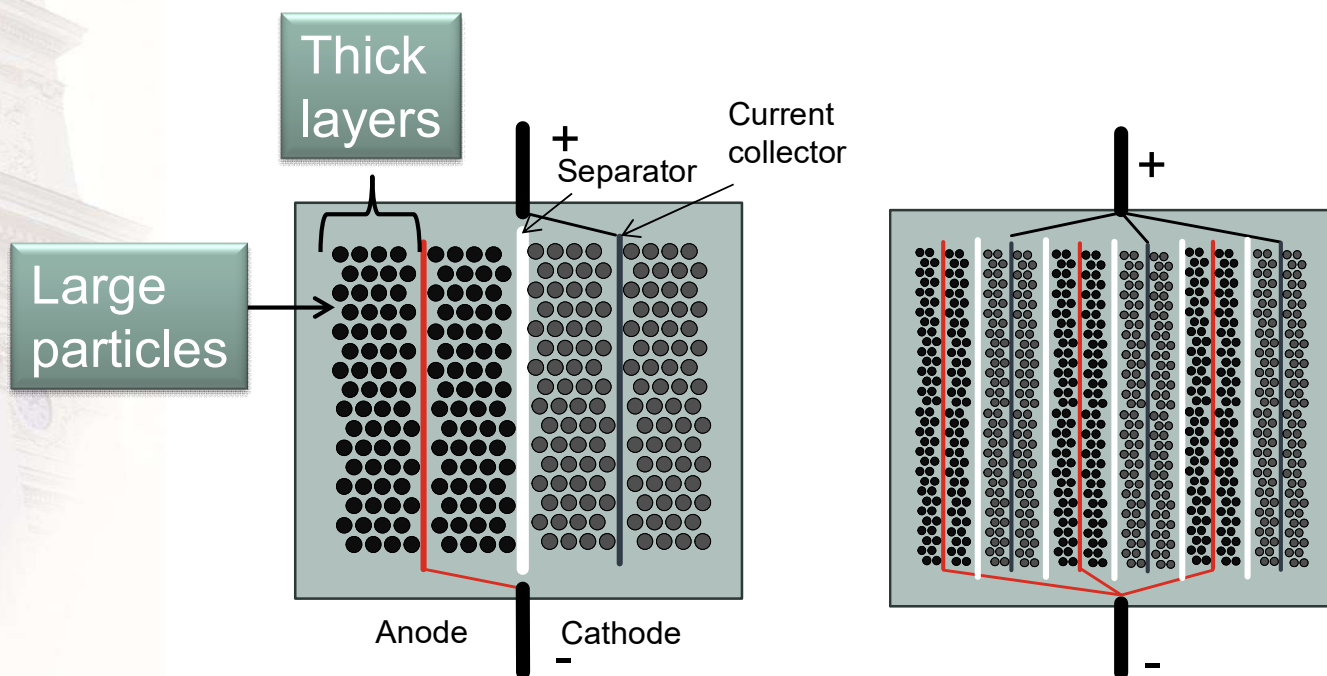


LiFePO_4 : SEM 60k

Highly porous structure to maximise active area



Power- vs. Energy- optimised



Battery Cell Properties

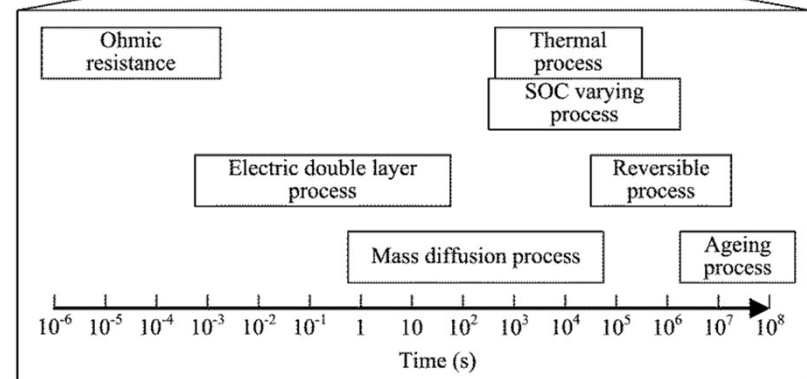
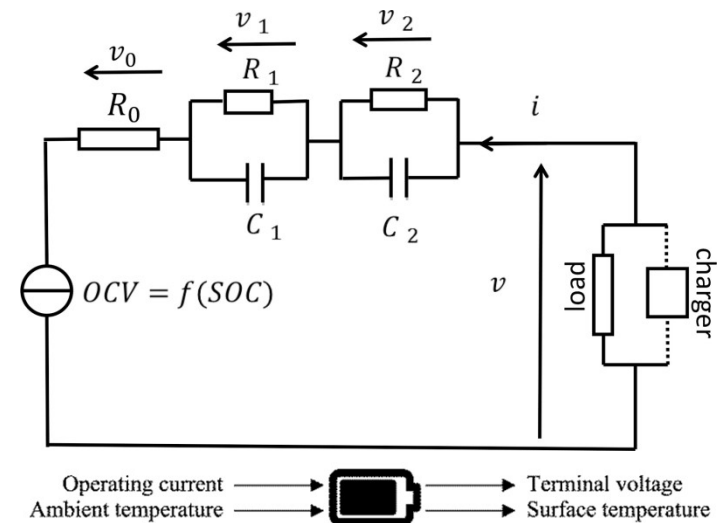
- All parameters are non-linear
- Strong dependence on SOC, temperature, current rate & direction, and history.
- OSV = Open Circuit Voltage
- R_0 , R_1 , C_1 describe the battery resistance, charger transfer and “double layer effect”¹
- R_2 , C_2 describe the diffusion² effect

1) *The first layer, the surface charge (either positive or negative), consists of ions adsorbed onto the object due to chemical interactions. The second layer is composed of ions attracted to the surface charge via the Coulomb force, electrically screening the first layer.*

[https://en.wikipedia.org/wiki/Double_layer_\(surface_science\)](https://en.wikipedia.org/wiki/Double_layer_(surface_science))

2) *Movement of Lithium ions*

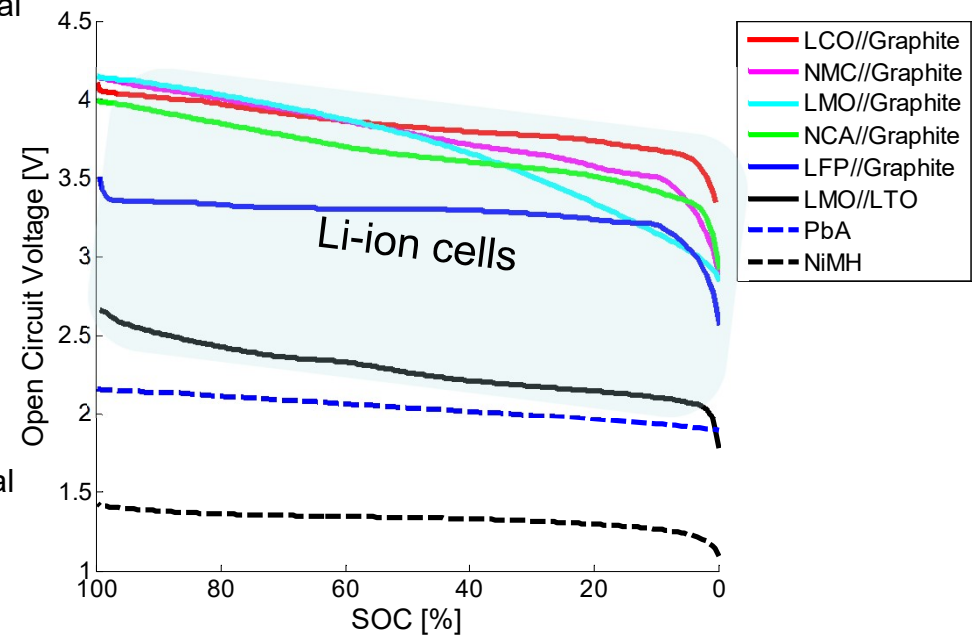
<https://www.sciencedirect.com/science/article/pii/S0360544217317127#fig1>



<http://www.mdpi.com/1996-1073/9/6/444>

Short introduction to Battery Characteristics

- Cell capacity [Ah]
 - Defined by cell design & choice of active material
 - Typical range: 5-100Ah
- Cell voltage [V]
 - Defined by choice of active material
 - Typical range 1-4.5V
- Cell energy [Wh]
 - Energy = Capacity • Voltage
 - Typical range: 5-400Wh
- Cell power [W]
 - Power = Voltage • Current
 - Limited by cell design & choice of active material
 - Typical range: 5-5000W



Short introduction to Battery Characteristics

SOH – state-of-health

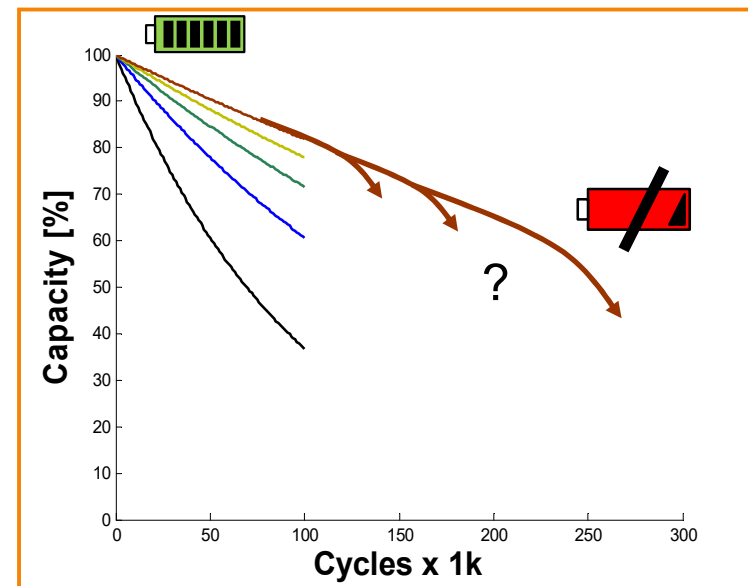
- Actual capacity related to reference capacity
- 0-1 or 0-100%, sometimes 80-100%
- No standard definition

BOL – beginning of life

- Fresh / unused battery

EOL – end of life

- Aged battery with insufficient performance
- Can be defined by capacity, energy, power etc.
- No standard definition

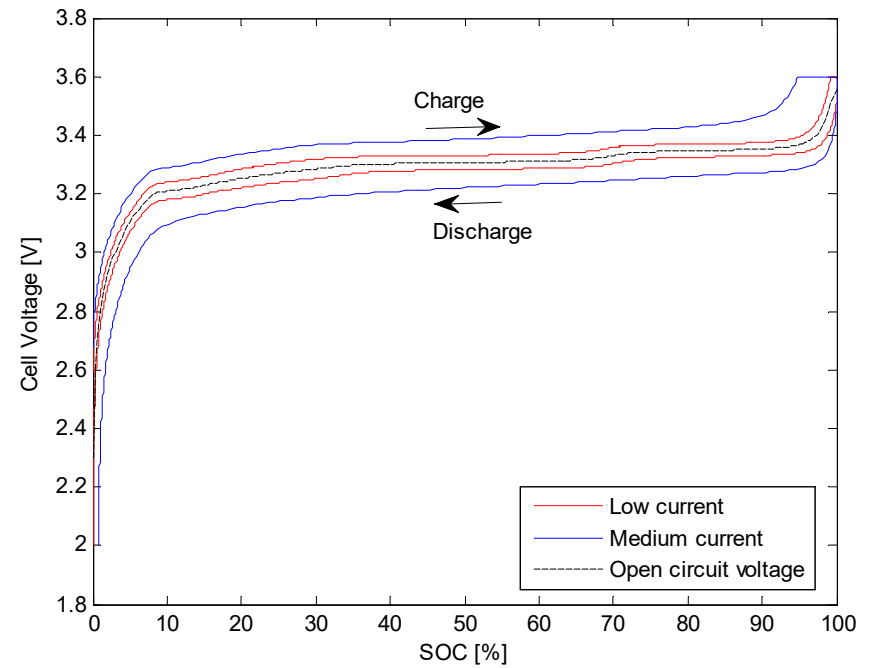


Short introduction to Battery Characteristics

Cell Voltage = $f(\dots)$

- Material Selection
- State of Charge (SOC)
- Current / power
- Temperature
- Age

→ *No equilibrium / steady-state*

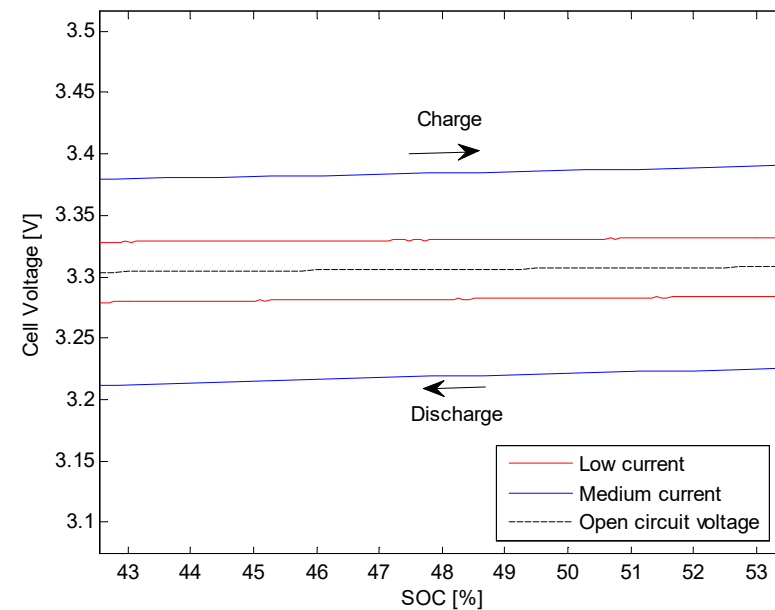


Short introduction to Battery Characteristics

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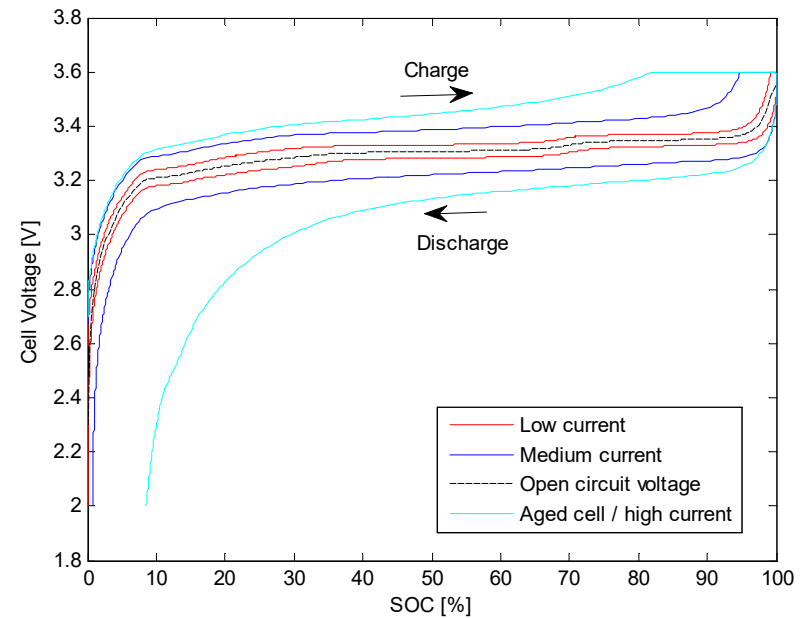


Short introduction to Battery Characteristics

Cell Voltage = $f(\dots)$

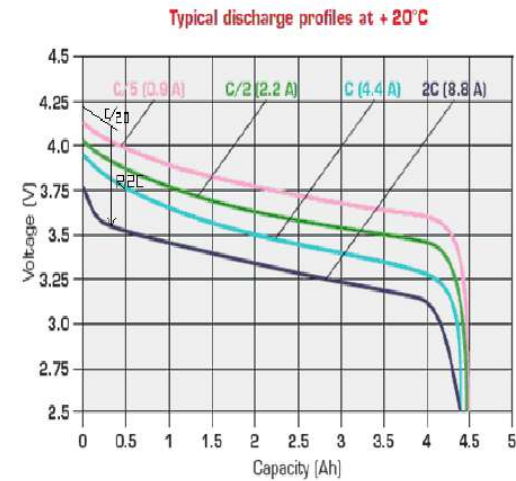
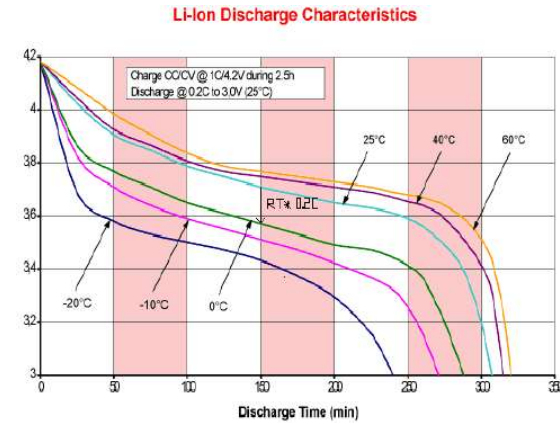
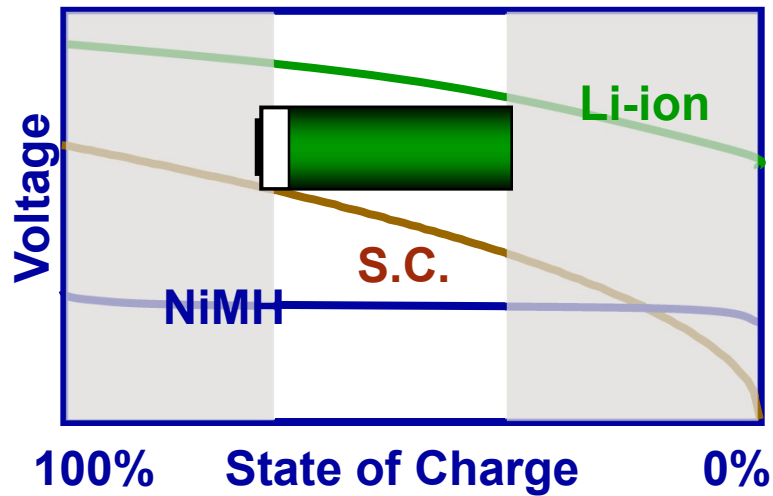
- Material Selection
- State of Charge (SOC)
- Current / power
- Temperature
- Age

→ *No equilibrium / steady-state*



State of Charge

- C-rate = $P[\text{kW}] / W[\text{kWh}]$
 - C=1: 1 h full charge
 - C=2: 30 min full charge

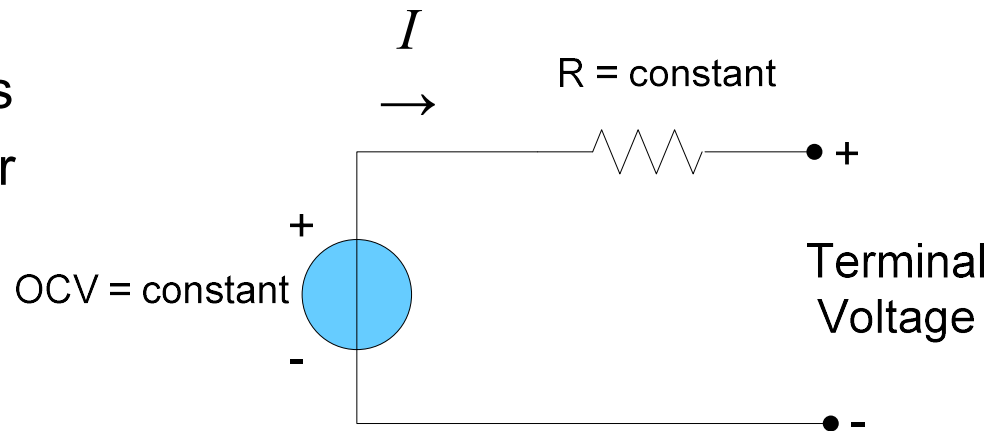


https://www.researchgate.net/figure/Discharge-curves-for-different-technologies-and-different-batteries_fig1_224395476

Applied Modelling: Performance & Accuracy

Simple Battery model

- Constant parameters
- Very fast, suitable for optimisation



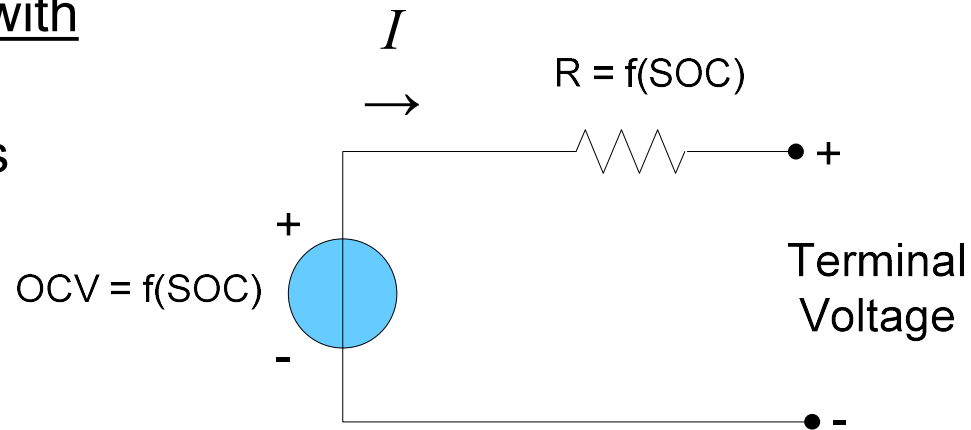
- Very limited accuracy

$$P = U \cdot I = U \left(\frac{U - OCV}{R} \right) = \frac{1}{R} (U^2 - U \cdot OCV)$$

Applied Modelling: Performance & Accuracy

Simple Battery model with SOC

- Simple look-up tables
- Still fast, suitable for optimisation



$$OCV = f(\text{SOC})$$

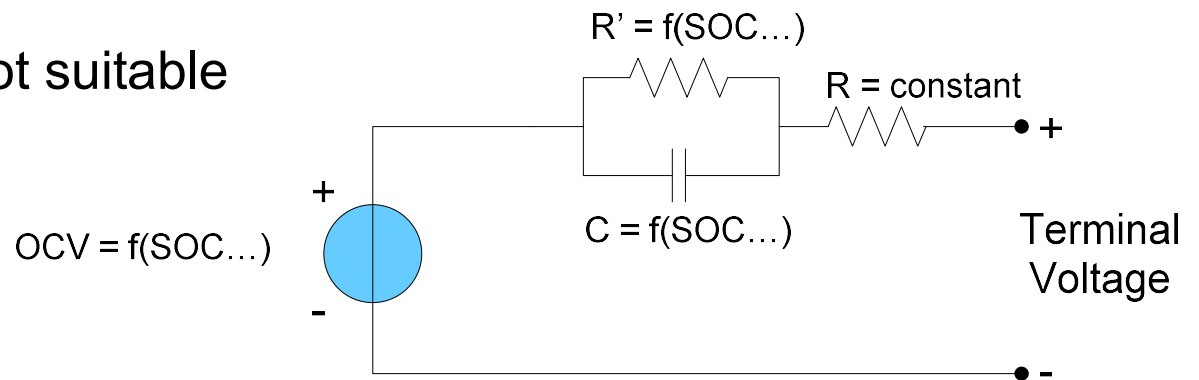
$$U = OCV(\text{SOC}) - R(\text{SOC}) \cdot I$$

- Still limited accuracy

Applied Modelling: Performance & Accuracy

Enhanced battery model with SOC & t

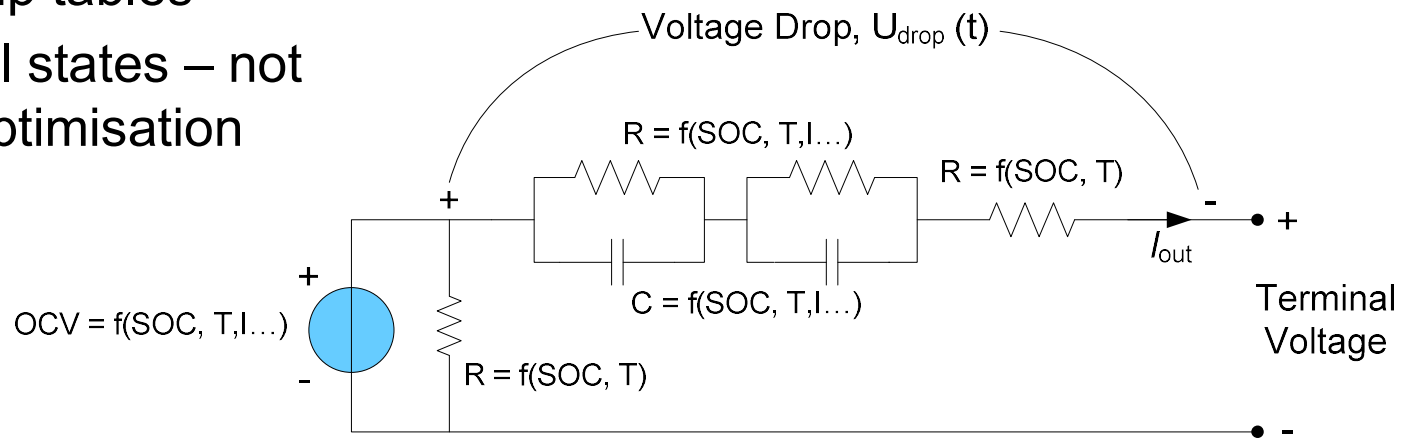
- Simple look-up tables
- Two internal states – not suitable for optimisation



Applied Modelling: Performance & Accuracy

Advanced battery model with SOC, T & t

- Simple look-up tables
- Three internal states – not suitable for optimisation

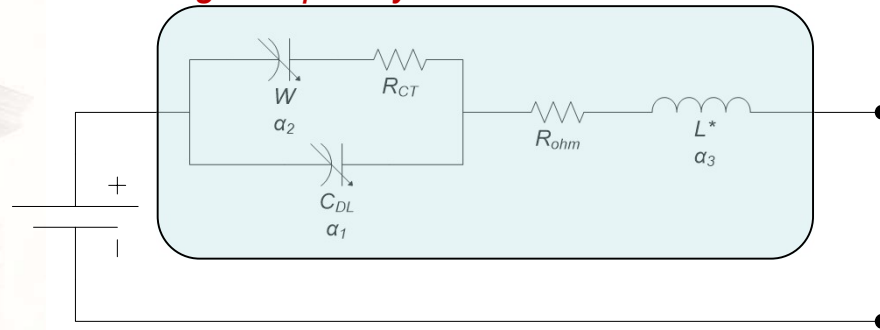


GSP Model

Applied Modelling: Performance & Accuracy

Separate high-frequency models are needed for specific cases

batteries act as inductors at high frequency and as capacitors at low



Models available from Volvo GTT-ATR for specific cells upon request

$$Z = (j\omega)^{\alpha_3} L^* + R_{Ohm} + \frac{1}{\frac{1}{R_{ct} + \frac{1}{C_w(j\omega)^{\alpha_2}}} + \frac{1}{C_{dl}(j\omega)^{\alpha_1}}} \quad [Ohm]$$

Battery simulation model : I

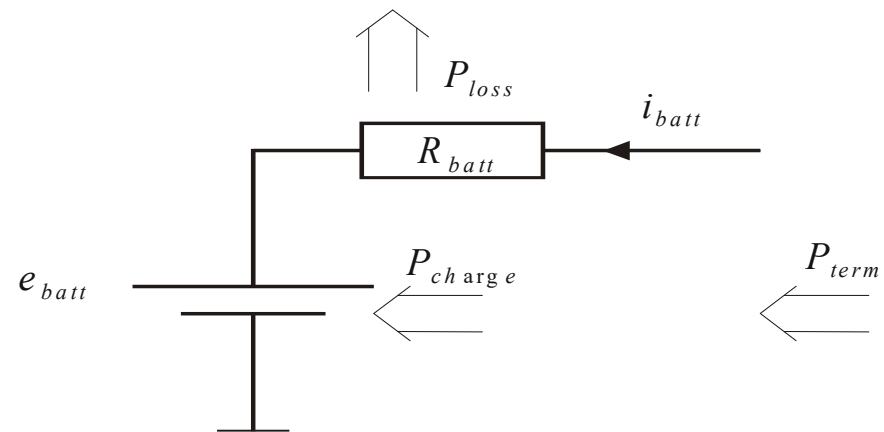
$$P_{term} = (e_{batt} + R_{batt} \cdot i_{batt}) \cdot i_{batt} = e_{batt} \cdot i_{batt} + R_{batt} \cdot i_{batt}^2$$

$$i_{batt} = -\frac{e_{batt}}{2 \cdot R_{batt}} \pm \sqrt{\left(\frac{e_{batt}}{2 \cdot R_{batt}}\right)^2 + \frac{P_{term}}{R_{batt}}}$$

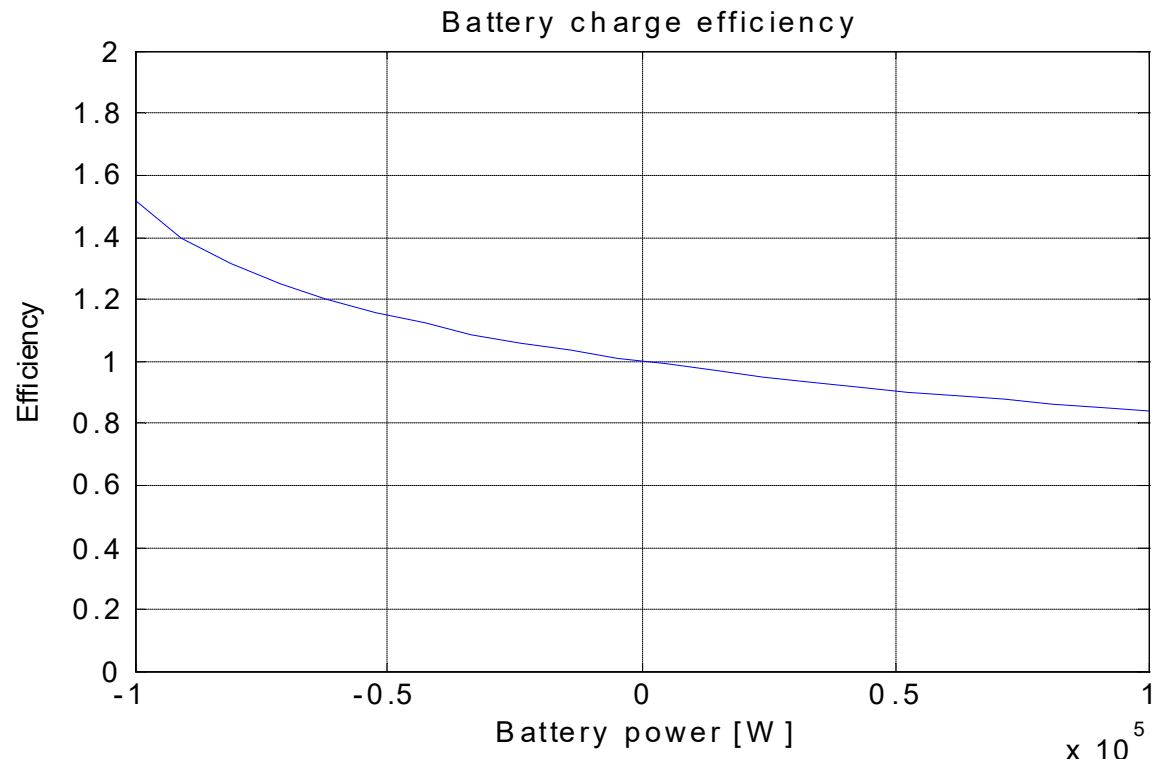
$$P_{loss} = R_{batt} \cdot i_{batt}^2$$

$$P_{charge} = P_{term} - P_{loss}$$

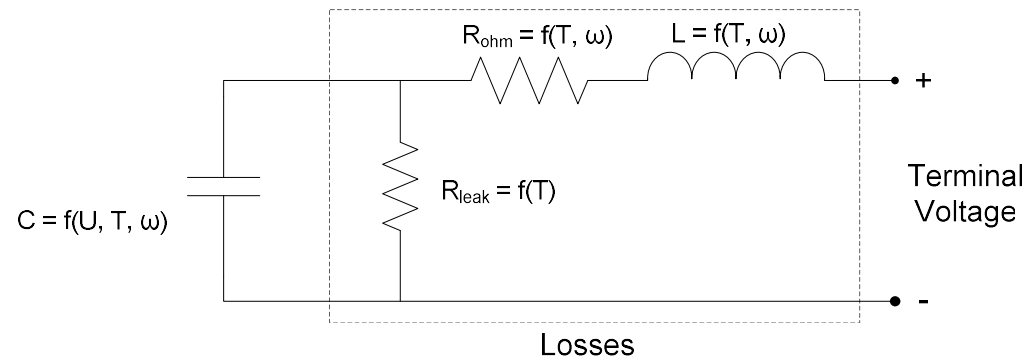
$$\eta_{batt} = \frac{P_{charge}}{P_{term}}$$



Battery Simulation model : II



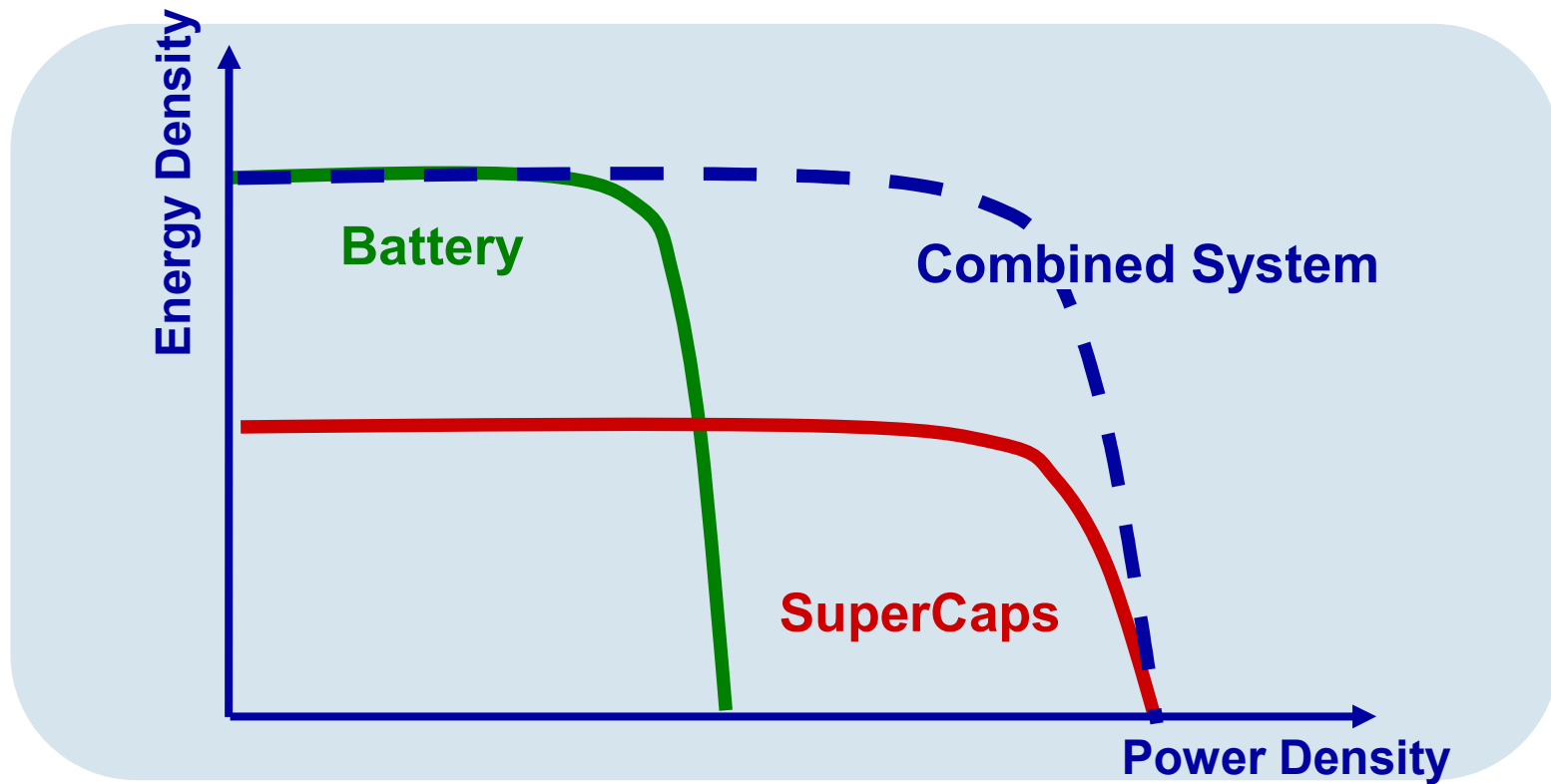
Super Capacitor Cell Properties



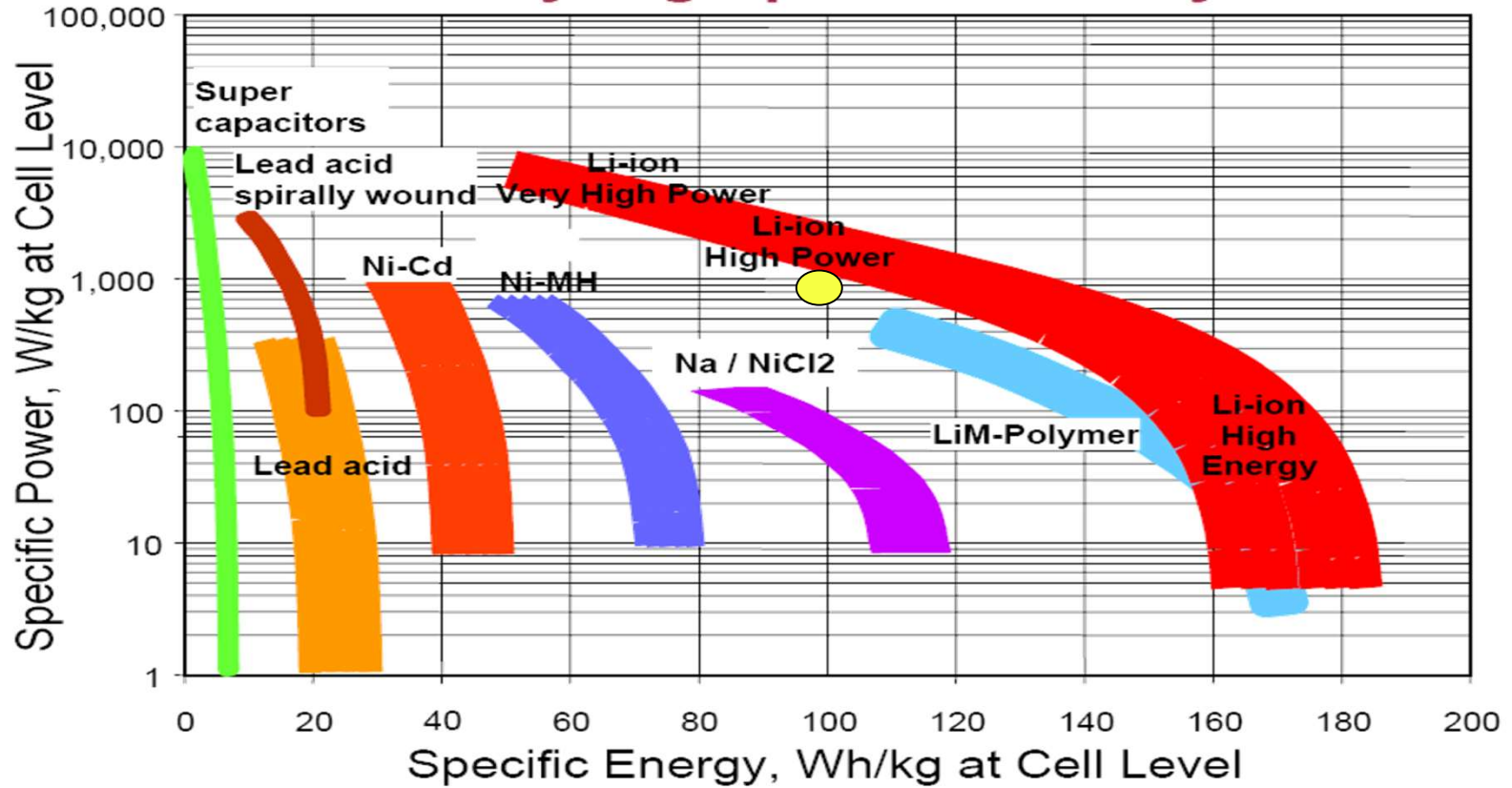
- Fewer non-linear elements
- Strong dependence on SOC & temperature only

"Semi" Steady-State Operation!

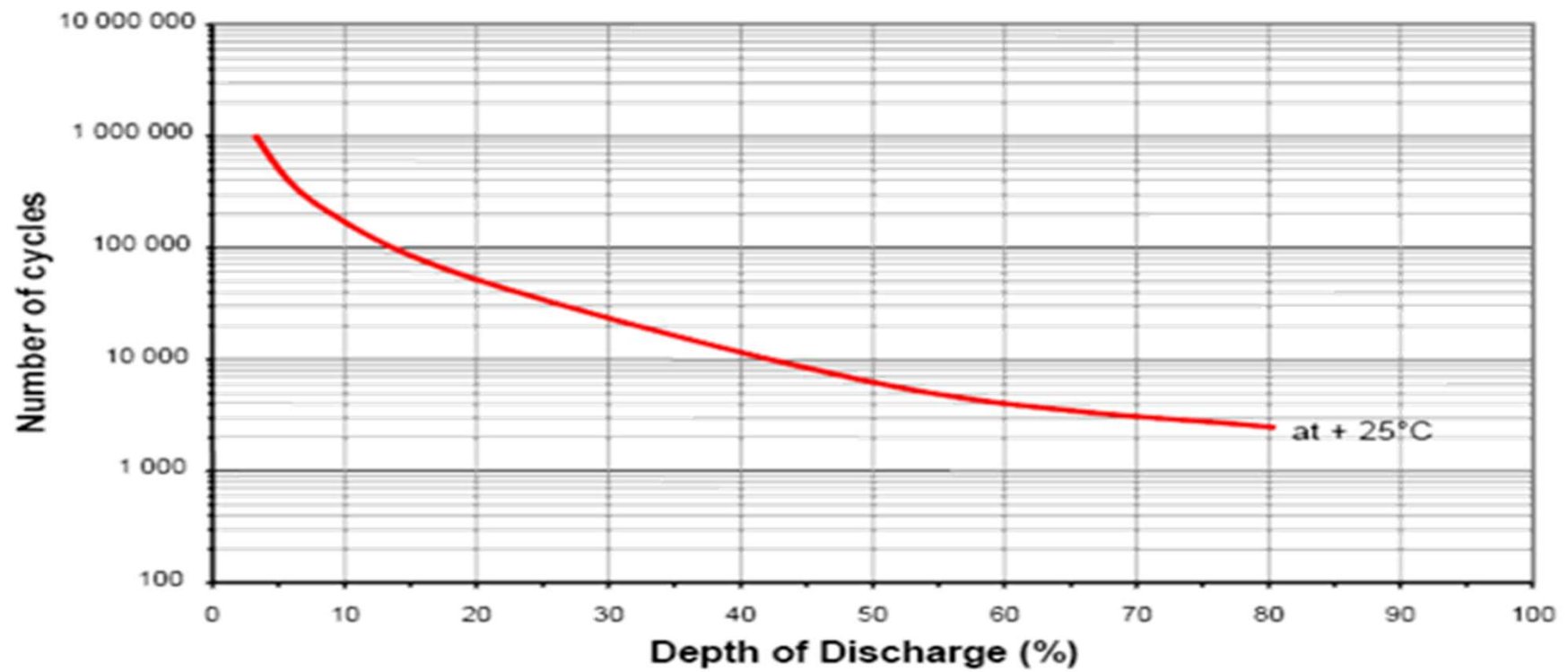
Combined Energy Storage System



SAFT data

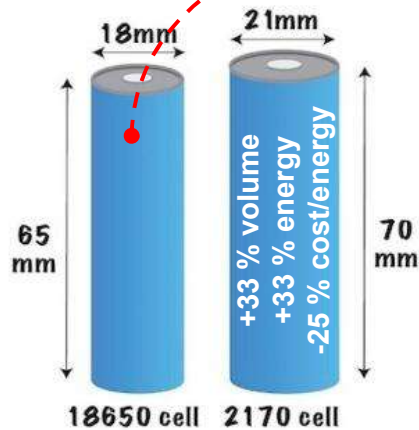


SAFT Lithium teknologi

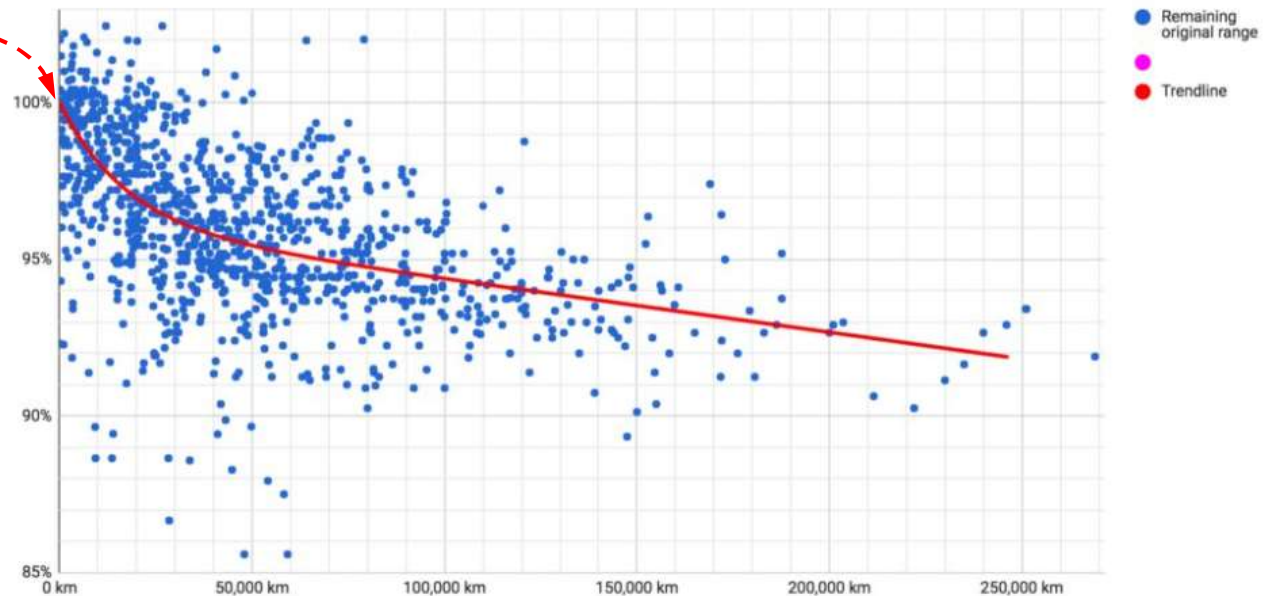


Tesla Model S&X, Milage vs remaining battery capacity

- Based on the 18650 cell.



- Model 3 use the 2170 cell



Tesla Model S and Model X battery degradation chart [Credit: Matteo via Maarten Steinbuch]

<https://www.teslarati.com/tesla-battery-life-80-percent-capacity-840km-1-million-km/>

Battery Lifetime : II

- Calculate the total converted energy as:

$$W_{conv} = W_{batt} * DoD / 100 * NoC$$

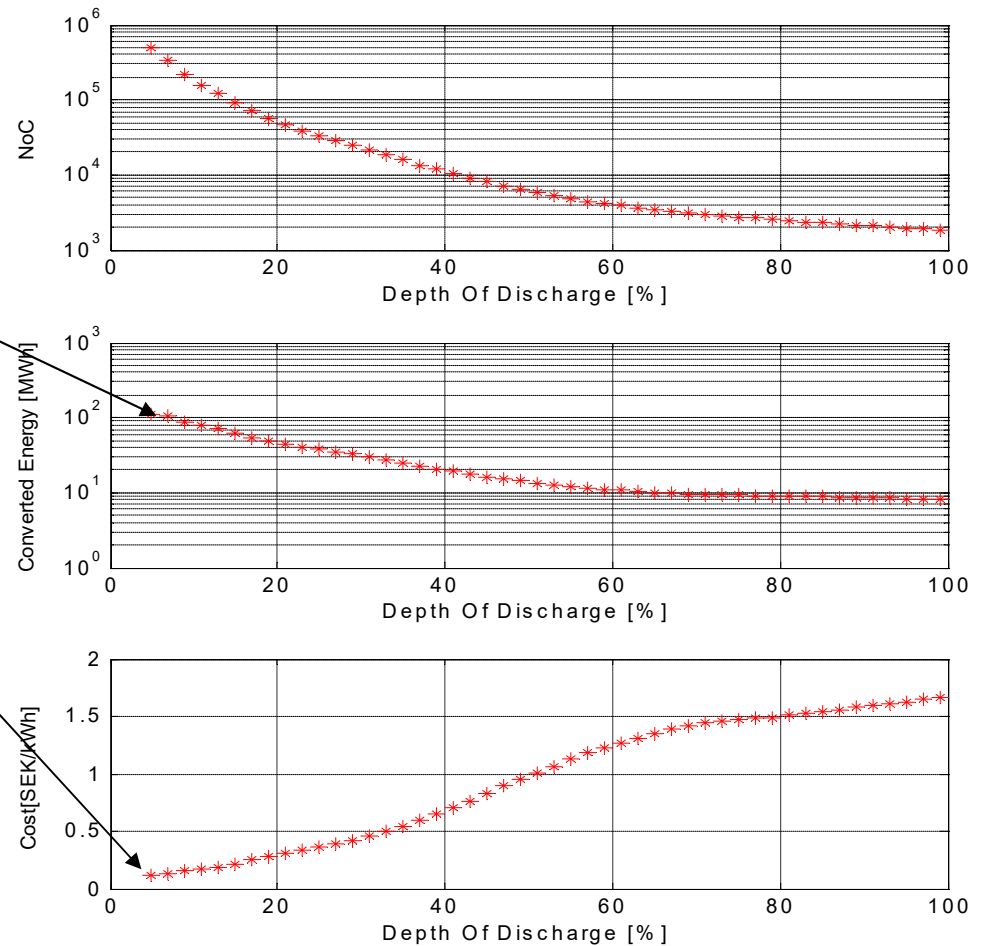
- And the battery cost per kWh converted energy as:

$$SEK_{perkWh} = CostPerkWh * W_{batt} / W_{conv}$$

$$CostPerkWh = 3000 [SEK]$$

Battery Lifetime : III

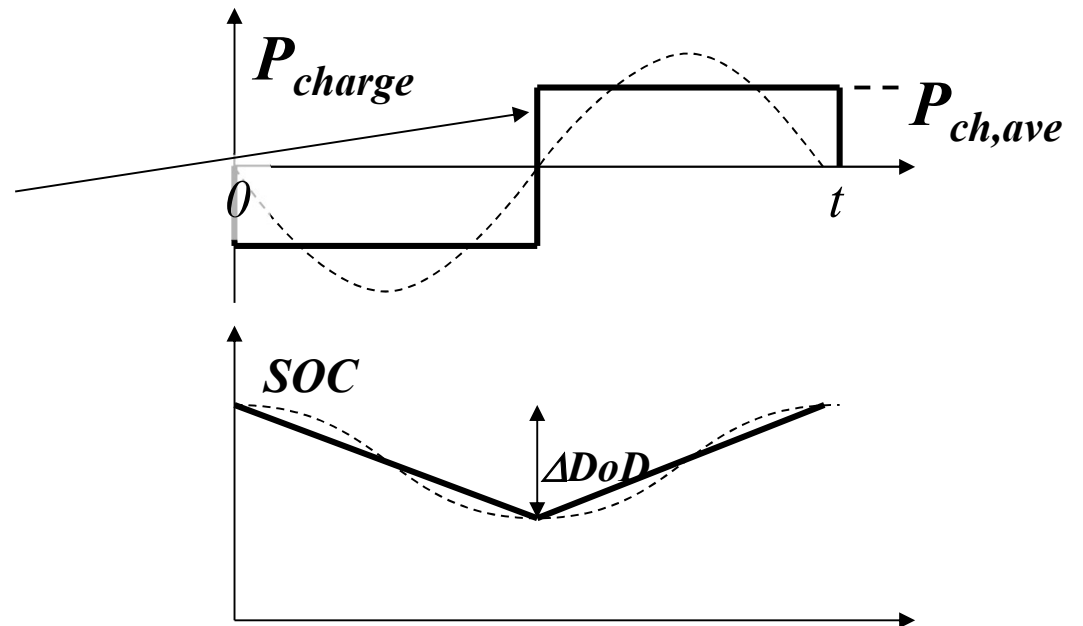
- Note! – The total converted energy falls fast with increased DoD !!!
- The cost per converted kWh increases rapidly with increased DoD



Battery Lifetime : IV

- Assume that DoD is driven by a constant charge Power
- then... :

$$\Delta DoD = \frac{P_{ch,ave} \cdot \frac{t}{2}}{W_{batt}} \cdot 100$$



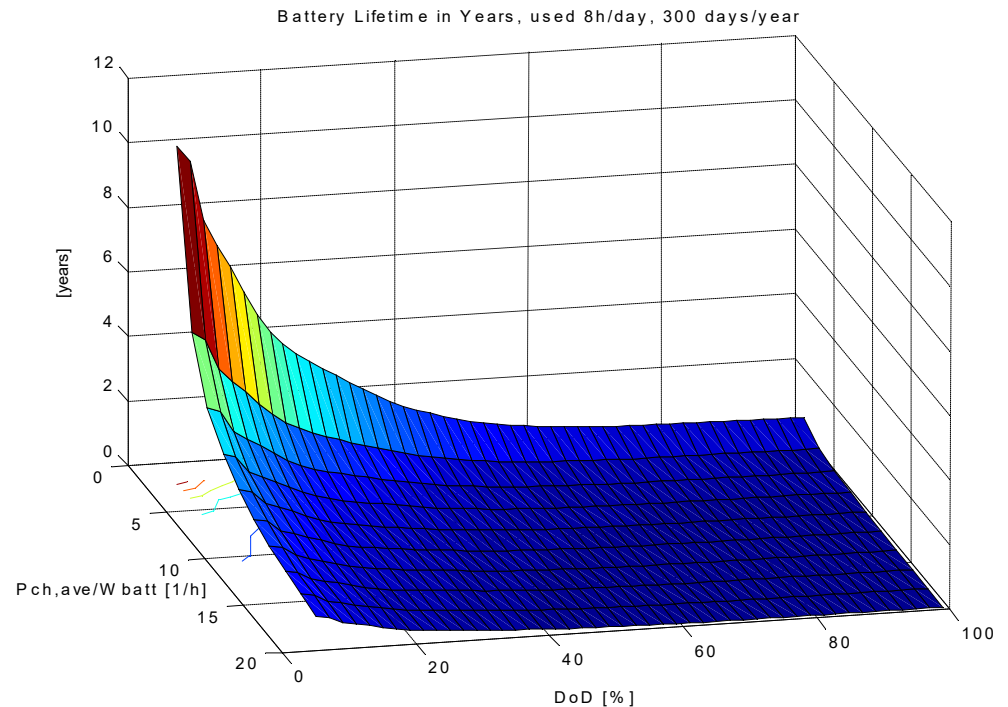
Battery Lifetime : V

- The battery lifetime can be calculated as:

$$\Delta DoD = \frac{P_{ch,ave} \cdot \frac{t}{2}}{W_{batt}} \cdot 100$$

$$\begin{aligned} \text{Lifetime} &= \frac{DoD \cdot W_{batt} \cdot 2}{P_{ch,ave} \cdot 100} \cdot NoC = \\ &= \frac{DoD \cdot 2}{\frac{P_{ch,ave}}{W_{batt}} \cdot 100} \cdot NoC \end{aligned}$$

- Now, plot the battery lifetime as a function of DoD and $(P_{ch,ave}/W_{batt})$
 - Assume 8h/day, 300 days/year



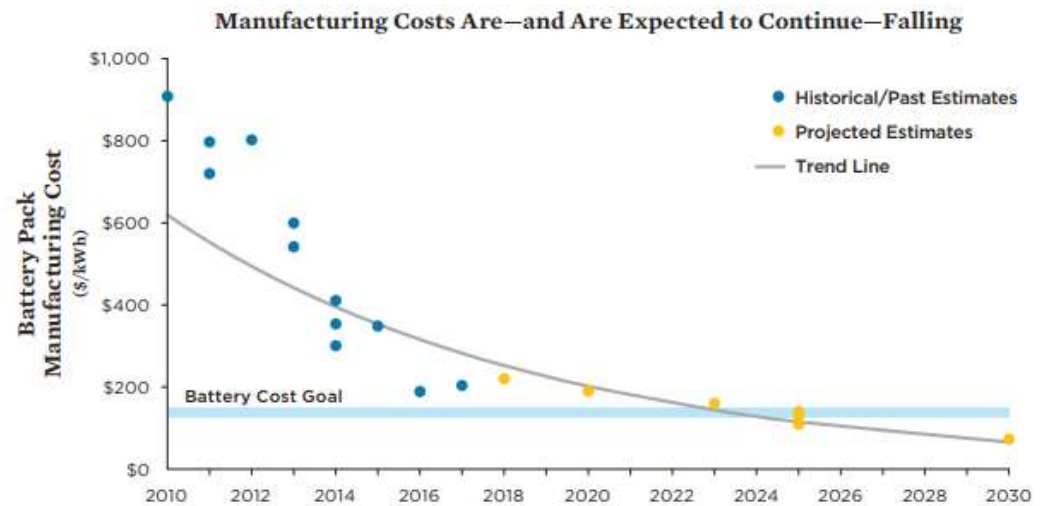


Cost and 2nd life

Battery Cost development

- The Cost of EV Traction Batteries is falling fast

EV Battery Pack Manufacturing Costs Predicted to Fall over Time

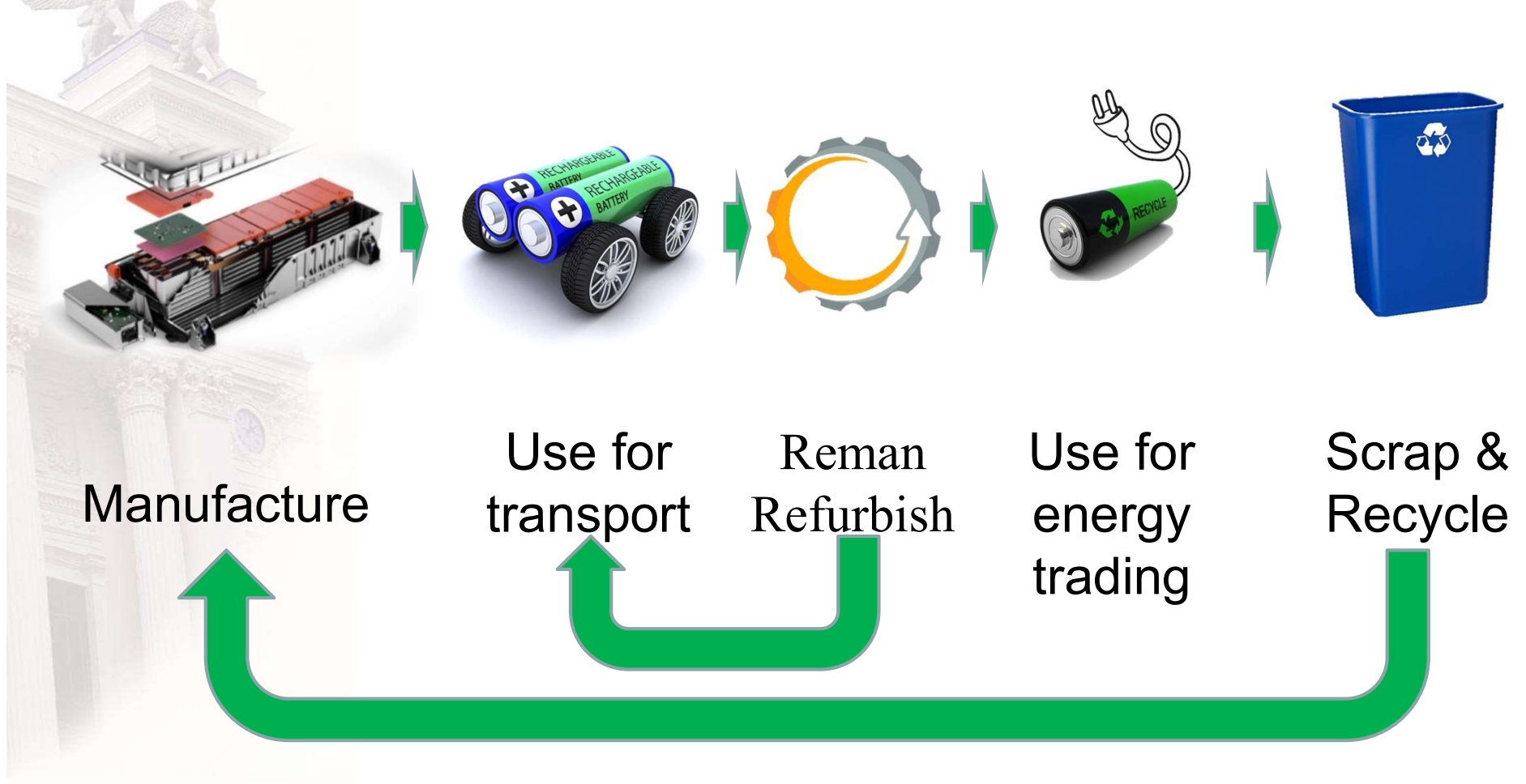


If battery costs continue to decline as EV production increases, within several years they will reach the \$125–\$150 target that makes EVs competitive with conventional gasoline vehicles.

Note: Battery cost estimates include both academic analysis and statements from automakers. Multiple data points in a given year represent estimates from multiple analyses. Trend line represents exponential best fit of battery cost data.

SOURCES: ARB 2017; SOULOPOULOS 2017; VOELCKER 2017; SLOWIK, PAVLENKO, AND LITSEY 2016; VOELCKER 2016; NYKVIST AND NILSSON 2015.

The traction battery life, and business, cycle



Battery 2nd Life

- EV's use the battery until 85-90 % capacity is left
- After leaving the vehicle, another 5...10 years of life remains.
- Grid Energy storage is an important application

Daimler, 10's of MW and MWh



Tesla, 560 MW and 129 MWh

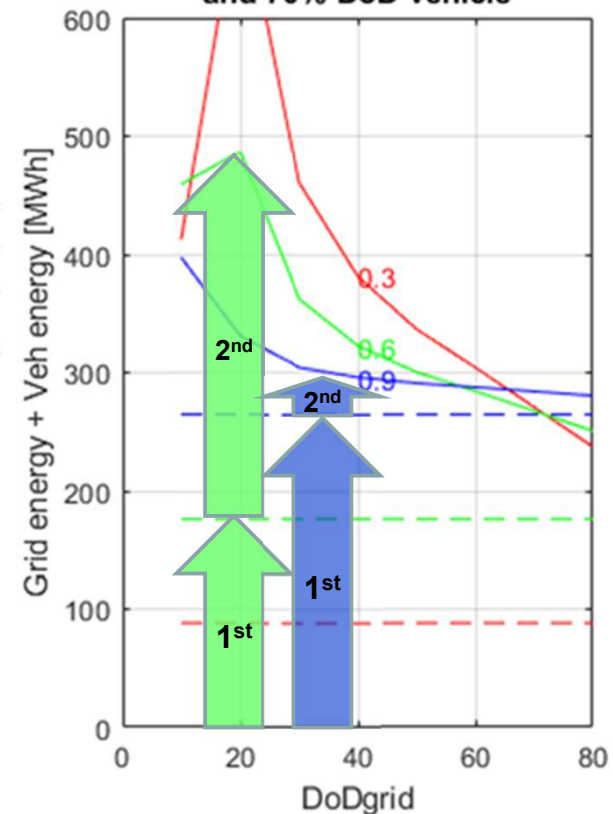


Grid Battery Energy Storage – a Swedish perspective

		"Fast charger and BIG Battery World"				
	# of vehicles in Sweden	Battery Size [kWh]	Battery lifetime in vehicle [years]	Battery 2'nd life in grid [years]	Grid battery energy [GWh]	Supply time @ 10 GW [h]
Cars	5 000 000	75	10	5	188	19
Heavy Trucks	50 000	500	4	2	13	1
Total					200	20

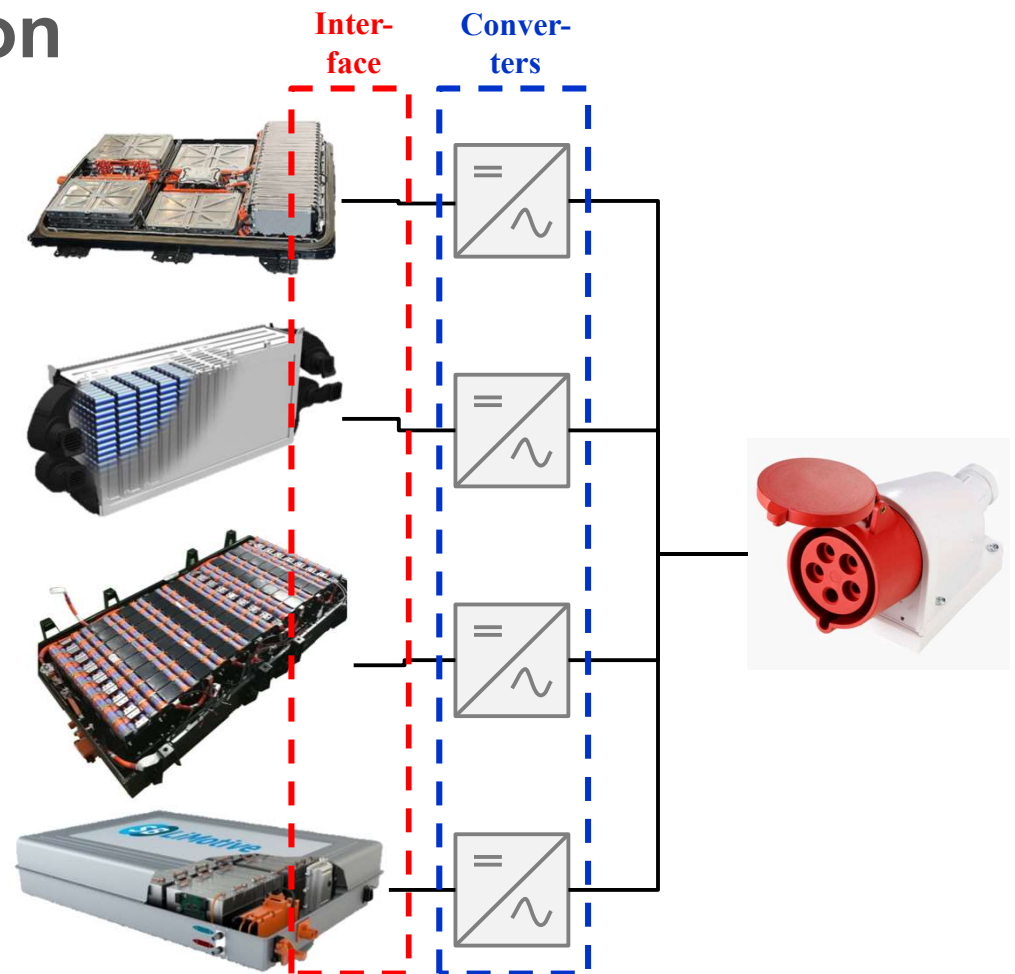
- Independent of Scenario – The **2'nd Life** batteries represent a **Significant Energy** supply for the electricity grid
- The (re)used EV battery is a **value carrier** into the electricity grid, together with renewable electricity generation

Veh+Grid energy turnover until 10 years lifetime @ 10 Grid cycles/day [MWh] and 70% DoD vehicle



Battery 2nd life application

- In a few years we will have 1000's of batteries to handle.
- These will be more or less different ...
- These have to be joined in a systematic way
- Important to have a well defined **interface** and efficient **converters**
- "Design for 2nd life"-requirements needed on ESS HW and SW!



Battery Lifetime: Conclusions

- Deep discharge of a battery costs lifetime
 - Best with less than 10 % DoD
- The battery may have a high power density BUT using it is expensive, i.e. a high power/energy ratio costs lifetime
 - Best with an average battery power in the range 3...4 x the battery energy capacity.

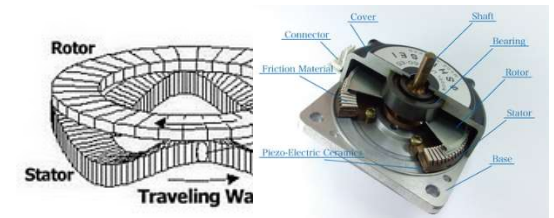


Electric Motor Drive Systems

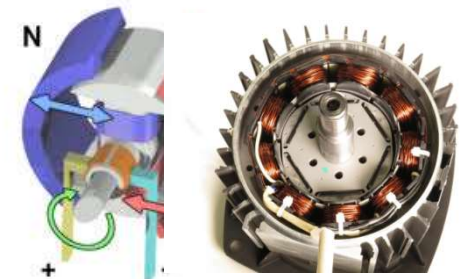
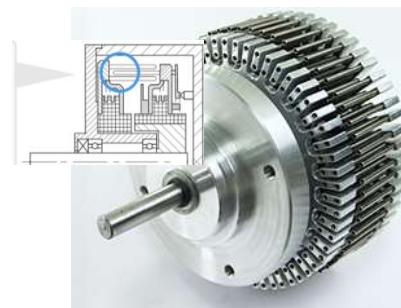


What is an Electrical Machine?

- "A device that convertes between Electric and Mechanic energy, both ways"
- Physical principles?
 - › Some kind of field ...
 - » *Acoustic field*
 - » *Piezoelectric deformation*
 - » *Electrostatic field*
 - » *Magnetic field*



<http://www.shinsei-motor.com/English/techno/index.html>



Energy density, Magnetic vs Electric

$$\approx 1 \text{ [Tesla]}$$

Magnetic Energy Density

$$\frac{B^2}{2\mu_0}$$

$$\mu_0 = 4\pi \cdot 10^{-7}$$

$$\approx 4 \cdot 10^5 \text{ [Ws / m}^3\text{]}$$

Electric Energy Density

$$\varepsilon_0 \frac{E^2}{2}$$

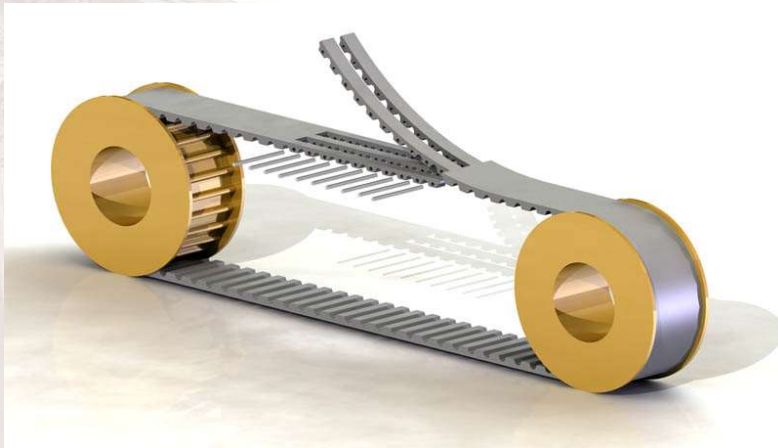
$$\approx 4 \cdot 10^1 \text{ [Ws / m}^3\text{]}$$

$$\varepsilon_0 = 8.85 \cdot 10^{-12}$$

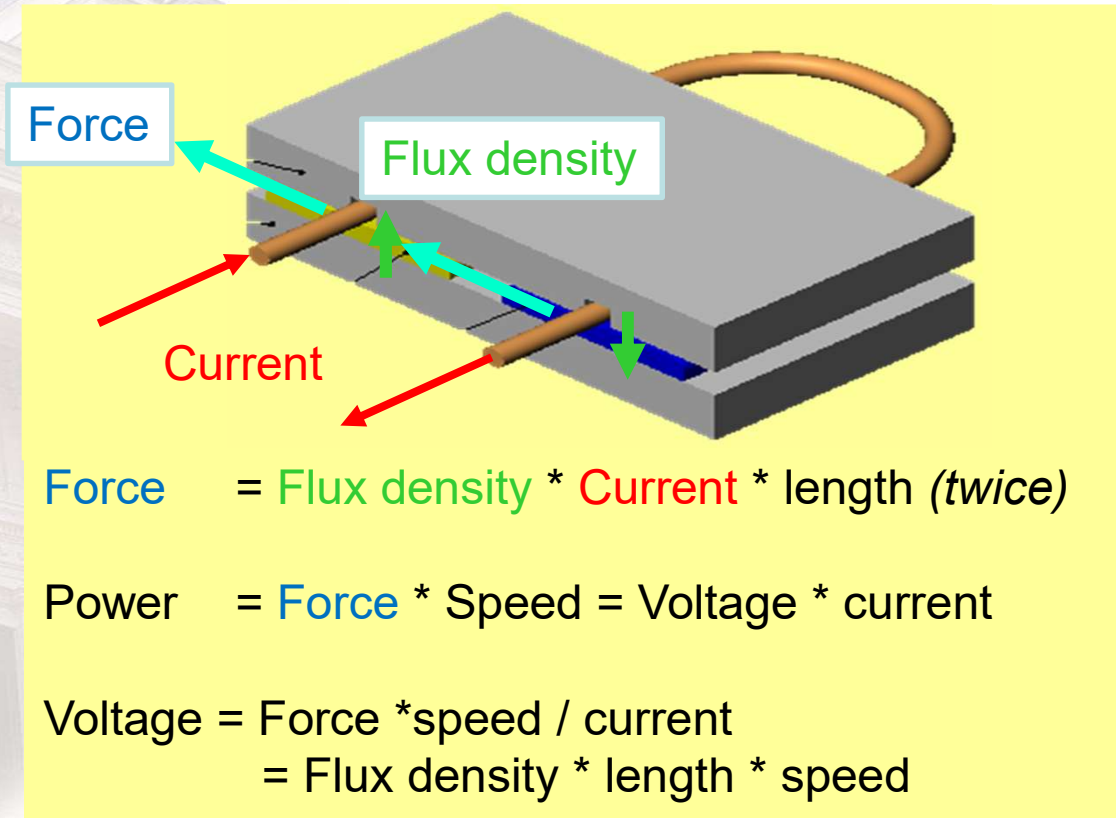
$$\approx 3 \text{ [MV/m]}$$

Linear Motion

- In many applications the “most wanted”
- Often translated from a rotation

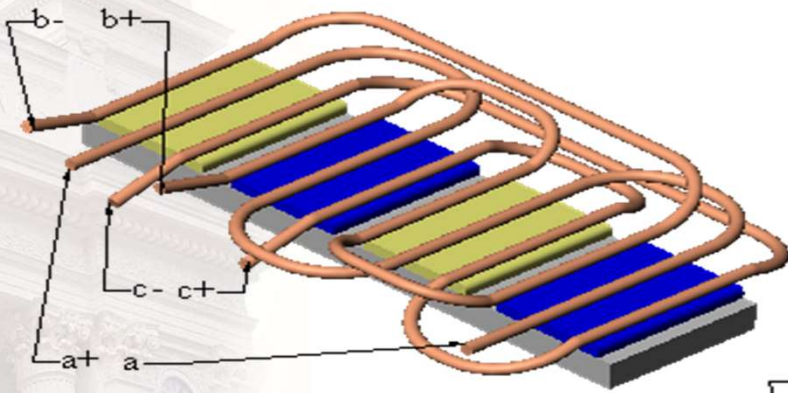


Generic Force

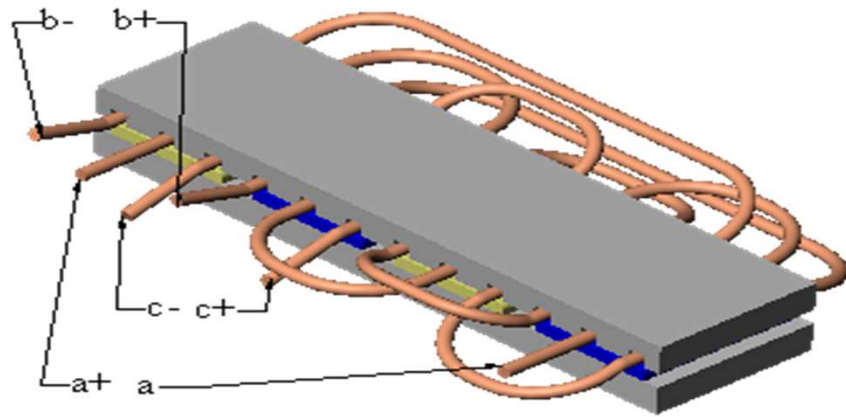


***Lorentz force
= current in magnetic field***

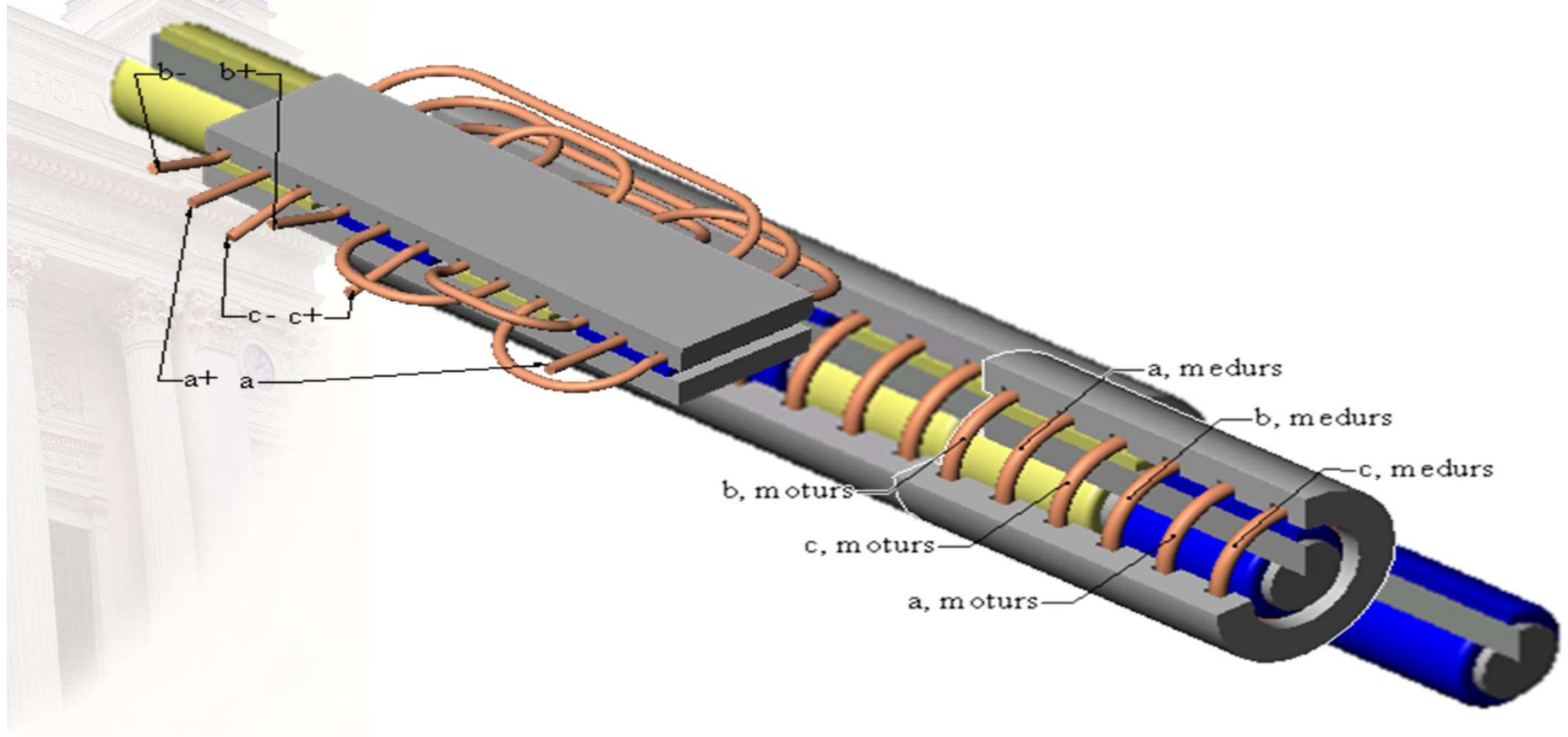
Multi Phase, otherwise it stops



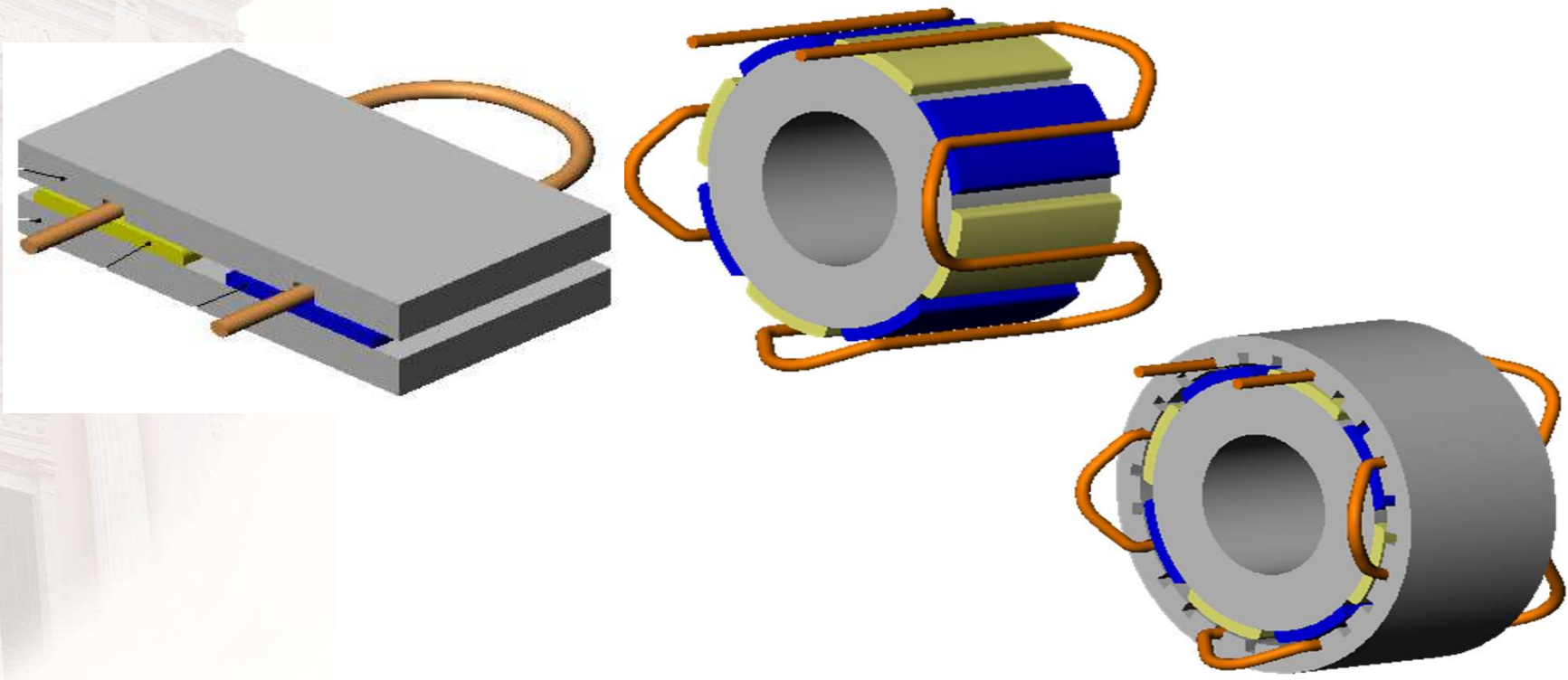
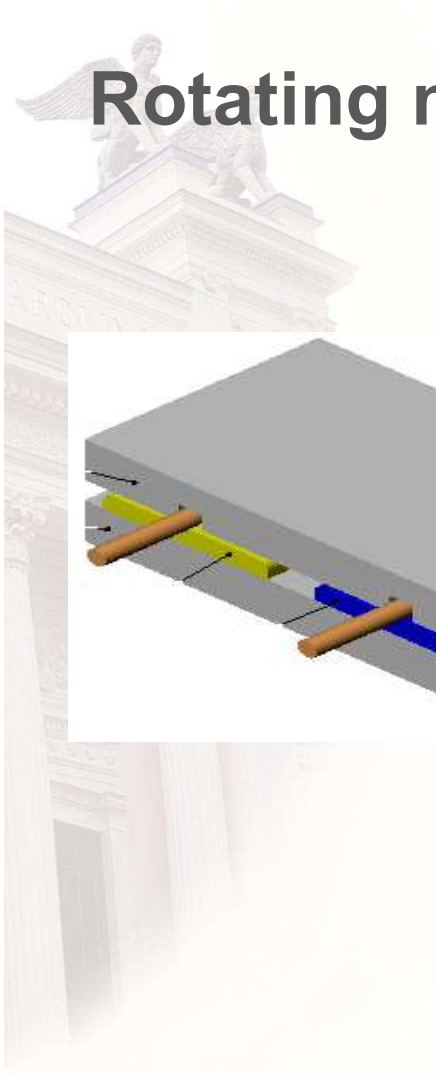
End windings don't contribute to torque, but takes a lot of space ...



Linear movement from generic force



Rotating movement from generic force





Conclusions on force and movement

- *The same generic circuit accomplish both linear and rotating movement.*
- *One phase is not enough for continuous force.*
- *Qualitative:*

Voltage ~ Speed

Current ~ Force

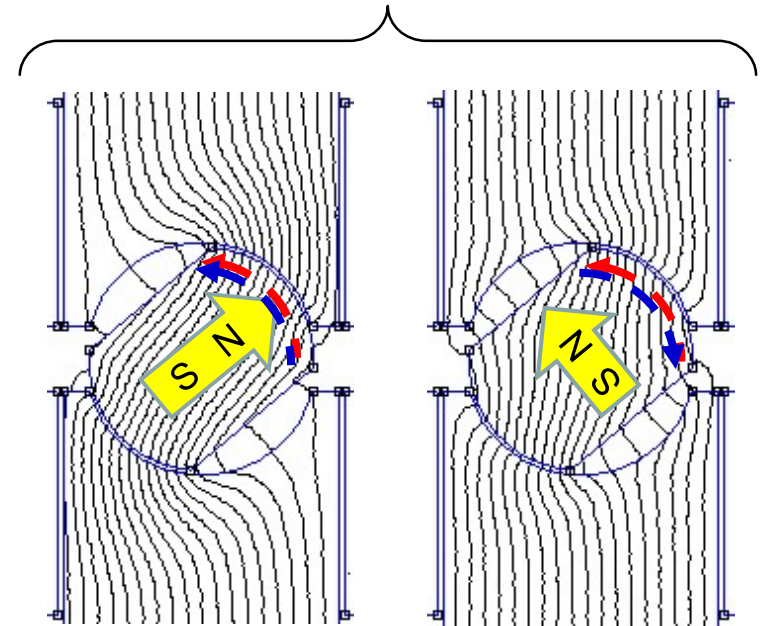
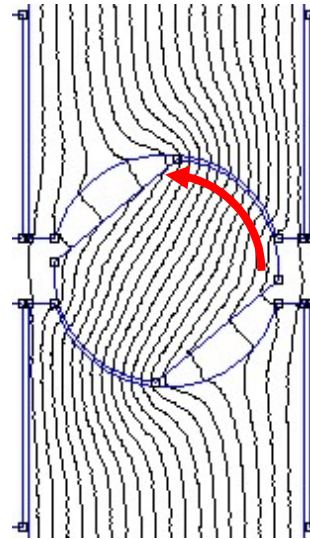
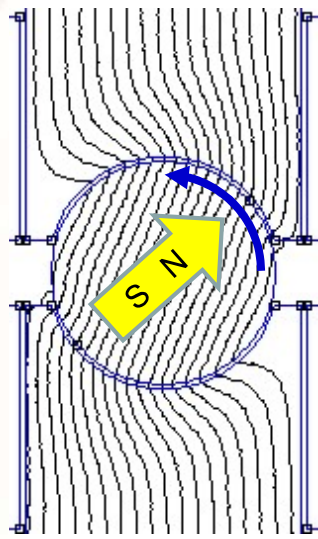
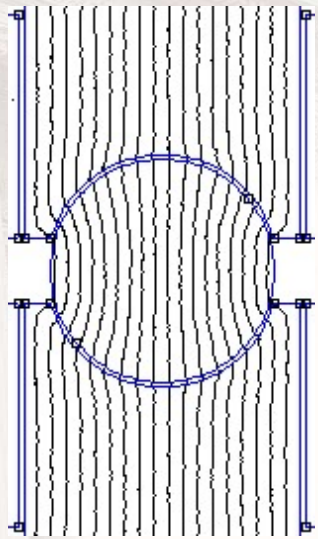
Add the Reluctance torque ...

No Torque

Electrically or permanently magnetized =
Lorentz forces

Only
Reluctance force

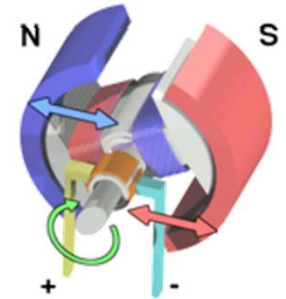
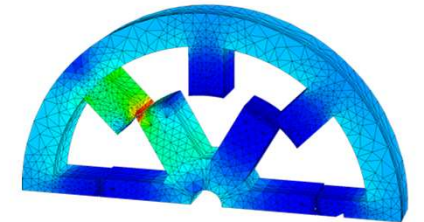
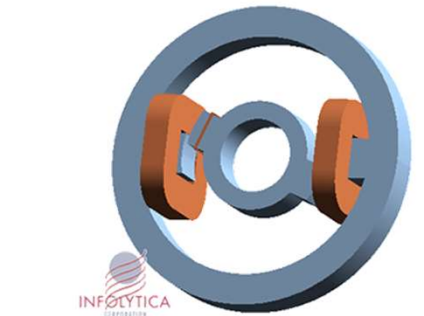
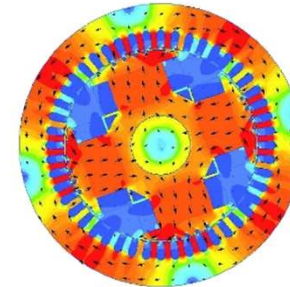
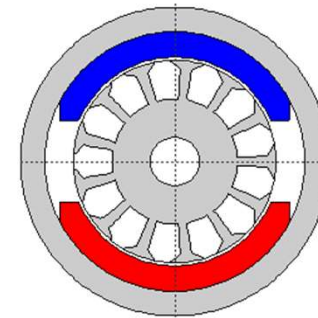
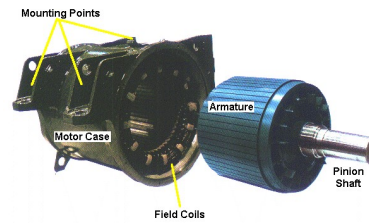
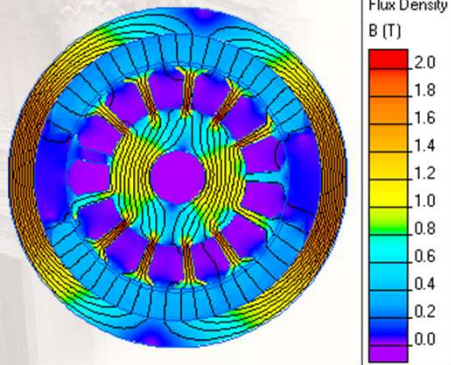
Lorentz
AND
Reluctance forces



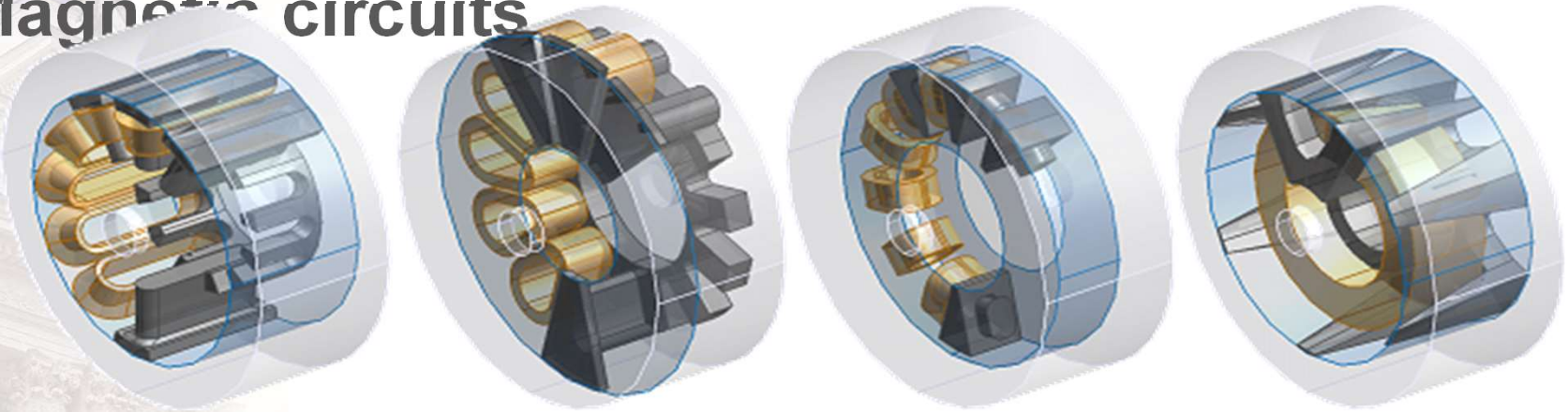
External Magnetic Field

Different mechanical arrangements

- Windings in the rotor
- Windings in the stator
- Windings in both sides



Magnetic circuits



- The **magnetic core** provides mechanic support and construction
- Soft and hard magnetic materials
 - Solid, laminated or powder cores – high μ & B_{sat} vs P_{loss}
 - Discrete, multipole magnets – high $B_r H_c$ vs cost & integration
- Establish magnetic coupling – linkage vs leakage

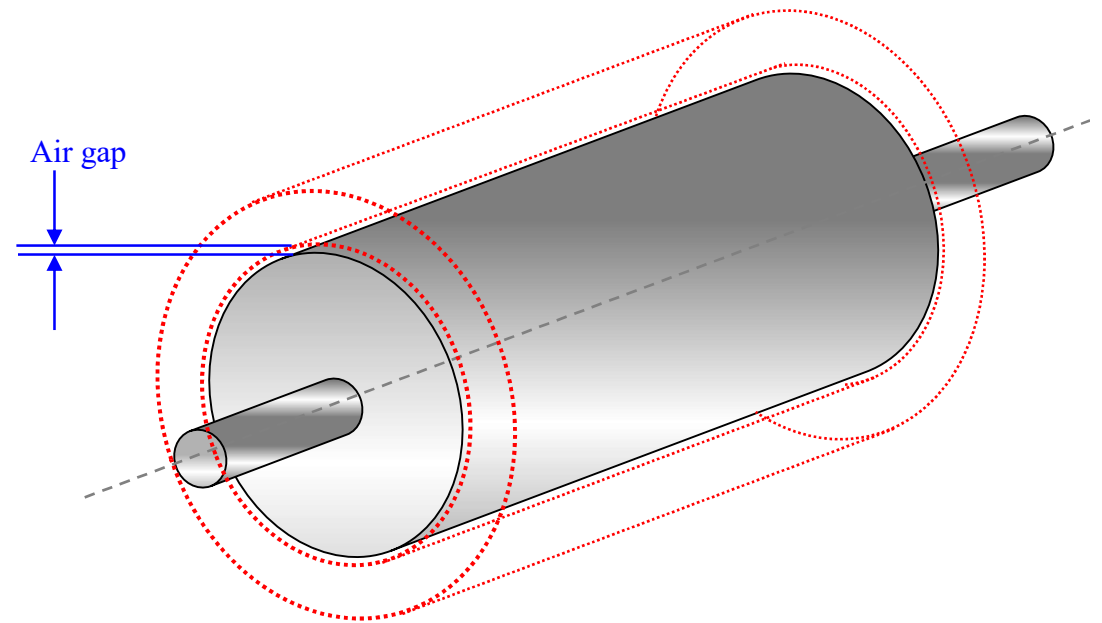
Electric circuits



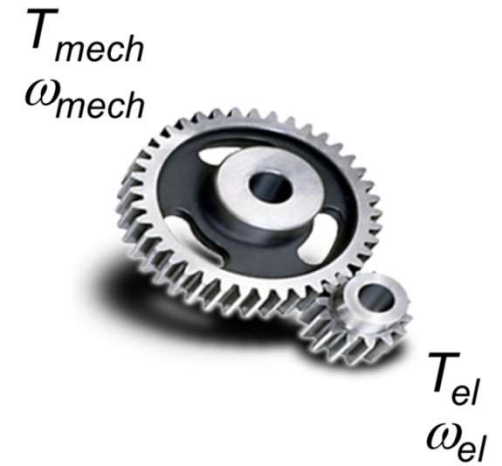
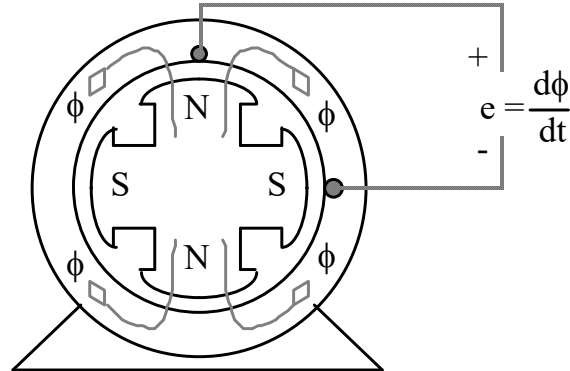
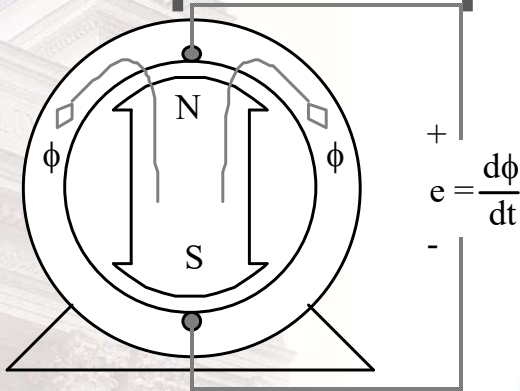
- Initially referred as 3ϕ symmetric and sinusoidal
- Insulated **electric conductor (Wire) wound as coils** or formed as waves
- Distributed or concentrated windings
 - Arrangement measured by winding factor - ratio of actual MMF to full-pitched winding MMF
- Manufacturability and assembling

Stator, Rotor and Airgap

- The **stator** is static (not moving)
- The **rotor** rotates
- The **air gap** separates them
 - Usually < 1 mm



of poles = p

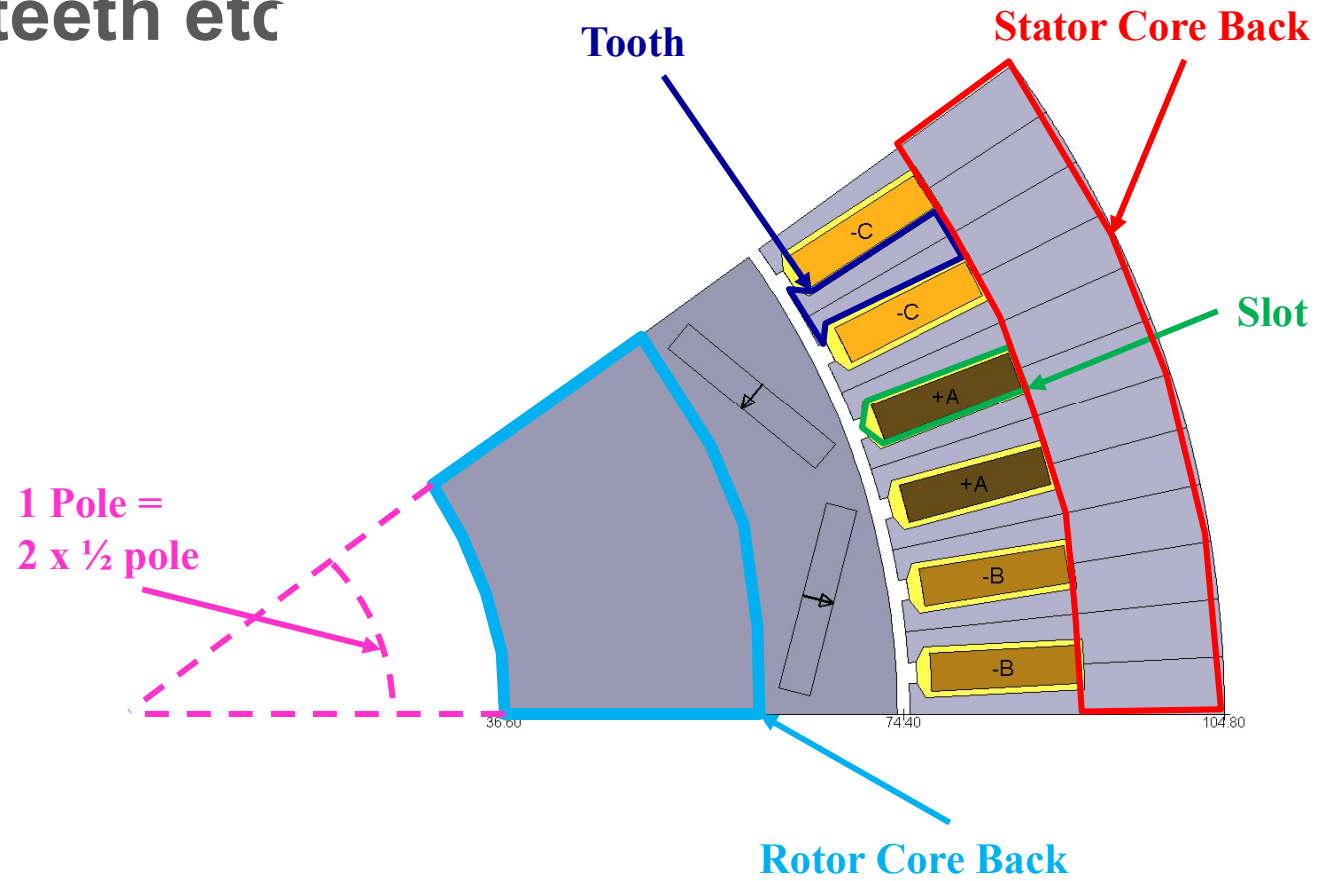


$$\omega_{el} = \frac{p}{2} \cdot \omega_{mech}$$

$$T_{mech} = \frac{p}{2} \cdot T_{el}$$

$$\text{Mechanical Power} = \omega_{mech} \cdot T_{mech} = \frac{p}{2} \cdot \omega_{el} \cdot \frac{2}{p} \cdot T_{el} = \omega_{el} \cdot T_{el}$$

Core back, teeth etc



Torque and Power

- Torque = Force * radius
on the shaft

$$T = F * r$$

- Power = Torque * Speed
on the shaft

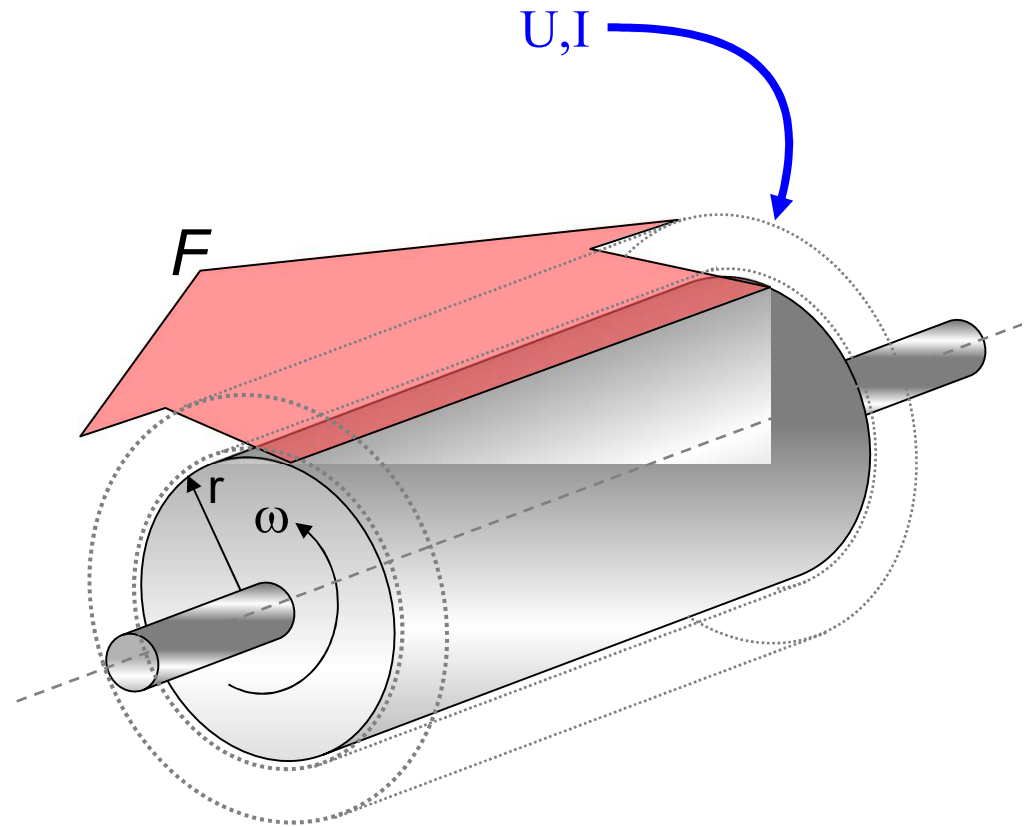
$$P = T * \omega$$

but, also ...

- Power = Voltage *
Current

on the electrical terminals

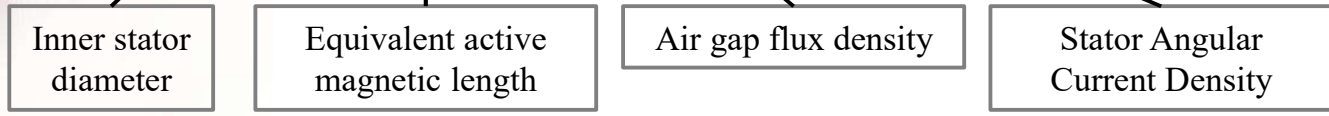
$$P = U * I$$



Tangential force

- Interaction of flux and current
- "F = B i l"

$$F_{\text{tangential}} = \frac{\pi}{2} D_{is} l_e B_{gm1} K_{s1}$$



Stator current / meter air gap periphery

Angular current density

Number of winding turns

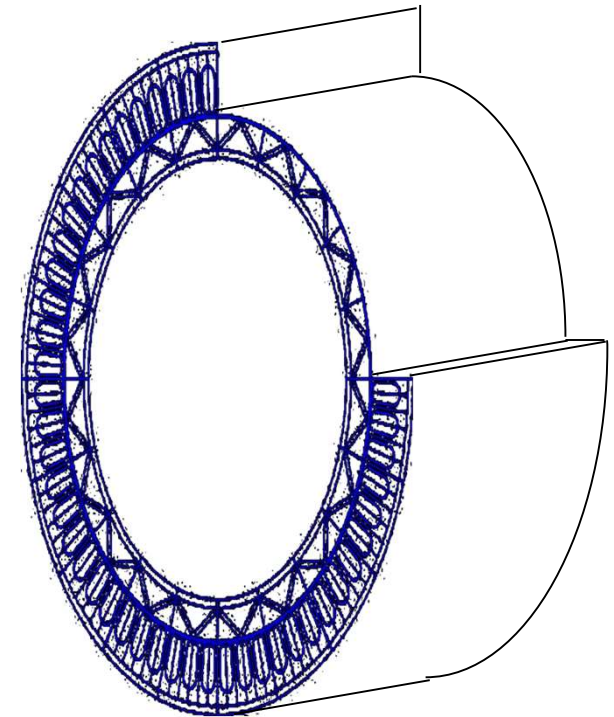
$$K_s(\theta) = \frac{6}{\pi} \cdot \frac{k_1 N_s}{D_{is}} \cdot \hat{i}_s \cdot \cos\left(\frac{p\theta}{2} - \omega_e t\right) \quad [A/m]$$

Winding factor

Number of poles

Phase current amplitude

Stator inner diameter

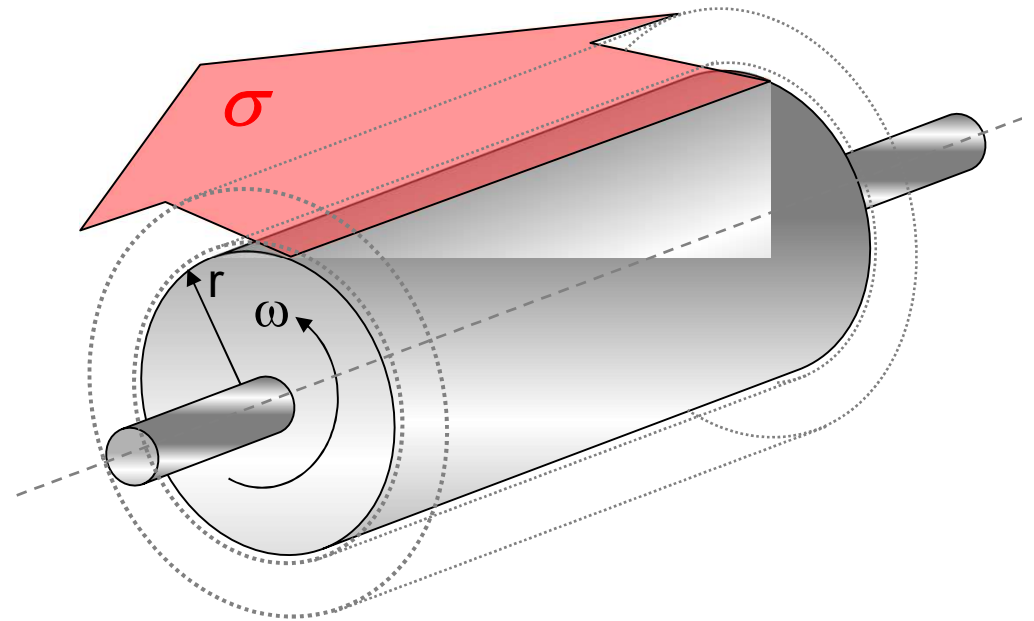


Shear Force & Torque

- Current and Flux interact for tangential force
 σ = Force/Unit area is a key figure
- A good design accomplish about
 $\sigma = 10\ 000 \dots 30\ 000 \text{ [N/m}^2\text{]}$
- ... in continuous operation and
 2...4 times that in transient operation

$$\sigma = \left(\frac{F}{A} \right)_{avg} = \frac{\frac{\pi}{2} D_{is} l_e B_{gm1} K_{s1}}{\pi D_{is} l_e} = \frac{B_{gm1} K_{s1}}{2} \quad [\text{N/m}^2]$$

$$T = \frac{\pi}{4} D_{is}^2 l_e \cdot B_{gm1} K_{s1}$$



Conclusion on torque

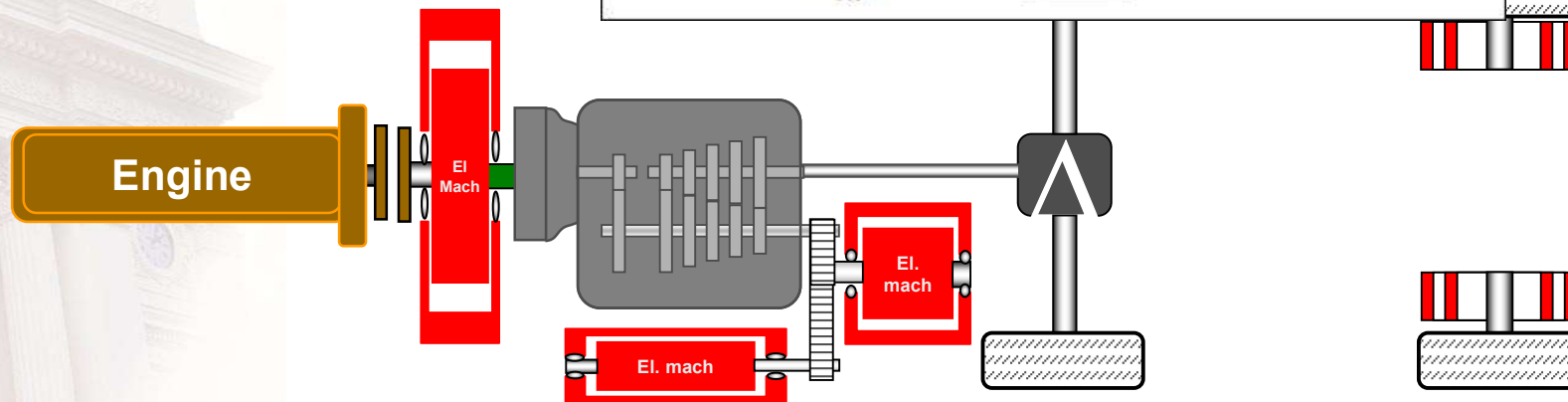
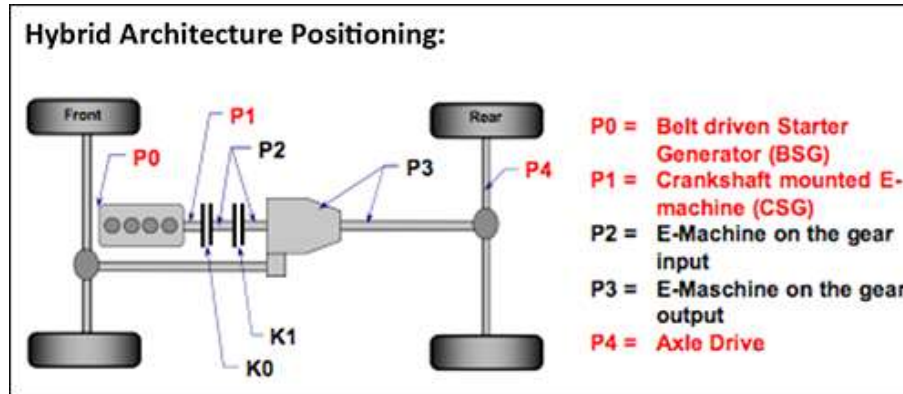
The torque is proportional to the:

- Magnetic flux density – Limited by material properties to about 1.0 ... 1.5 Tesla
- Spatial current "density" – Limited by cooling capability
- Axial length of the machine
- Diameter SQUARED !

} = Rotor Volume

$$T = \frac{\pi}{4} D_{is}^2 l_e \cdot B_{gm1} K_{s1}$$

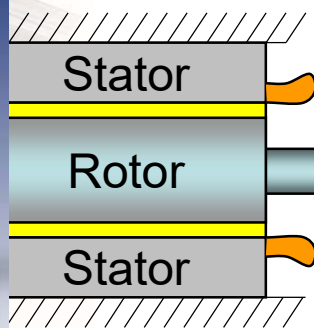
Hybrid Topologies



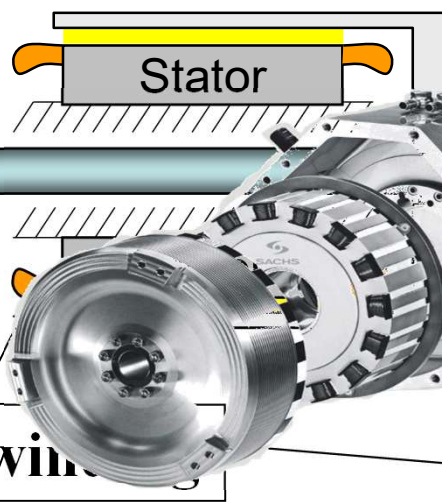
Inner or Outer Rotor, Radial or Axial flux ...



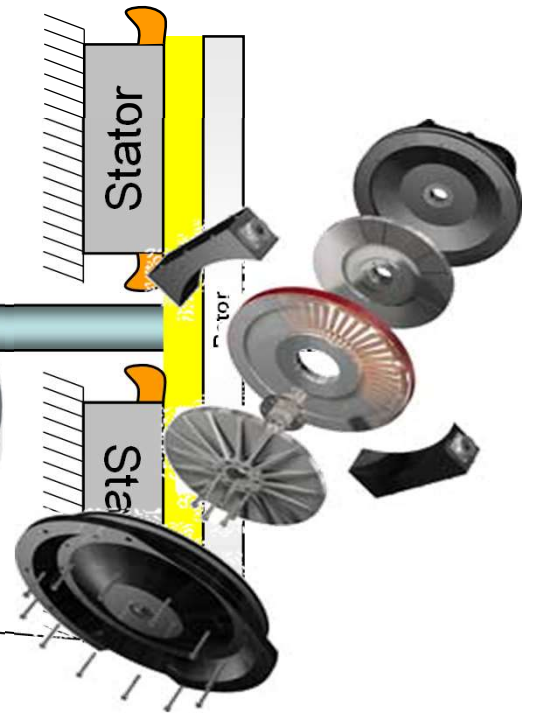
**Inner rotor
Radial flux**



**Outer rotor
Radial flux**



Axial flux



Distributed or Concentrated winding



Axially shorter end winding
Cheaper assembly
Lower torque quality



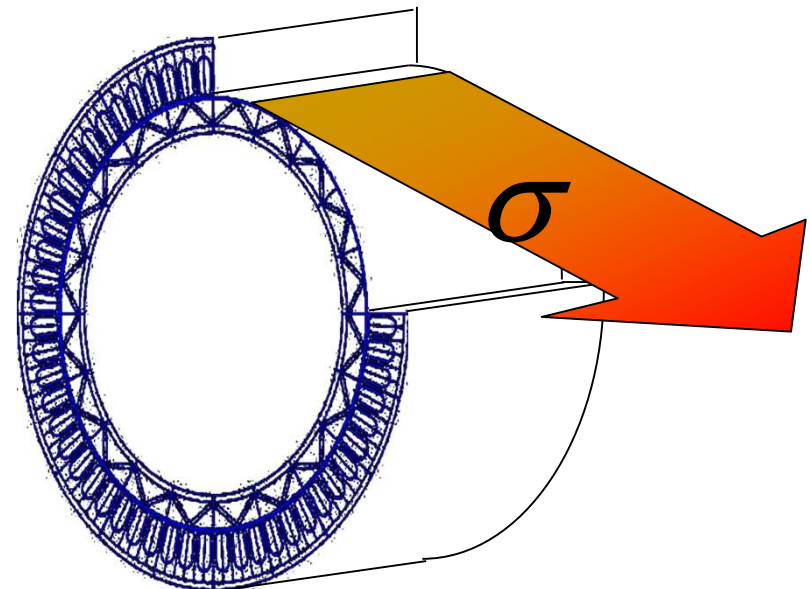
Longer end winding
More expensive assembly
Higher torque quality

Shear Force & Torque

Current and Flux interact for tangential force

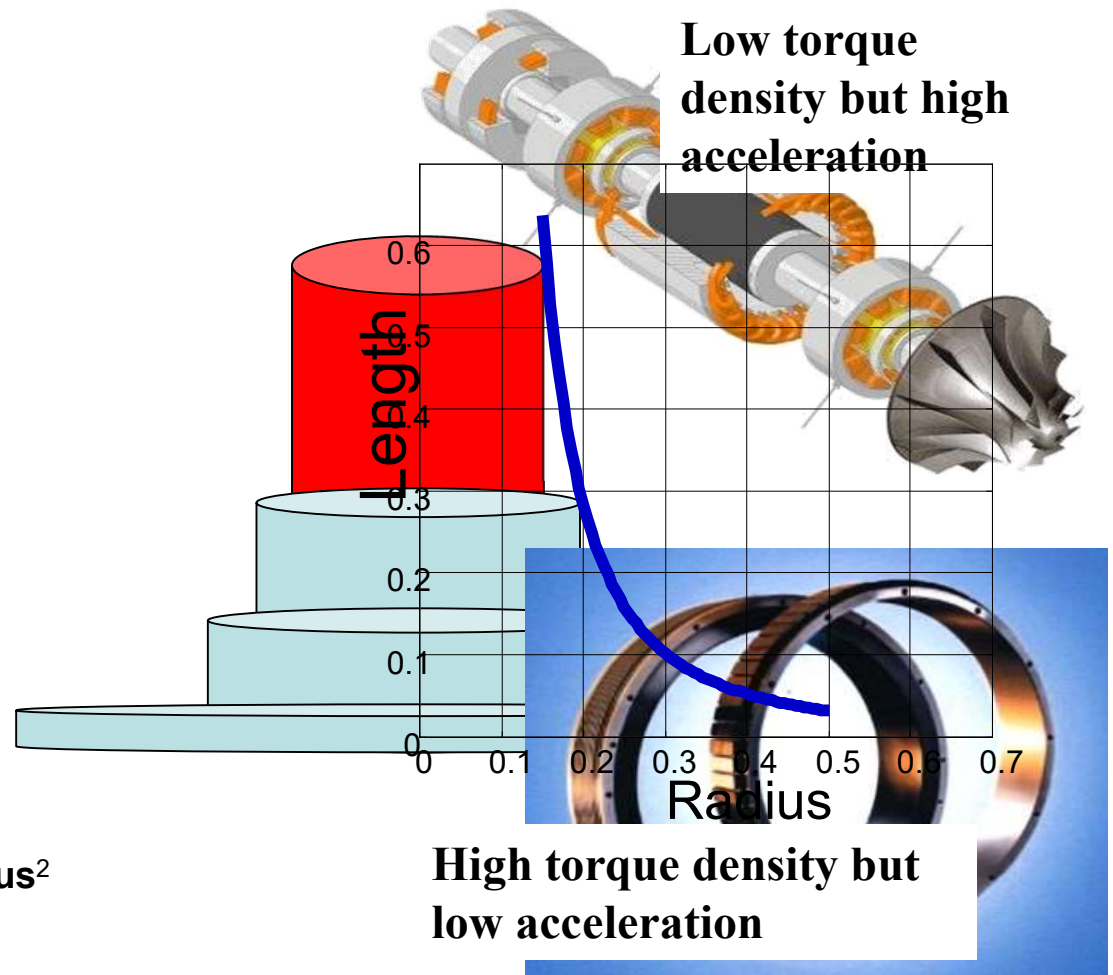
σ = Force/Unit area is a key figure

A good design accomplish about $\sigma = 10000-30000$ [N/m²] in continuous operation and 2..4 times more in transient operation

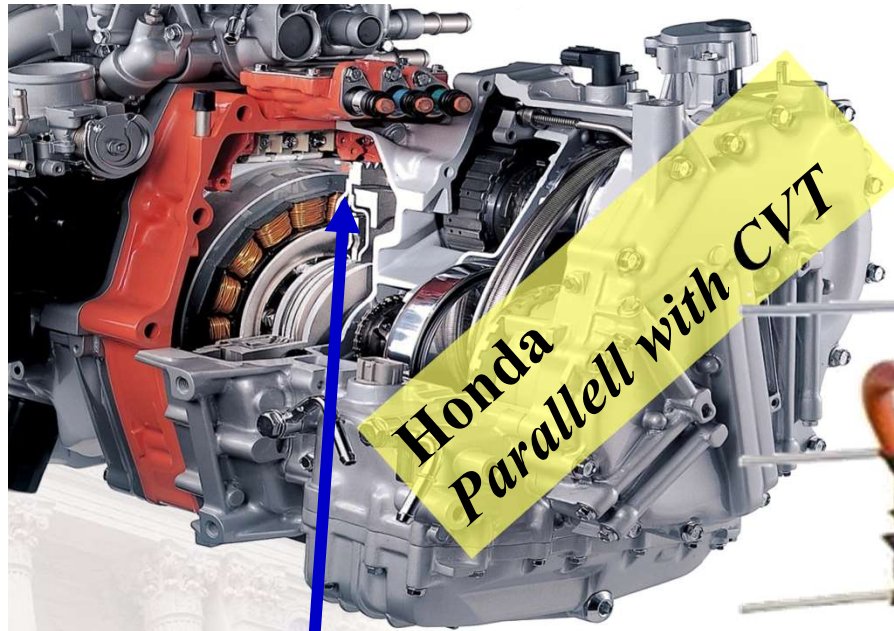


Form Factor

- For the same torque, a machine can be either short and wide, or long and slender ...
- Assume 25000 [N/m²], and a desired torque of 1000 Nm, AND that the stator outer radius is 0.15 – 0.5 meter.
 - How long will the machine be to fulfill the torque requirement?
- **The long and slender machine will accelerate faster**
 - Torque \sim radius²*length
 - Inertia \sim radius⁴*length
- **Acceleration = Torque/Inertia \sim 1/radius²**

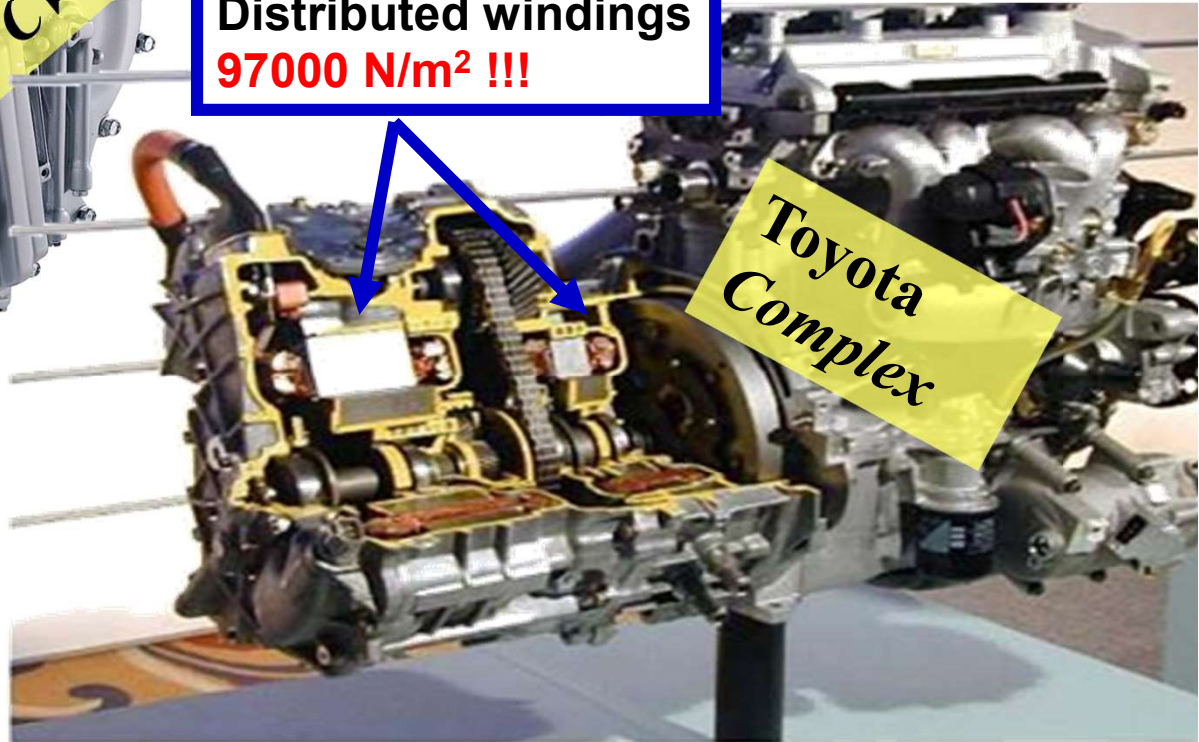


Ποιοι αλλη
Toyota



**Honda
Parallel with CVT**

**Complex
PM Machines
Distributed windings
97000 N/m² !!!**



**Toyota
Complex**

**"ISAM"
PM Motor
Concentrated
windings"**

Volvo V90 T8

Conventional 4-cylinder gasoline
320 hp / 240 kW

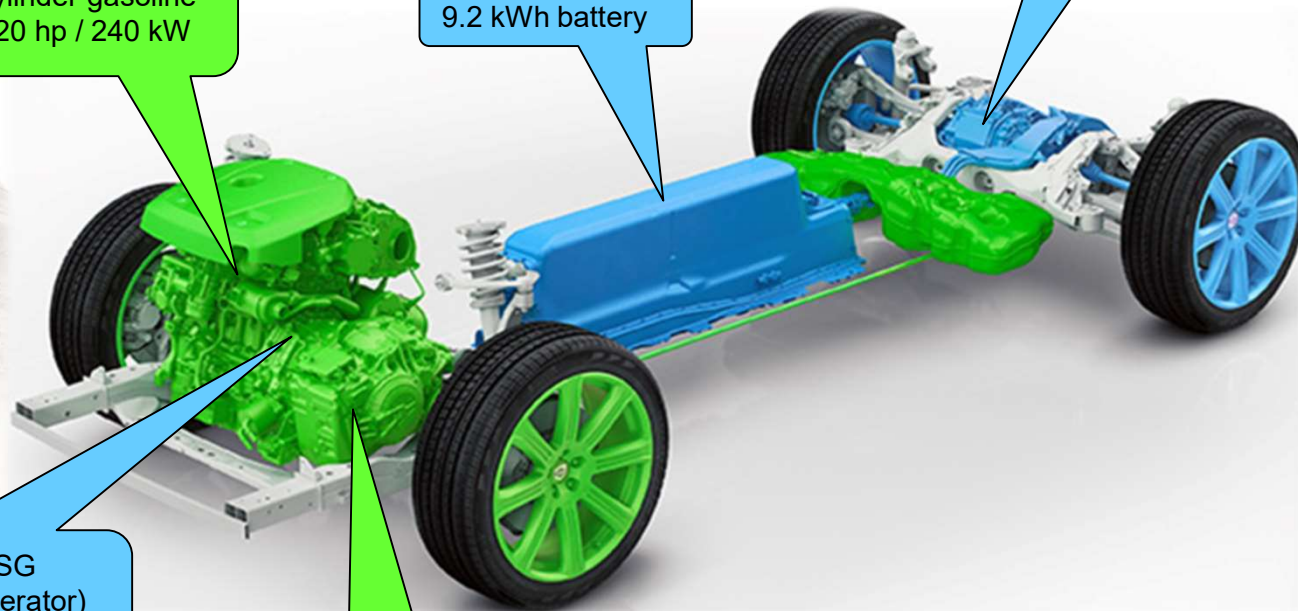
9.2 kWh battery

PM machine
electric axle
82 hp / 61 kW



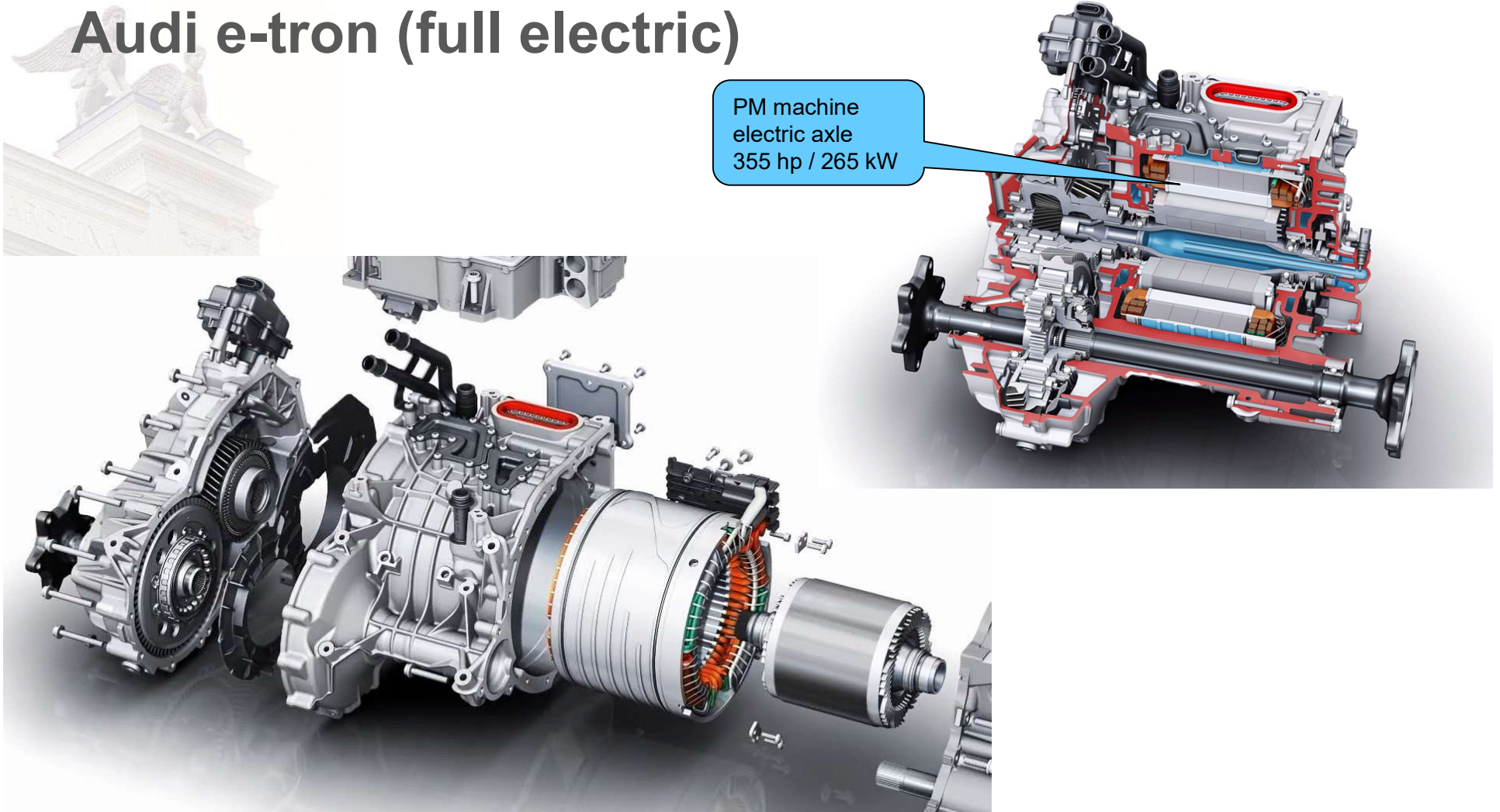
Crankshaft mounted ISG
(Integrated Starter generator)
45 hp / 34 kW

8-speed automatic
transmission

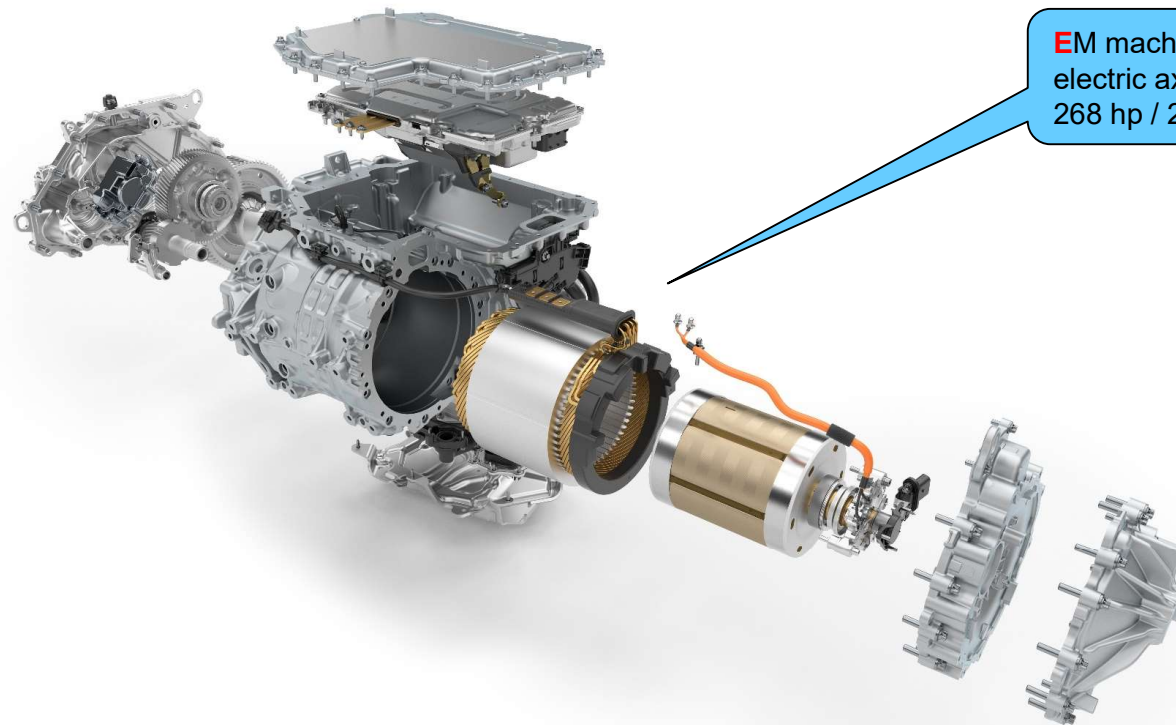


Audi e-tron (full electric)

PM machine
electric axle
355 hp / 265 kW



BMW iX3 drive train



EM machine
electric axle
268 hp / 200 kW



Conclusions on force and movement

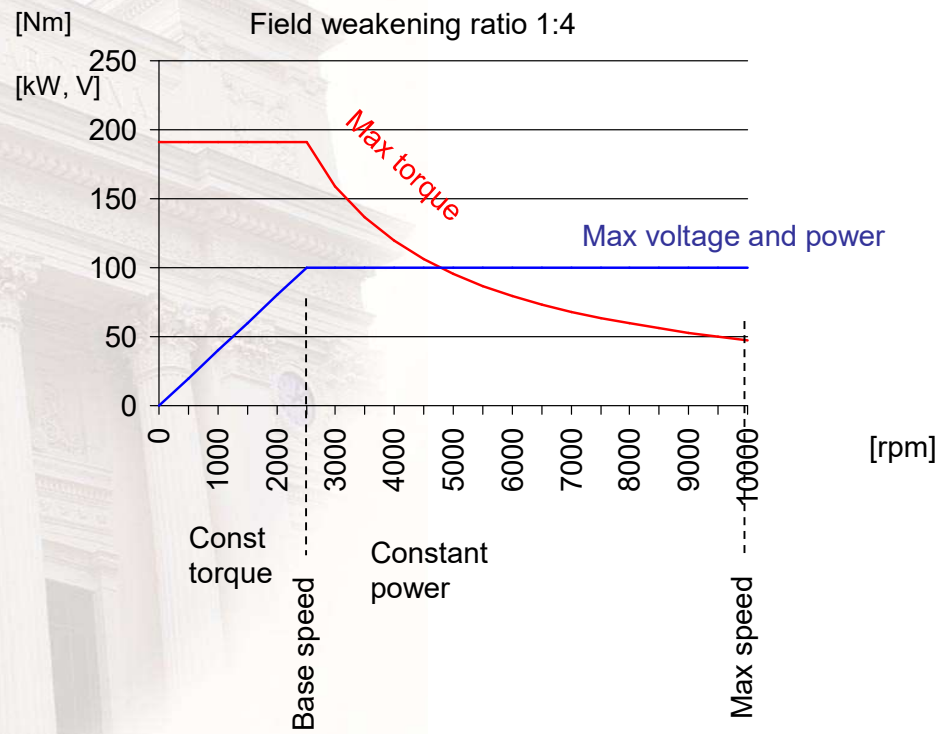
- *The same generic circuit accomplish both linear and rotating movement.*
- *One phase is not enough for continuous force*
- *Qualitative:*

- *Voltage ~ Speed*
- *Current ~ Force*

Field Weakening : I

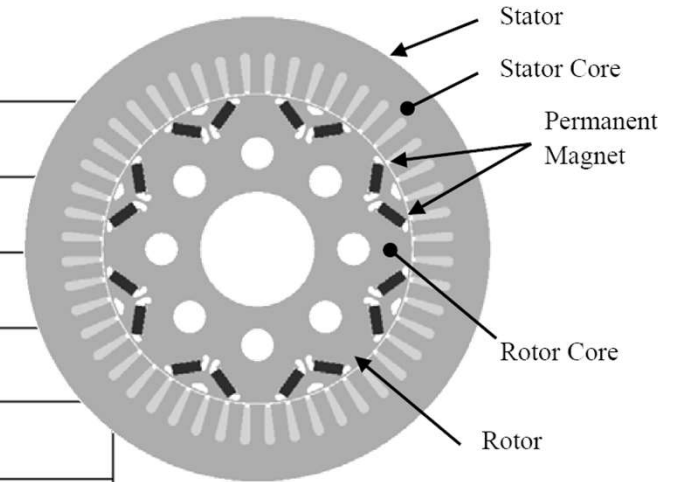
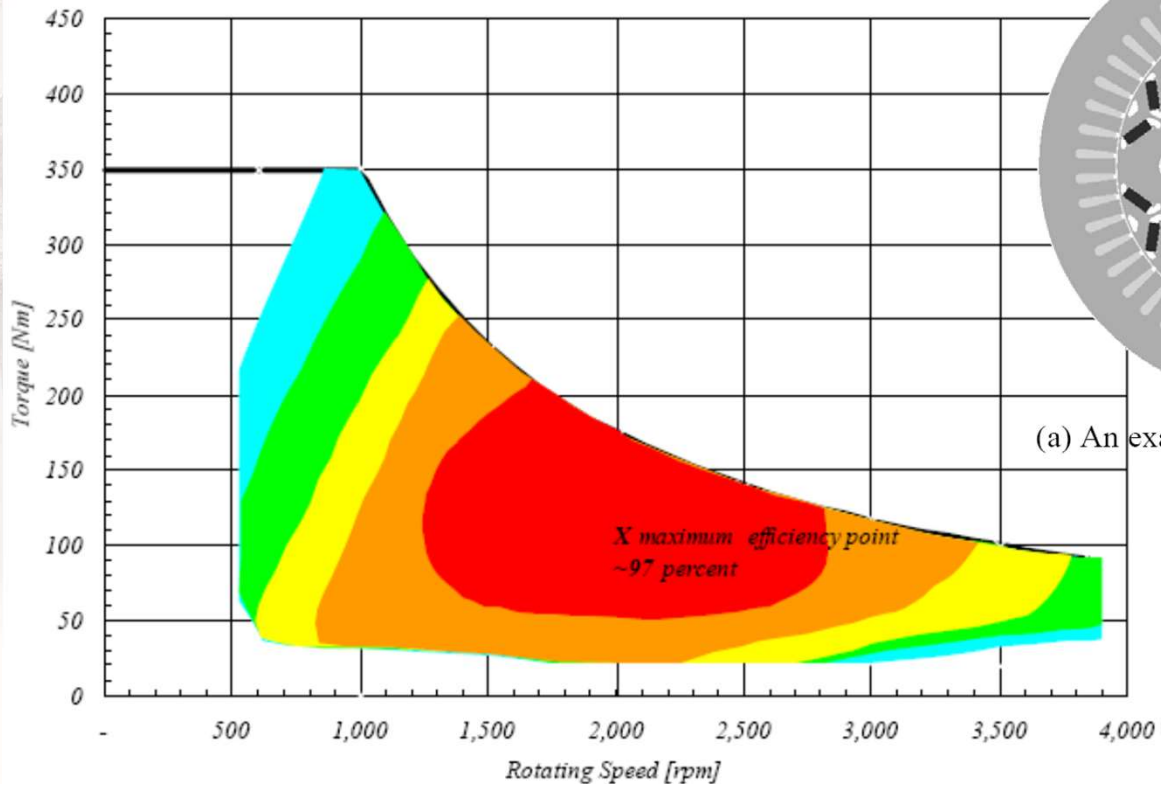
- Remember:
 - Voltage ~ flux density * speed
 - Torque ~ flux density * current
 - Power ~ speed*torque = voltage*current
- The required voltage "hits the roof" at some speed. What to do, to increase speed beyond?
 - Answer: Reduce flux density'
 - Consequences:
 - *The voltage requirement is kept constant, as desired*
 - *The torque capability drops as the flux density.*
 - *The power is kept constant, since the speed increases in the same rate as the torque drops with increasing speed.*

Field Weakening : II



Example from Toshiba

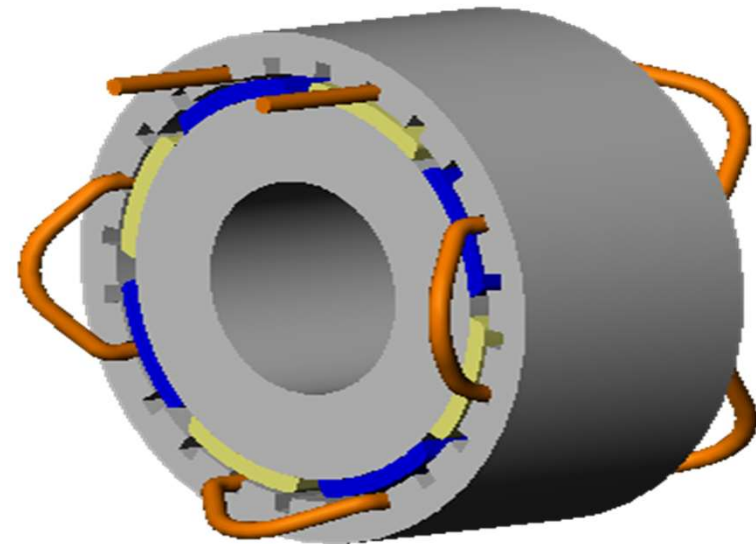
”Large Torque and High Efficiency Permanent Magnet Reluctance Motor for A Hybrid Truck” - Masanori Arata et. Al, EVS-22



(a) An example of cross section of the PRM

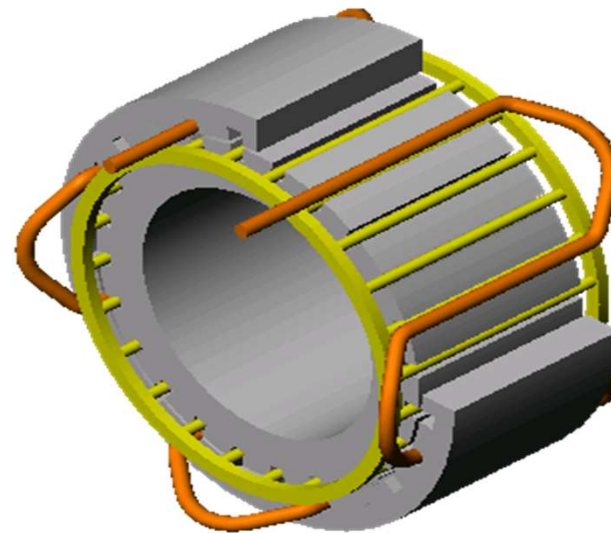
Permanent Magnet Synchronous Machines

- Same as the generic machine
- Voltage and frequency proportional to speed
- Current proportional to torque
- High torque density
 - 1...10 Nm/kg
 - Compare to ICE 1...2 Nm/kg
- High efficiency
 - Up to 97%
- Higher efficiency, higher torque density and more expensive than other machines
 - Due to the permanent magnets.

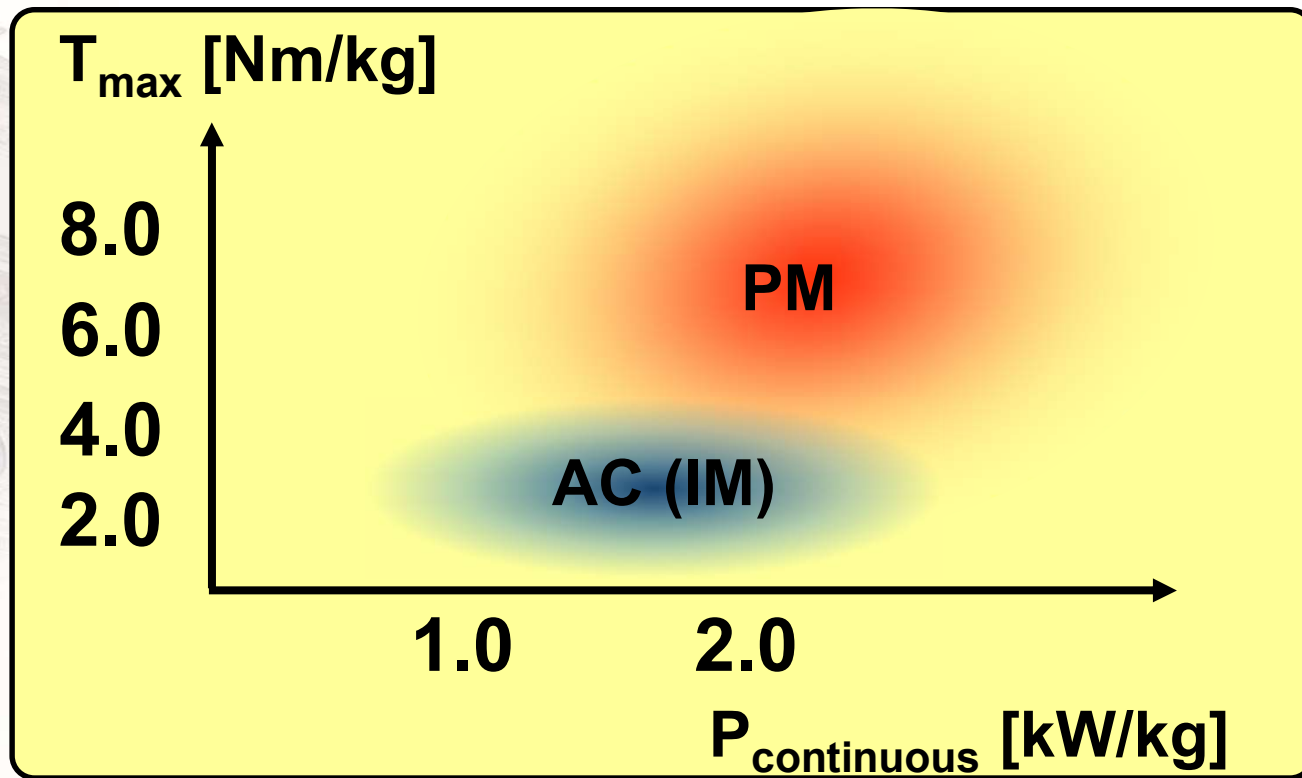


The Induction Machine : I

- Same stator as the PMSM
- The rotor is a short circuited "cage"
- The rotor current must be induced magnetically
 - Losses related to magnetization "competes" with losses due to torque generation.
- Robust construction
- Low cost
- Low/no maintenance
- Heavily standardized for industrial applications
- Voltage and frequency proportional to speed, like PMSM



Comparison – Electric Drives



Electrical machine losses

- Several types:

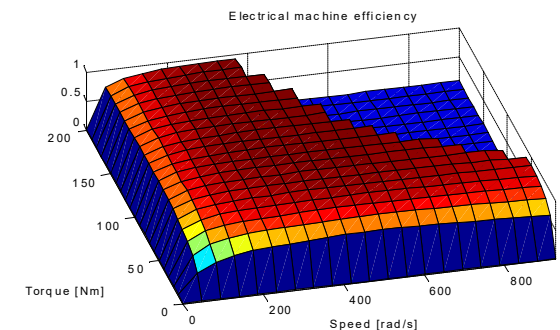
- Copper losses (Ri^2)
- Iron losses ($k_1*f*B^2 + k_2*(f*B)^2$)
- Windage losses (surface speed)
- Other friction losses

- Approximately calculated by

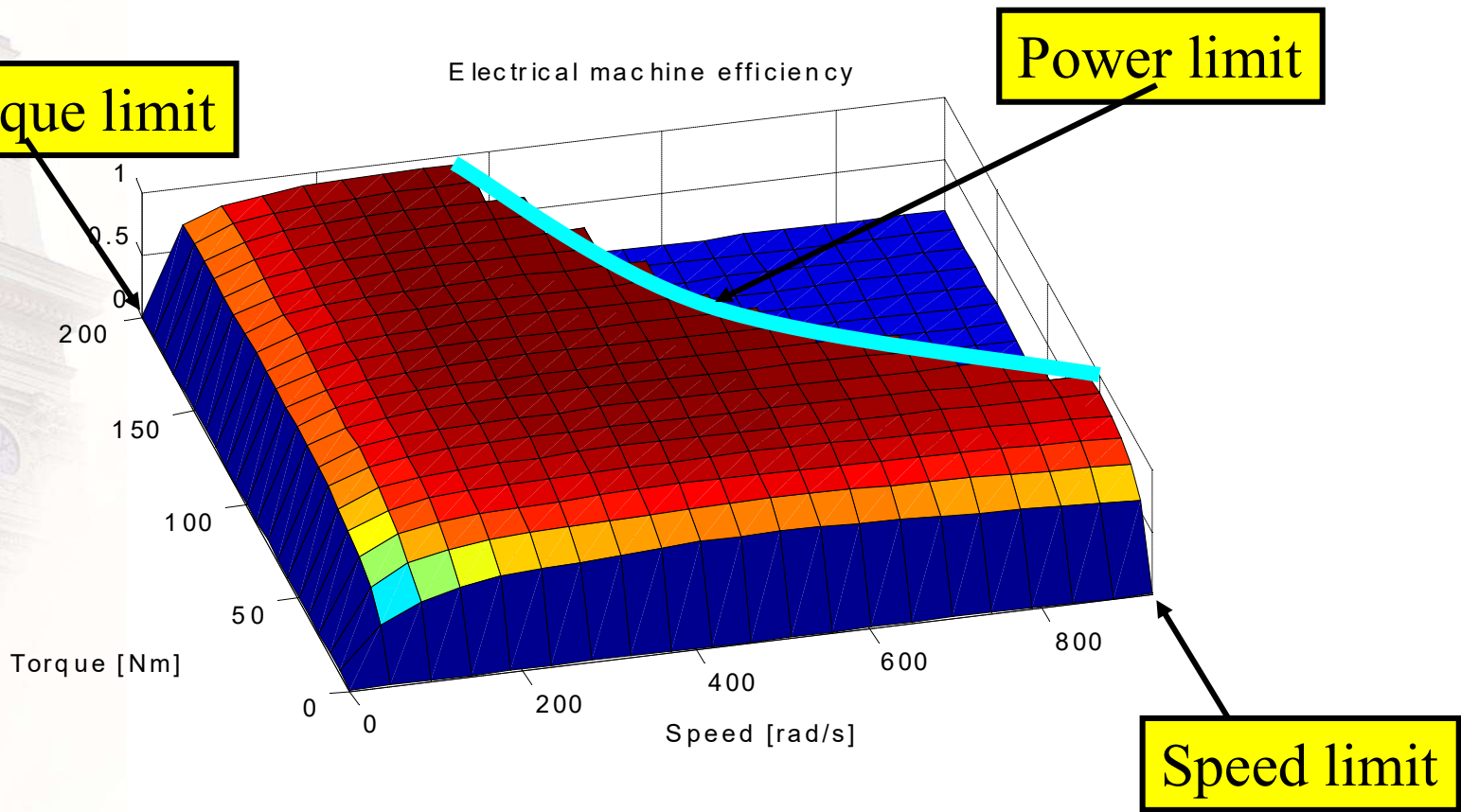
- `[EtaEM, Tem, Wem] = CreateEMmap(Pem_max, wem_max, Tem_max)`

Hysteresis losses

Eddy current losses



The Traction motor efficiency

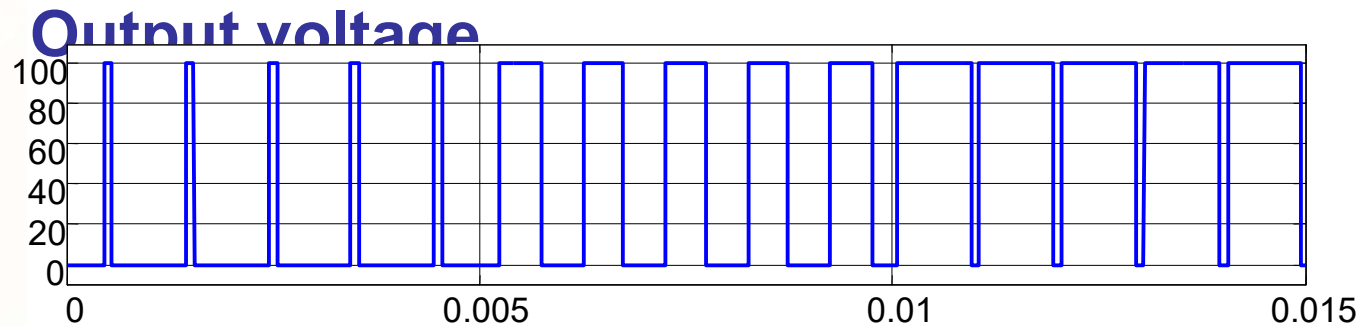
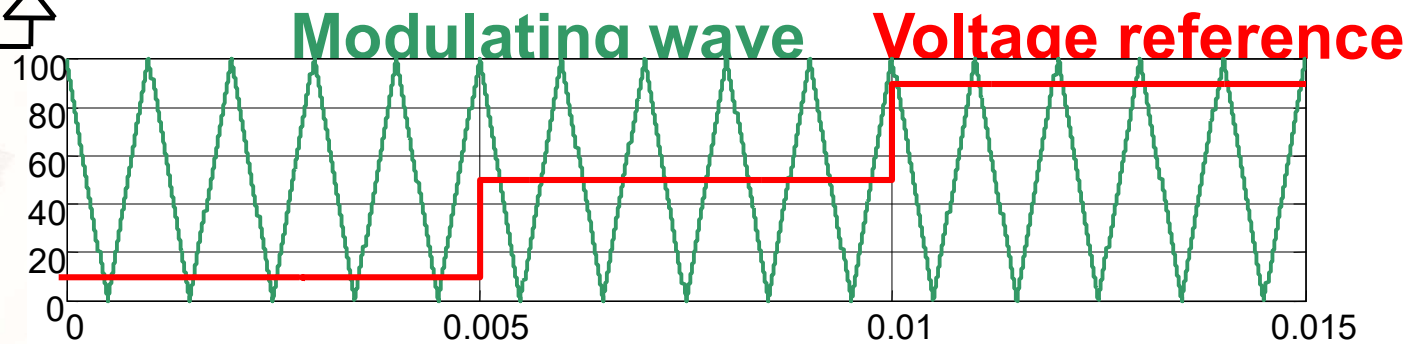
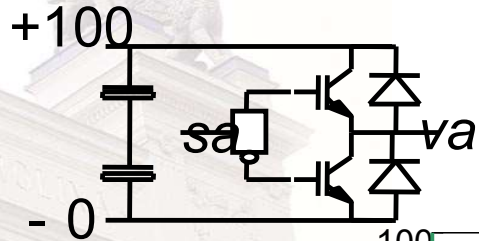




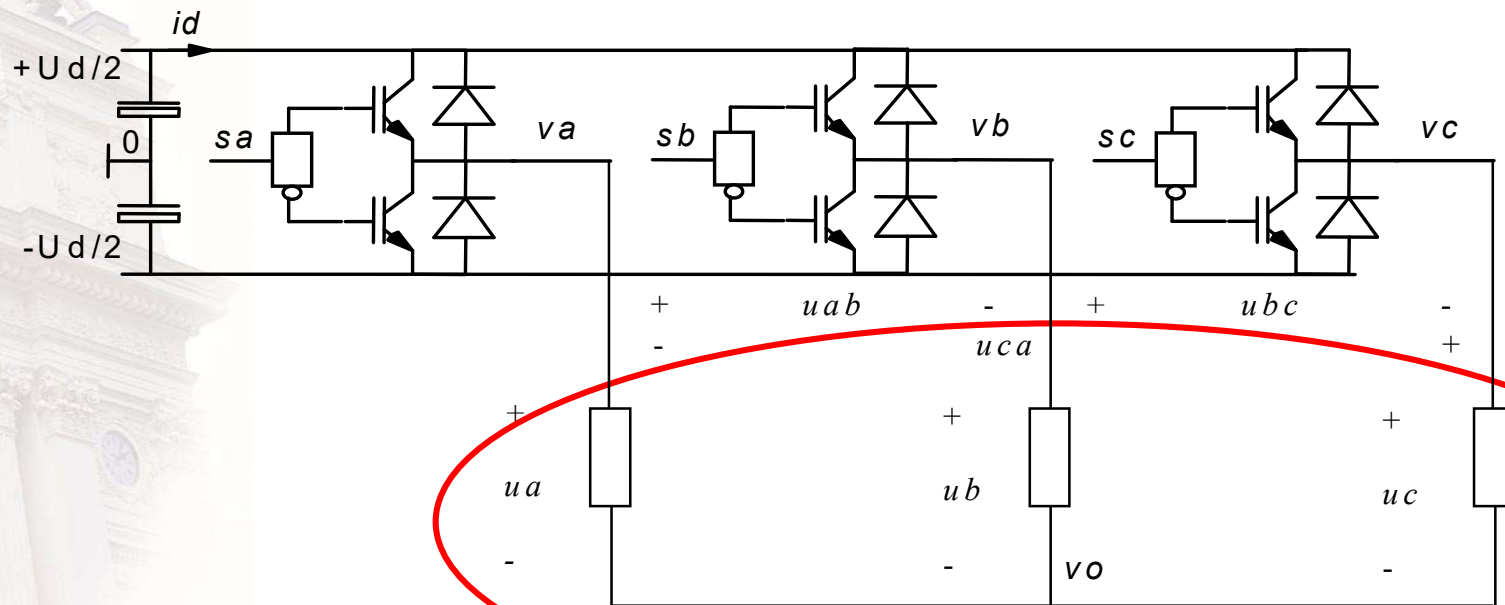
Power Electronics

- Needed to condition the battery voltage to the different electrical drives
- Use switching technology for high efficiency
 - Conventional converters (like loudspeaker amplifiers) efficiency 25-60 % due to continuous control of the voltage.
 - Switching means “on/off” control of voltage, leading to efficiency above 95 %.

1 Phase Pulse Width Modulation (PWM)



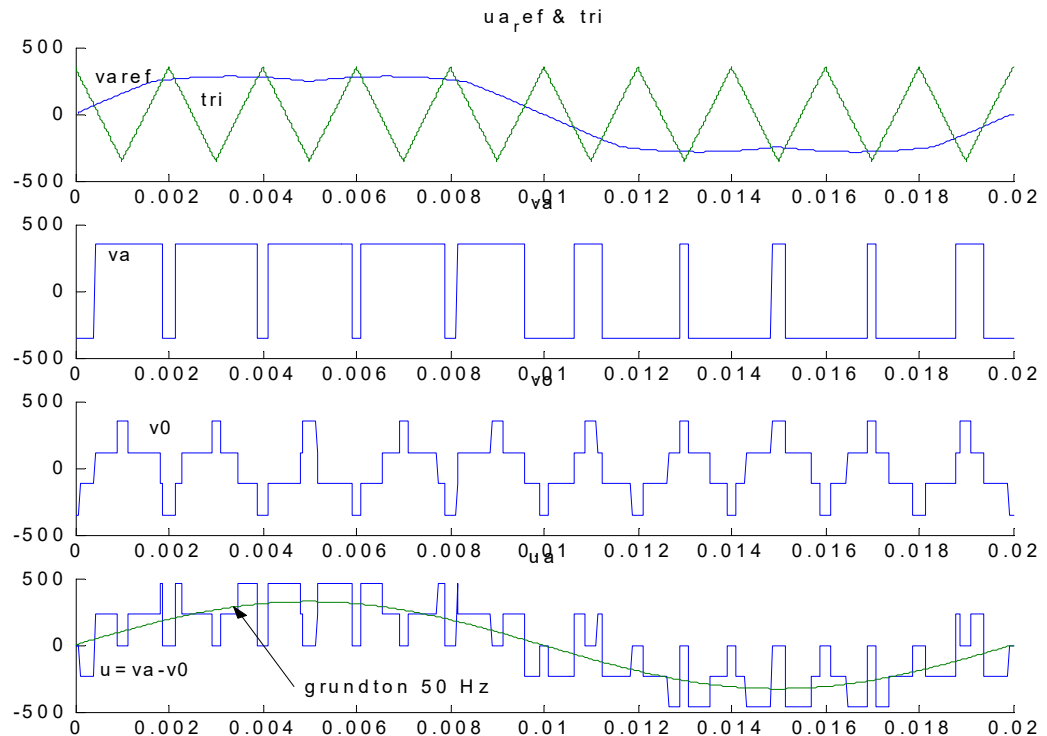
Three-phase Converters



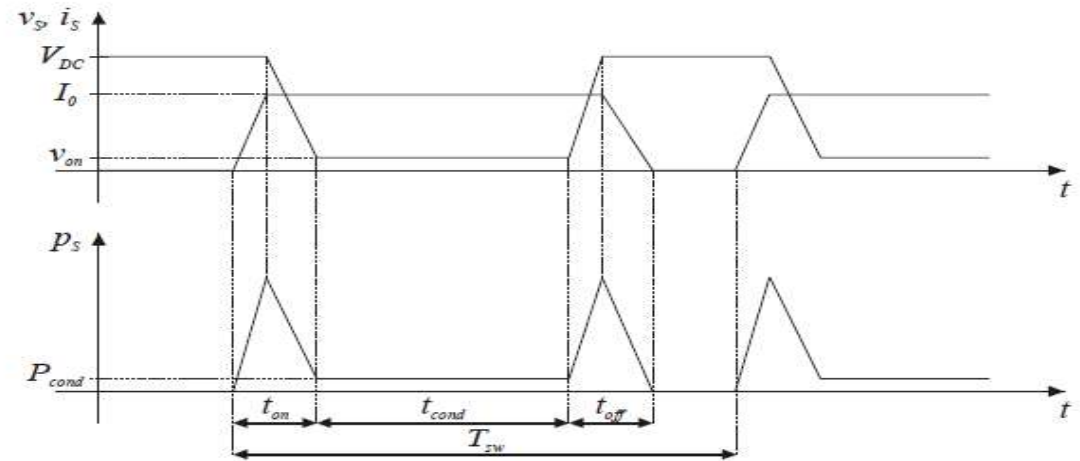
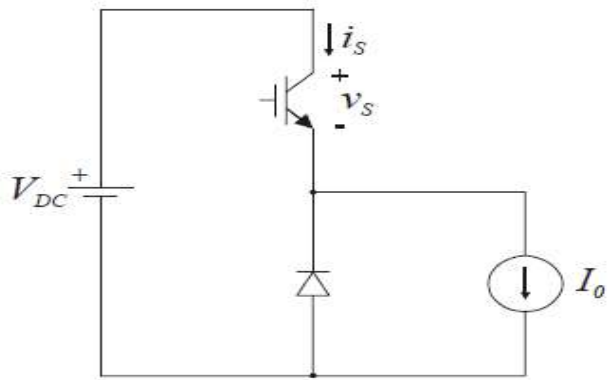
Traction motor

Three-phase Pulse Width Modulation

- The voltages contain high harmonics that cause:
 - Audible noise
 - Torque ripple
 - EMC problems



Simple Converter Loss Model



$$p_S(t) = v_S(t) \cdot i_S(t)$$

Switching and Conduction losses

Energy losses: $E_S(T_{sw}) = \int_{T_{sw}} p_S(\tau) d\tau = E_{S,on}(T_{sw}) + E_{S,cond}(T_{sw}) + E_{S,off}(T_{sw})$

$$E_{S,on}(T_{sw}) = \int_{t_{on}} p_S(\tau) d\tau = V_{DC} \cdot I_0 \cdot \frac{t_{on}}{2}$$

$$E_{S,cond}(T_{sw}) = \int_{t_{cond}} p_S(\tau) d\tau = V_{S(on)} \cdot I_0 \cdot t_{cond} \quad \text{Note} \quad V_{S(on)} = V_{S0} + R_S \cdot I_0$$

$$E_{S,off}(T_{sw}) = \int_{t_{off}} p_S(\tau) d\tau = V_{DC} \cdot I_0 \cdot \frac{t_{off}}{2}$$

Power losses: $P_S(T_{sw}) = \frac{E_S(T_{sw})}{T_{sw}} = P_{S,on}(T_{sw}) + P_{S,cond}(T_{sw}) + P_{S,off}(T_{sw})$

$$P_{S,on}(T_{sw}) = \frac{E_{S,on}(T_{sw})}{T_{sw}} = E_{S,on}(T_{sw}) \cdot f_{sw} = \frac{V_{DC} \cdot I_0 \cdot t_{on}}{2} \cdot f_{sw}$$

$$P_{S,cond}(T_{sw}) = \frac{E_{S,cond}(T_{sw})}{T_{sw}} = V_{S(on)} \cdot I_0 \cdot \frac{t_{cond}}{T_{sw}} = V_{S(on)} \cdot I_0 \cdot D_S$$

$$P_{S,off}(T_{sw}) = \frac{E_{S,off}(T_{sw})}{T_{sw}} = E_{S,off}(T_{sw}) \cdot f_{sw} = \frac{V_{DC} \cdot I_0 \cdot t_{off}}{2} \cdot f_{sw}$$



$$P_{S,sw}(T_{sw}) = P_{S,on}(T_{sw}) + P_{S,off}(T_{sw})$$

Reverse recovery Losses

If specified, use:

$$E_{S,on}(T_{sw}) = \frac{E_{on,n}}{V_{DC,n} \cdot I_{0,n}} \cdot V_{DC} \cdot I_0$$

$$E_{S,off}(T_{sw}) = \frac{E_{off,n}}{V_{DC,n} \cdot I_{0,n}} \cdot V_{DC} \cdot I_0$$

For the freewheeling diode:

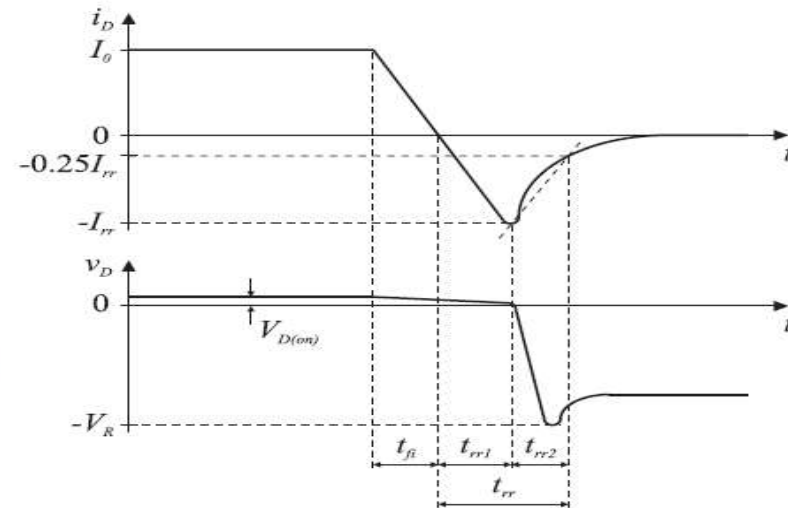
$$P_{D,cond}(T_{sw}) = V_{D(on)} \cdot I_0 \cdot D_D \quad V_{D(on)} = V_{D0} + R_D \cdot I_0$$

$$D_D \approx 1 - D_S$$

$$P_{D,rr} = V_{DC} \cdot Q_f \cdot f_{sw} \quad Q_f \approx \frac{1}{S+1} \cdot Q_{rr} \quad \text{where } S = \frac{t_{rr1}}{t_{rr2}}$$

If specified, use:

$$P_{D,off} = E_{D,off}(T_{sw}) \cdot f_{sw} \quad E_{D,off}(T_{sw}) = \frac{E_{off,n}}{V_{DC,n} \cdot I_{0,n}} \cdot V_{DC} \cdot I_0$$

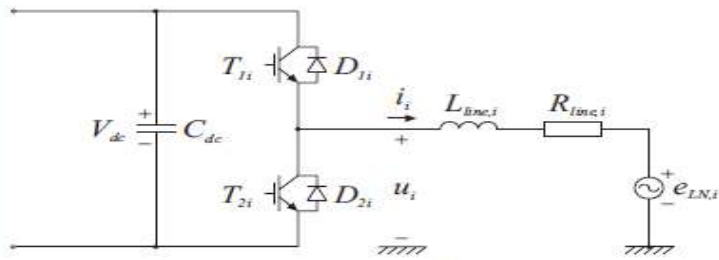


Fig

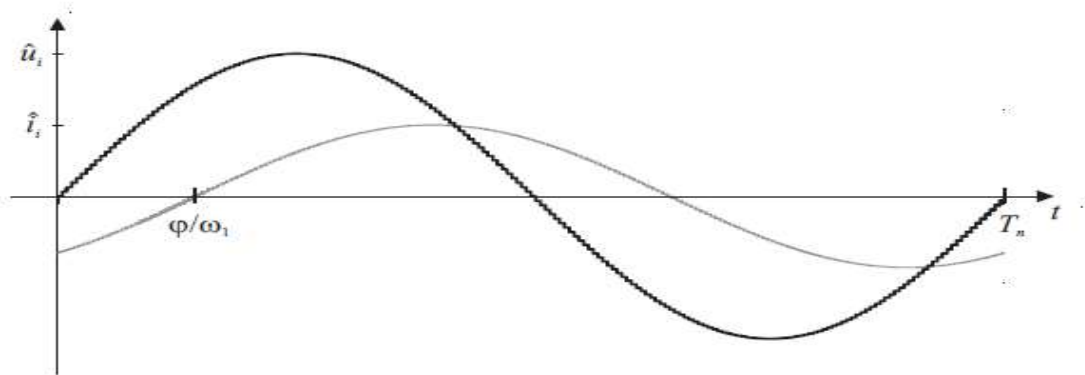
$$Q_f = \frac{Q_{f,n}}{I_{0,n}} \cdot I_0$$



3-phase converter losses



One half-bridge of a three-phase voltage source converter.



Converter output voltage and current. The current is displaced by an angle φ relative to the voltage.

Loss estimation

Switching losses:

$$\bar{P}_{Ti,sw} = \frac{1}{T_n} \int (P_{on} + P_{off}) dt = \frac{f_{sw}}{T_n} \int (E_{on} + E_{off}) dt = \frac{E_{on,n} + E_{off,n}}{V_{dc,n} \cdot I_n} \cdot \frac{V_{dc} f_{sw}}{T_n} \int |\hat{i}_i \sin(\omega_1 t - \varphi)| dt = \frac{2\sqrt{2}}{\pi} \cdot \frac{E_{on,n} + E_{off,n}}{V_{dc,n} \cdot I_n} \cdot V_{dc} I_i f_{sw}$$

$$\bar{P}_{Di,sw} = \frac{1}{T_n} \int (P_{on} + P_{off}) dt = \frac{f_{sw}}{T_n} \int (E_{on} + E_{off}) dt = \frac{2\sqrt{2}}{\pi} \cdot \frac{E_{off,n}}{V_{dc,n} \cdot I_n} \cdot V_{dc} I_i f_{sw} = \frac{2\sqrt{2}}{\pi} \cdot \frac{E_{Drr,n}}{V_{dc,n} \cdot I_n} \cdot V_{dc} I_i f_{sw}$$

Conduction losses:

$$\bar{P}_{Ti,cond} = \left(\frac{\sqrt{2}}{\pi} \cdot V_{T0} I_i + \frac{1}{2} \cdot R_{T(on)} I_i^2 \right) + \left(V_{T0} I_i + \frac{4\sqrt{2}}{3\pi} \cdot R_{T(on)} I_i^2 \right) \cdot \frac{U_i \cos(\varphi)}{V_{dc}}$$

$$\bar{P}_{Di,cond} = \left(\frac{\sqrt{2}}{\pi} \cdot V_{D0} I_i + \frac{1}{2} \cdot R_{D(on)} I_i^2 \right) - \left(V_{D0} I_i + \frac{4\sqrt{2}}{3\pi} \cdot R_{D(on)} I_i^2 \right) \cdot \frac{U_i \cos(\varphi)}{V_{dc}}$$

Example :

- $V_{to} = 0.95; \% [V]$
- $V_{do} = 1.65; \% [V]$
- $R_{t_on} = 0.5/300; \% [Ohm]$
- $R_{d_on} = 0; \% [Ohm]$
- $E_{d_rr} = 0.0485; \% [J]$
- $E_{on} = 26e-3; \% [J]$
- $E_{off} = 55.5e-3; \% [J]$



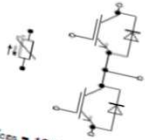
- $V_{dc_n} = 600; \% [V]$
- $I_n = 450; \% [A]$

- $U_{dc} = 600; \% [V]$
- $P_{max} = 200000; \% [W]$

Technische Information / Technical Information
FF450R12ME4

IGBT-Module
 IGBT-modules

EconoDUAL™3 Modul mit Trench/Feldstopp IGBT4 und Emittor Controlled HE Diode und NTC
 EconoDUAL™3 module with Trench/Feldstopp IGBT4 and Emittor Controlled HE diode and NTC

$V_{CES} = 1200V$
 $I_{C,nom} = 450A / I_{C,DM} = 900A$

Typische Anwendungen

- Motorantriebe
- Servoantriebe
- USV-Systeme
- Windgeneratoren

Typical Applications

- Motor Drives
- Servo Drives
- UPS Systems
- Wind Turbines

Elektrische Eigenschaften

- Niedriges $V_{CE(sat)}$
- $T_{d,sp} = 150^{\circ}C$

Electrical Features

- Low $V_{CE(sat)}$
- $T_{d,sp} = 150^{\circ}C$

Mechanische Eigenschaften

- Standardgehäuse

Mechanical Features

- Standard Housing

Module Label Code
 Barcode Code 128

DMX - Code

Content of the Code

Digit	Content
1 - 5	Module Serial Number
6 - 11	Module Material Number
12 - 19	Production Order Number
20 - 21	Datecode (Production Year)
22 - 23	Datecode (Production Week)

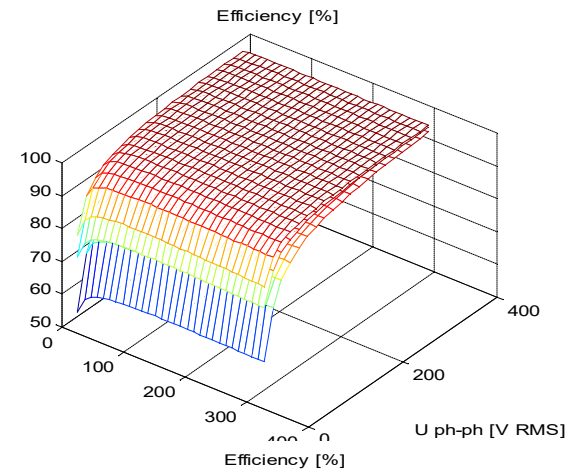
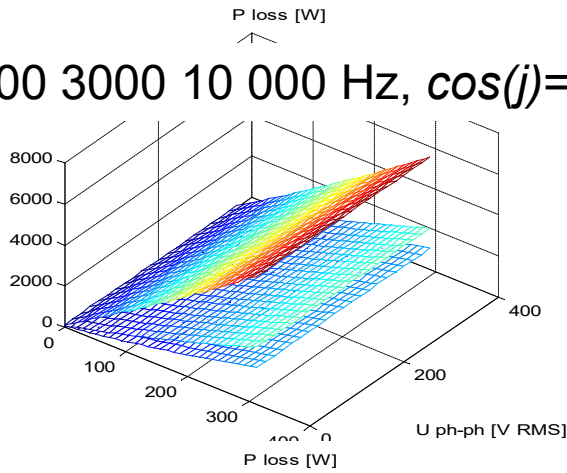
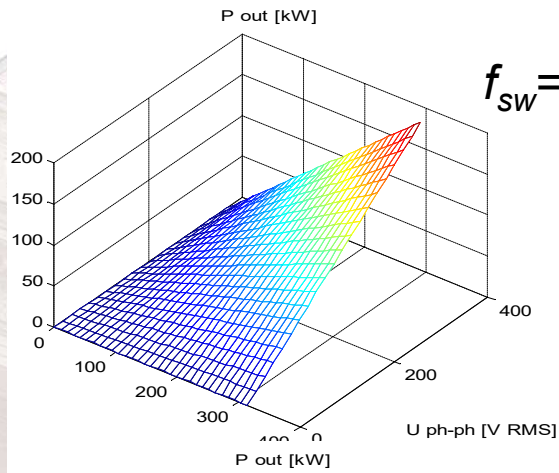
Prepared by: CU
 Approved by: MB

Date of publication: 2013-11-04
 Revision: 3.1

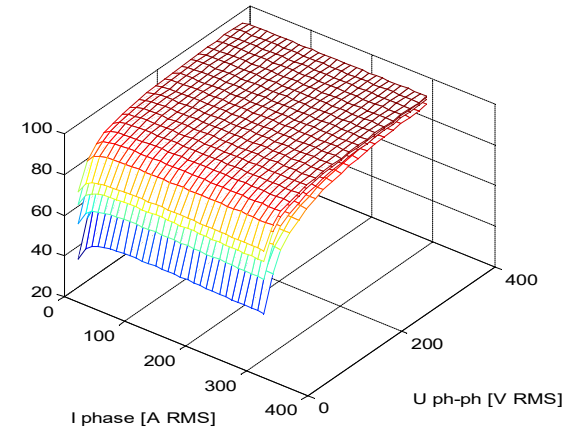
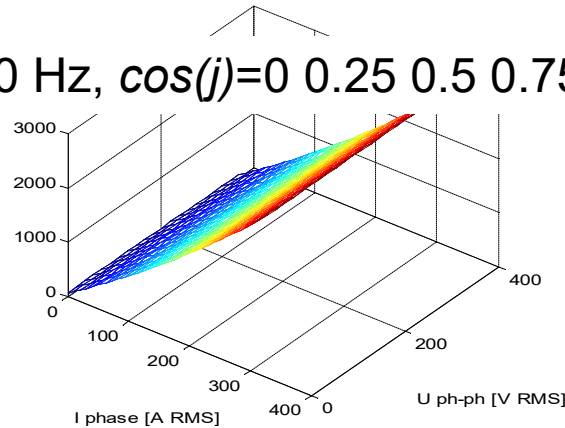
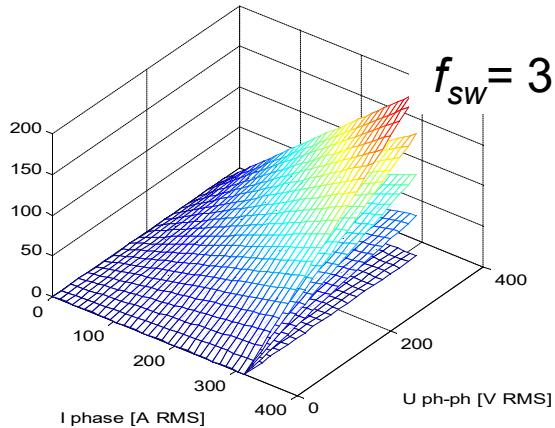
UL approved (E92335)

Converter Efficiency for different f_{sw} & $\cos(\varphi)$

$f_{sw} = 1000 \ 3000 \ 10 \ 000 \text{ Hz}, \cos(j) = 1$



$f_{sw} = 3000 \text{ Hz}, \cos(j) = 0 \ 0.25 \ 0.5 \ 0.75 \ 1$



Power Electronic Efficiency

- Mostly depending on the ratio

$$\frac{\text{Output voltage}}{\text{DC link voltage}}$$

- Almost constant over wide operating range
- Can be represented by a constant, e.g. 0.97

